

# **Intel® 81341 and Intel® 81342 I/O Processors**

**Design Guide**

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*May 2007*



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Date	Revision	Description
January 2007	002	Updated product name terminology Updated signal descriptions and names in Chapter 11
September 2006	001	Initial release







## 1.0 Introduction

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### 1.1 About This Document

This document provides layout information and guidelines for designing platform or add-in board applications with Intel® 81341 and Intel® 81342 I/O Processors.

It is recommended that this document be used as a guideline. Intel recommends employing best-known design practices using board-level simulation, signal integrity testing and validation to create a robust design. Designers, note that this guide focuses on specific design considerations for this part and is not intended to be an all-inclusive list of good design practices. It is recommended that this guide is used in conjunction with empirical data to optimize the particular design.

The simulation conditions used for each of the interfaces are listed in the [Appendix A, "Terminology, Definitions and Conditions"](#). The simulations were performed for motherboard and adapter card topologies. The impedance used for the motherboard is 50 ohm +/- 15% and the adapter card trace impedance is 60 ohm +/- 15%. These results are based on the six layer board stackup that is provided in [Chapter 3.0](#).



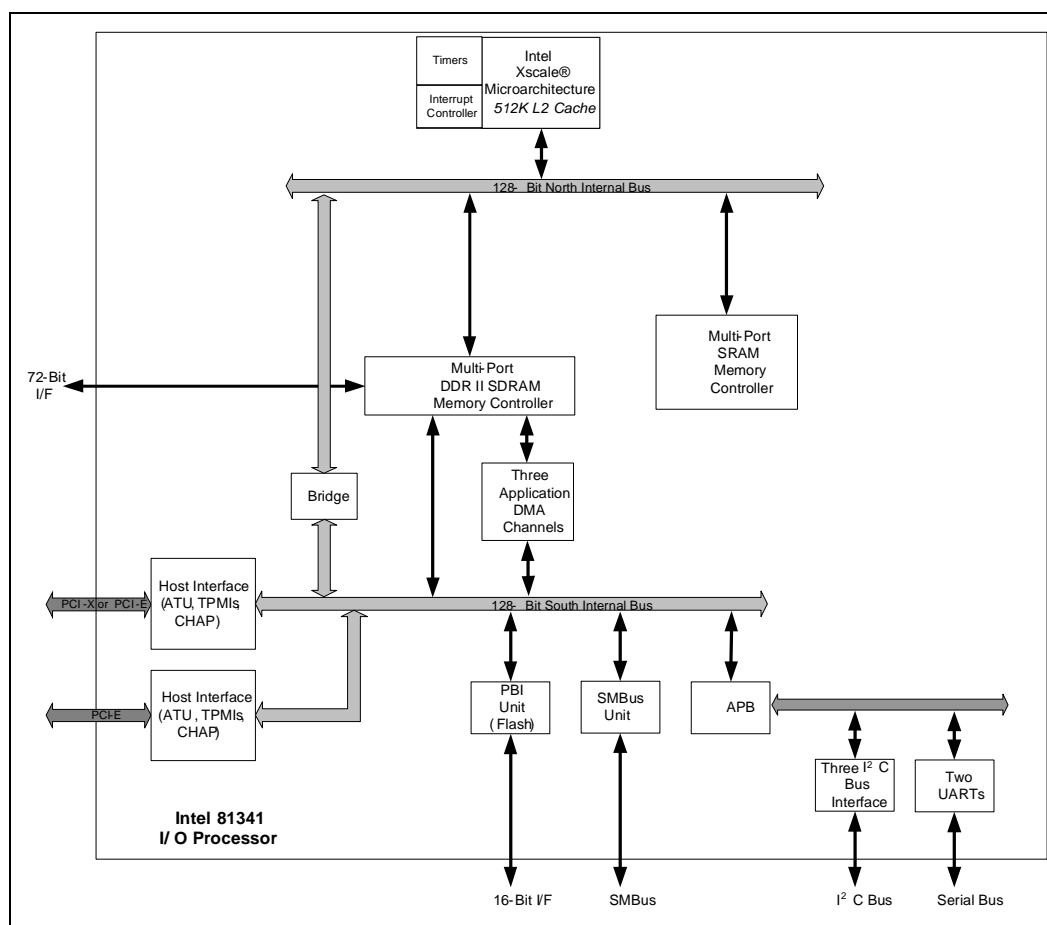
This document is partitioned into the following chapters:

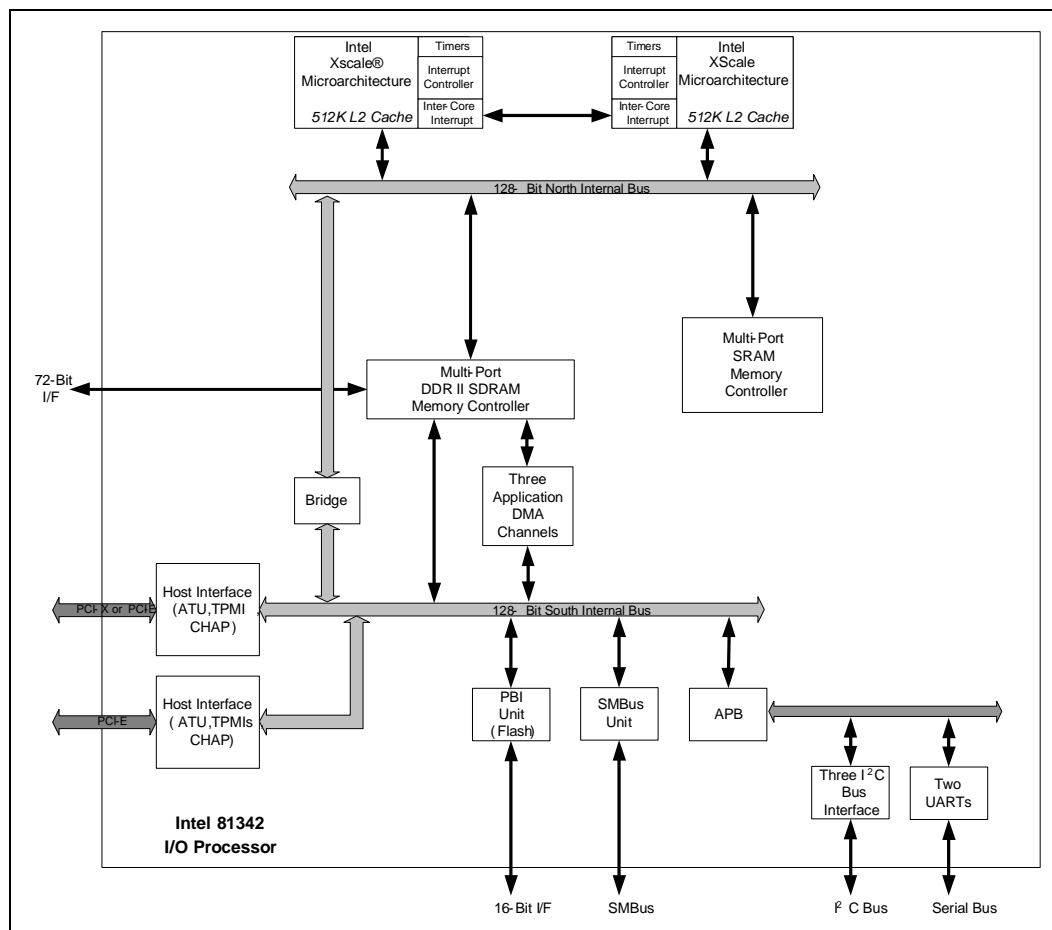
- The top level block diagram and package dimensions are provided in [Chapter 2.0, "Package Information"](#).
- The example stackups for a motherboards and adapter cards are provided in [Chapter 3.0, "Board Layout Guidelines"](#).
- The layout guidelines external interfaces are listed in the following chapters: [Chapter 6.0, "PCI-X Layout Guidelines"](#), [Chapter 5.0, "PCI Express Layout"](#), [Chapter 4.0, "Memory Controller"](#) and [Chapter 7.0, "Peripheral Local Bus"](#).
- The required terminations and reset straps configurations are listed in [Chapter 11.0, "Terminations"](#). This chapter also details the recommended filtering.
- The summary of the layout guidelines for each of the interfaces and the filters is listed in [Chapter 12.0, "Layout Checklist"](#).
- The details on power sequencing and decoupling recommendations are provided in [Chapter 10.0, "Power Delivery"](#).
- The JTAG information is listed in [Chapter 8.0, "JTAG Circuitry for Debug"](#). The details on test equipment are listed in [Chapter 9.0, "Debug and Test"](#).
- The references are listed in [Chapter 13.0, "References"](#).
- The definitions and the simulation conditions (used for all the simulations described in this document) are provided in [Appendix A](#).
- The details on recommended heatsink solutions are listed in Intel® 81341 and Intel® 81342 I/O Processors *Thermal Application Note*.

The Intel® 81341 and Intel® 81342 I/O Processor PCI Bus is capable of 133 MHz operation in PCI-X mode as defined by the *PCI-X Addendum to the PCI Local Bus Specification*, Revision 1.0b. Also, the processor supports a 66 MHz conventional PCI mode as defined by the *PCI Local Bus Specification* Revision 2.2. Intel® 81341 and Intel® 81342 I/O Processor support PCI Express interface lane widths of x1, x2, x4 and x8.

This I/O processor is available as the single processor, Intel® 81341 I/O Processor (Figure 1) or two processor version, Intel® 81342 I/O Processor (Figure 2). Intel® 81341 and Intel® 81342 I/O Processors provide a dual interface that supports both PCI-X 1.0b and PCI Express. When PCI-X 1.0b is selected as the upstream (host) I/O interface, PCI Express is available as a private (not visible to the host), downstream I/O interface. Likewise, when PCI Express is selected as the upstream I/O interface, PCI-X 1.0b is available as a private, downstream I/O interface. The selection of the upstream I/O interface is a reset strap option.

**Figure 1. Single-processor Intel® 81341 I/O Processor Functional Block Diagram**

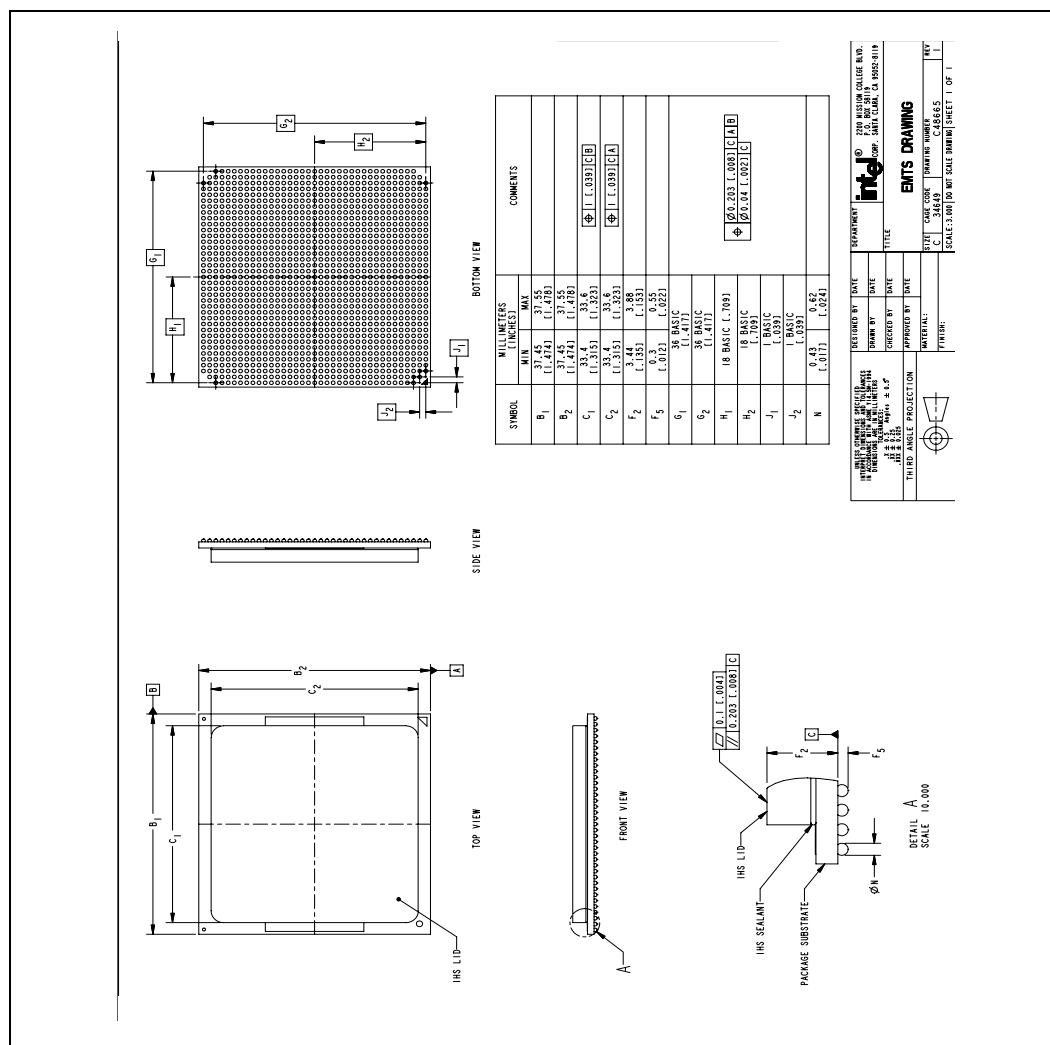



**Figure 2. Two-processor Intel® 81342 I/O Processor Functional Block Diagram**


## 2.1 Package Introduction

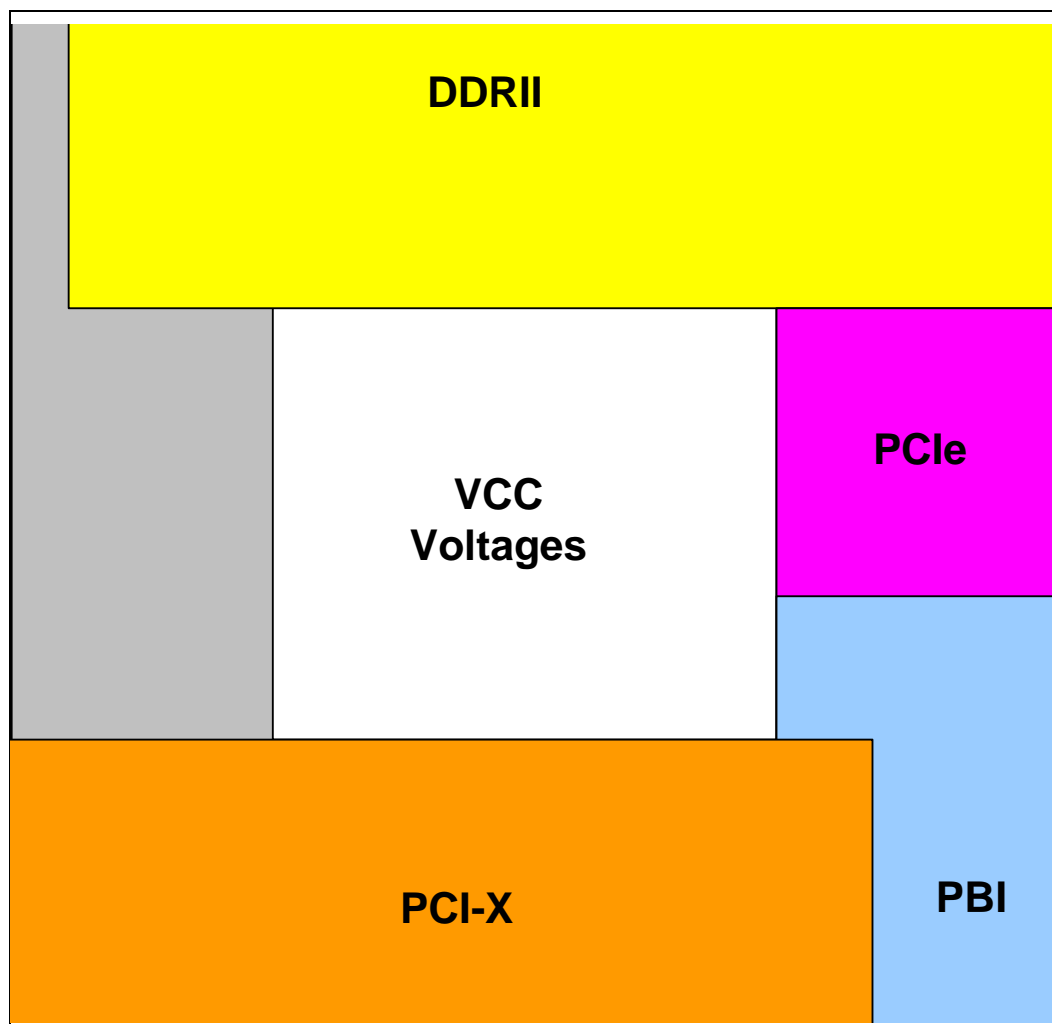
Figure 4 shows the top view of the package with the interfaces labeled and color coded. This figure is helpful during board layout. The signals are located on the FCBGA package to simplify signal routing and system implementation.

**Figure 3. Intel® 81341 and 81342 I/O Processors 1357-ball FCBGA Package Diagram**





**Figure 4. Top View Ball Map With Interfaces**





## 3.0 Board Layout Guidelines

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This chapter provides an example of a motherboard and adapter card stackup implementation. This stackup was used for all simulations listed in this design guide. It is highly recommended that signal integrity simulations be conducted to verify each PCB layout. This is especially true when the layout deviates from the recommendations listed in these design guidelines.





### 3.1 Motherboard Stack Up Information

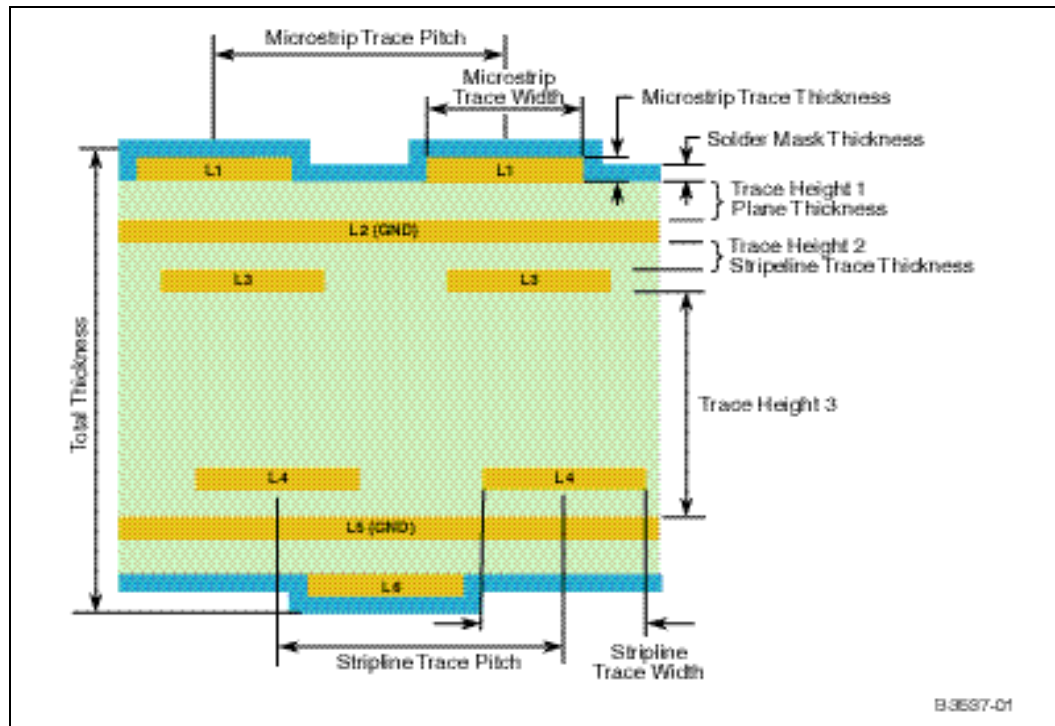
When the 81341 and 81342 are used in server and workstation Raid On Mother Board (ROMB) applications, the motherboard is implemented on six layers. The specified impedance range for all board implementations is 50ohms +/-15%. Adjustments are made for interfaces specified at other impedances. Table 1 defines the typical layer geometries for a six layer board.

The motherboard impedance guidelines are based on the typical server/workstation impedance for their processor and memory subsystem of 50-ohms. Dimensions and tolerances for the motherboard are listed in Table 1. Refer to Figure 5 for location of variables in Table 1.

**Table 1. Motherboard Stack Up, Stripline and Microstrip**

Variable	Type	Nominal	Minimum	Maximum	Notes
Solder Mask Thickness (mil)	N/A	0.8	0.6	1.0	
Solder Mask $E_r$	N/A	3.65	3.65	3.65	
Core Thickness (mil)	N/A	9.8	9.6	10	
Core $E_r$	N/A	4.30	3.75	4.85	2113 material
Plane Thickness (mil)	Power	2.7	2.5	2.9	
	Ground	1.35	1.15	1.55	
Trace Height (mil)	1	3.5	3.3	3.7	The trace height is determined to achieve a nominal 50 ohms.
	2	3.5	3.3	3.7	
	3	10.5	9.9	11.1	
Preg $E_r$	Microstrip	4.30	3.75	4.85	
	Stripline1	4.30	3.75	4.85	
	Stripline2	4.66	4.19	5.13	
Trace Thickness (mil)	Microstrip	1.75	1.2	2.3	
	Stripline	1.4	1.2	1.6	
Trace Width (mil)	Microstrip	5.0	3.5	6.5	
	Stripline	4.0	2.5	5.5	
Trace Spacing (mil)	Microstrip	15.0	-	-	Each interface sets the trace spacing based on its signal integrity of differential impedance requirements. For the purposes of the building the transmission line models, it is assumed the artwork is very accurate and therefore a constant. Thus, all the variability in the trace spacing is the result of the tolerances of the trace width.
	Stripline	12.0	-	-	
Total Thickness (mil)	FR4	62.0	56.0	68.0	
Trace Velocity (ps/in)	Microstrip		135	141	Velocity varies based on variation in $E_r$ . It is not controlled during the fab process.
	Stripline		167	178	
Trace Impedance (ohms)	Microstrip	50	42.5	57.5	
	Stripline	50	45	55	

**Figure 5. Motherboard Stackup Recommendations**





## 3.2 Adapter Card Topology

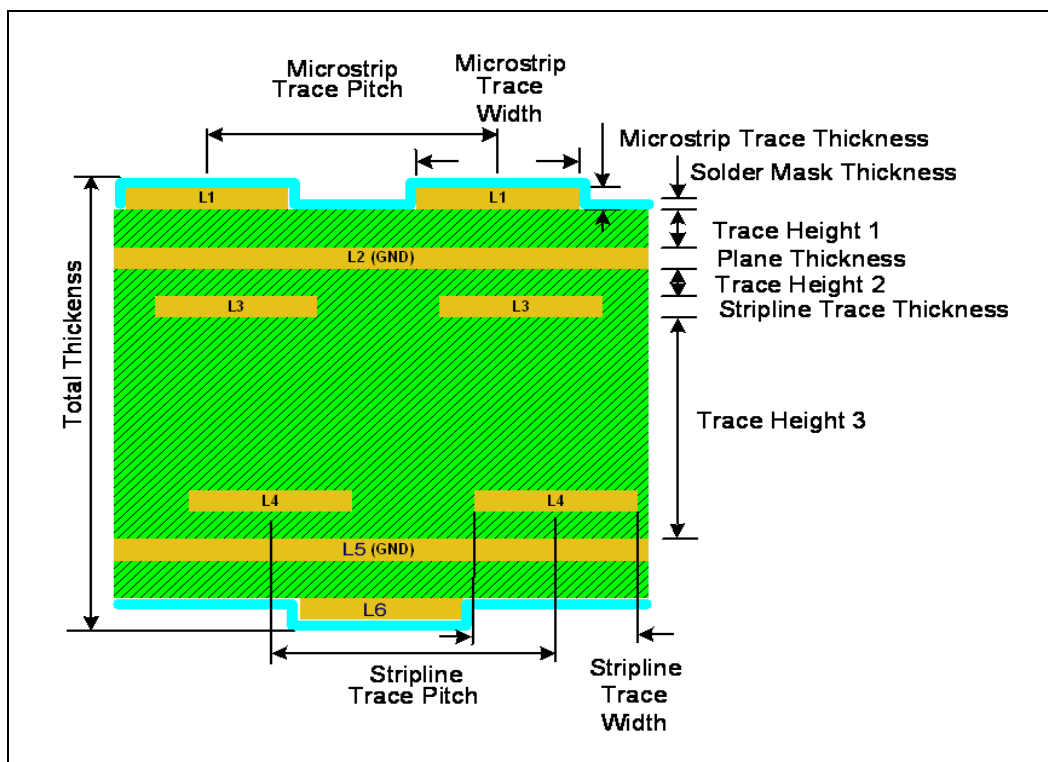
Intel® 81341 and Intel® 81342 I/O Processor are implemented on PCI Express or PCI-X adapter cards with six layers. The specified impedance range for all adapter card implementations is 60ohms +/-15%. [Table 2](#) defines the typical layer geometries for a six layer board. Note that the values are the same as the motherboard stack up with the exception of the impedance.

**Table 2. Adapter Card Stack Up, Microstrip and Stripline**

Variable	Type	Nominal	Minimum	Maximum	Notes
Solder Mask Thickness (mil)	N/A	0.8	0.6	1.0	
Solder Mask $E_r$	N/A	3.65	3.65	3.65	
Core Thickness (mil)	N/A	2.8	3.0	3.2	
Core $E_r$	N/A	4.3	3.75	4.85	2113 material
Plane Thickness (mil)	Power	2.7	2.5	2.9	
	Ground	1.35	1.15	1.55	
Trace Height (mil)	1	3.5	3.3	3.7	The trace height is determined to achieve a nominal 60 ohms.
	2	7.0	6.7	7.3	
	3	7.0	6.7	7.3	
Preg $E_r$	Microstrip	4.30	3.75	4.85	2113 material
	Stripline1	4.30	3.75	4.85	
	Stripline2	4.66	4.19	5.13	
Trace Thickness (mil)	Microstrip	1.75	1.2	2.3	
	Stripline	1.4	1.2	1.6	
Trace Width (mil)	Microstrip	4.0	2.5	5.5	
	Stripline	4.0	2.5	5.5	
Total Thickness (mil)	FR4	62.0	56.0	68.0	
Trace Velocity (ps/in)	Microstrip		135	141	Velocity varies based on variation in $E_r$ . It is not controlled during the fab process.
	Stripline		167	178	
Trace Impedance	Microstrip	60	51	69	
	Stripline	60	51	69	

**Note:** Each interface sets the trace spacing based on its signal integrity of differential impedance requirements. For the purposes of the building the transmission line models, it is assumed the artwork is very accurate and therefore a constant. Thus, all the variability in the trace spacing is the result of the tolerances of the trace width.

**Figure 6. Adapter Card Stackup**





### 3.3 PCB Impedance Targets

Table 3 provides the impedance ranges and the associated trace dimensions for single-ended and differential traces. Figure 5 shows an example of a differential trace.

**Table 3. Single-ended Trace Parameters**

Single Line						
Topology	Ohms	Actual Impedance Range			Width (mils)	Spacing (mils)
		Min	Max	Nominal		
Stripline	50	44.17	57.47	50.82	4	N/A
Stripline	60	51.16	66.62	58.89	4	N/A
Microstrip	50	42.97	57.46	50.22	5	N/A
Microstrip	60	51.30	67.89	59.60	4	N/A

**Table 4. Differential Trace Dimensions**

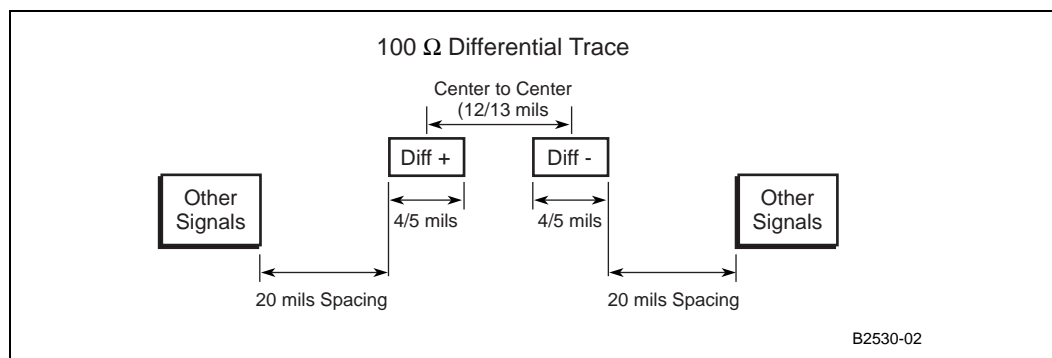
Differential Pair						
Topology	Ohms	Actual Impedance Range			Width (mils)	Edge-to-Edge Spacing (mils)
		Min	Max	Nominal		
Stripline	85	74.24	102.28	92	4	8
Stripline	100	87.06	121.84	100	4	8
Microstrip	85	71.56	119.36	88	5	7
Microstrip	100	80.36	114.28	100	4	8

#### 3.3.1 100 Ohm Differential Trace

The Figure 7 shows a 100 ohm differential trace constructed from various topologies based on the stackup listed in this chapter. These differential traces are used to route the DQS and clock lines.

1. Using two striplines of trace width 4 mils separated by 8 mils edge to edge (12 mils center to center).
2. Using two microstrips of trace width 5 mils separated by 8 mils edge to edge (13 mils center to center).

**Figure 7. An Example of 100 Ohm Differential Trace**



## 4.0 Memory Controller

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This chapter describes how to layout the physical memory interface for Intel® 81341 and Intel® 81342 I/O Processors.

### 4.1 Overview

The Intel® 81341 and Intel® 81342 I/O Processors integrates a high performance, multi-ported memory controller to provide a direct interface between Intel® 81341 and Intel® 81342 I/O Processors and its local memory subsystem. The Memory Controller supports:

- PC3200 and PC4300 Double Data Rate II (DDR2) Registered and Unbuffered DDR2 400MHz and DDR2 533MHz SDRAM
- 512 Mbit and 1 Gbit DDR2 SDRAM technology support
- Registered and Unbuffered DDR2 DIMM support
- Dedicated port for Intel XScale® processors to DDR2 SDRAM
- Between 256 MBytes and 2 GBytes of 64-bit DDR2 SDRAM
- 36-bit addressable
- Optimized core processor data processing 32-bit region
- Generation and/or Verification of Block Guard Data Integrity fields embedded in the data stream
- Single-bit error correction, multi-bit detection support (ECC)
- 32-, 40- and 64-, 72-bit wide Memory Interfaces (non-ECC and ECC support)
- The memory controller provides two chip enables to the memory subsystem. These two chip enables service the DDR2 SDRAM subsystem (one per bank).
- For 64-bit ECC memory, a 32-bit memory region is programmed to operate as 32-bit ECC memory for higher core write performance by avoiding Read-Modify-Write (RMW) operation of DDR2 SDRAM.
- One or two banks of DDR2 SDRAM (in the form of one two-bank dual inline memory module).



## 4.2 DDR2 533 Layout Guidelines

This section provides the DDR2 533 layout guidelines for both DIMM topology (for motherboard and adapter card topologies) and embedded memory down (for motherboard topology). For a DDR2 400 layout the same DDR2 533 layout guidelines are used.

- [Section 4.2.2](#) provides details on the DDR2 533 DIMM routing guidelines.
- [Section 4.2.4](#) provides details on the DDR2 533 embedded routing guidelines.

### 4.2.1 DDR2 533 DIMM Layout Guidelines

This section provides the layout guidelines for a DIMM topology for DDR2 533.

The DDR interface is divided up into three groups that each have special routing guidelines:

1. Source synchronous signal group: DQ/DQS/DQM/CB signals, [Section 4.2.2.1](#).
2. Clocked: M\_CLK signals, 6 clocks, three positive (**M\_CLK[2:0]**) and three negative (**M\_CLK[2:0]#**).
3. The 72-bit 2-bank unbuffered DDR SDRAM DIMM specification requires 6 clocks to distribute the loading across eighteen x8 DDR SDRAM components.
4. Control signals: Address/RAS/CAS/CS/WE/CKE/ODT signals, [Section 4.2.3](#).

The On Die Termination or ODT for DDR2 eliminates some of the termination resistors needed for the source synchronous signals.

The [Table 5](#) and [Table 6](#) list the DDR2 differential strobe alignment with each of the DQ groups.

**Table 5. x64 DDR Memory Configuration**

Data Group	Positive Strobe	Negative Strobe
DQ[7:0], DM[0]	DQS0	DQS0#
DQ[15:8], DM[1]	DQS1	DQS1#
DQ[23:16], DM[2]	DQS2	DQS2#
DQ[31:24], DM[3]	DQS3	DQS3#
DQ[39:32], DM[4]	DQS4	DQS4#
DQ[47:40], DM[5]	DQS5	DQS5#
DQ[55:48], DM[6]	DQS6	DQS6#
DQ[63:56], DM[7]	DQS7	DQS7#

**Table 6. x72 DDR Memory Configuration**

Data Group	Positive Strobe	Negative Strobe
DQ[7:0], DM[0]	DQS0	DQS0#
DQ[15:8], DM[1]	DQS1	DQS1#
DQ[23:16], DM[2]	DQS2	DQS2#
DQ[31:24], DM[3]	DQS3	DQS3#
DQ[39:32], DM[4]	DQS4	DQS4#
DQ[47:40], DM[5]	DQS5	DQS5#
DQ[55:48], DM[6]	DQS6	DQS6#
DQ[63:56], DM[7]	DQS7	DQS7#
CB[7:0], DM[8]	DQS8	DQS8#

## 4.2.2 DDR2 533 DIMM Layout Design

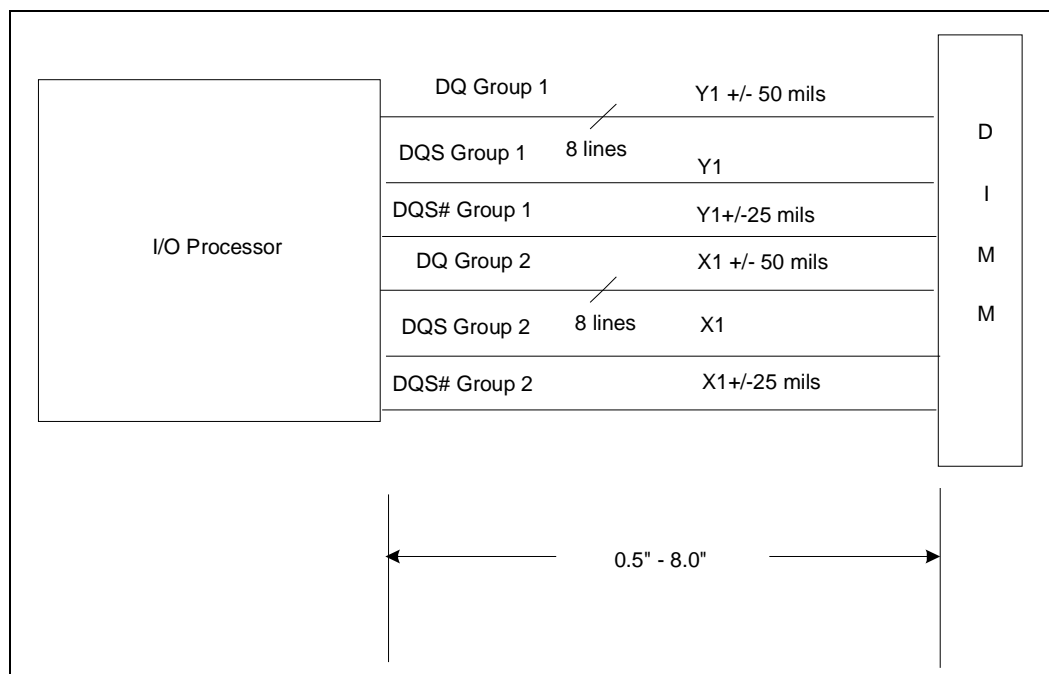
This section provides the source synchronous, clock and control layout guidelines for DDR2 533 unbuffered and registered DIMMs. The topologies that were simulated for this revision of the document include RAW Card A, B, C and registered DIMMs. Refer to the JEDEC specification for more details on these topology at <http://www.jedec.org>.

### 4.2.2.1 DDR2 DIMM Source Synchronous Routing

This section lists the recommendations for the DDR2 Source Synchronous Routing. These signals include all the DQ/DQS/DM/CB signals.

- Refer to [Figure 8](#) for a block diagram of the DQ and DQS group length matching relationship.
- Refer to [Figure 9](#) for a block diagram of the DQ/DQS group and length matching relationship with respect to the clock M<sub>CK</sub>/M<sub>CK</sub># signals.
- Refer to [Figure 10](#) for segment lengths of the DQ lines and [Figure 11](#) for the segment lengths of the DQS lines.
- [Table 7](#) lists the routing recommendations for DQ/DQS lines. [Table 8](#) lists the segment lengths for the DQ lines and [Table 9](#) lists segment lengths for the DQS lines.

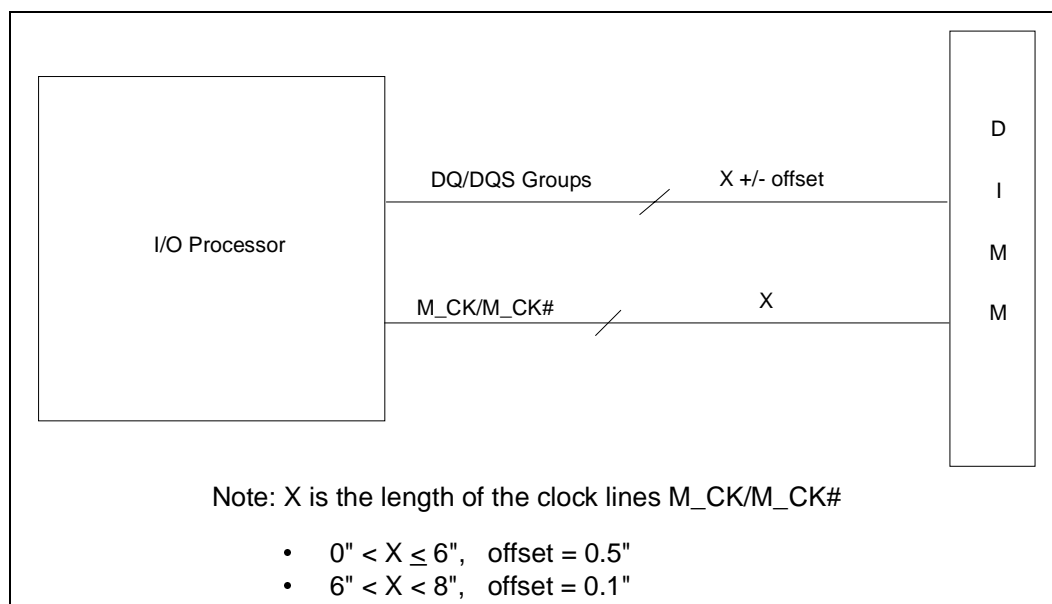
**Figure 8. DDR2 DIMM Source Synchronous Routing**







**Figure 9. DDR2 DIMM Length Matching DQ/DQS Group with Respect to Clocks M\_CK/M\_CK#**



**Table 7. DDR2 DIMM Source Synchronous Routing Recommendations**

Parameter	Routing Guideline
Reference Plane	Route over unbroken ground plane or unbroken power plane.
Preferred Layer	Stripline
Breakout	5 mils width 5 mils spacing. Maximum length of breakout region $\leq$ 500 mils microstrip
DQ signals Trace Impedance	Single ended stripline lines: 50 ohms $\pm$ 15% impedance for motherboards 60 ohms $\pm$ 15% impedance for Add-in cards
DQS Signals Trace Impedance	Differential stripline: Differential 85ohm $\pm$ 15% impedance for motherboards. Differential 100 ohm $\pm$ 15% impedance for add-in cards
DQ Group Spacing (edge to edge) <sup>1</sup>	Spacing within the same group: 12 mils minimum Spacing from other DQ groups: 20 mils minimum For DQS from any other signals: 20 mils minimum
Overall Trace Length: signal Ball to DIMM connector	0.5" minimum to 8" maximum (correlated with the clock length from ball to DIMM).
DQS Length Matching: Trace Length Matching within DQS group Within one DQS pair plus and minus	$\pm$ 0.05" within DQS group $\pm$ 0.0250"
Length Matching: <sup>1</sup> DQS with respect to clock (from controller to DIMM connector)	Total Length: 0" < total length $\leq$ 6", matching $\leq$ $\pm$ 0.5" 6" < total length $\leq$ 8", matching $\leq$ $\pm$ 0.1"
Number of Vias	$\leq$ 2 (for differential signals the number of vias on + and - signals must be the same)
DQ and DQS ODT	150 ohm ODT on Intel® 81341 and Intel® 81342 I/O Processors 75 ohm ODT on DRAM
Routing Guideline	Route all data signals and their associated strobes on the same layer.

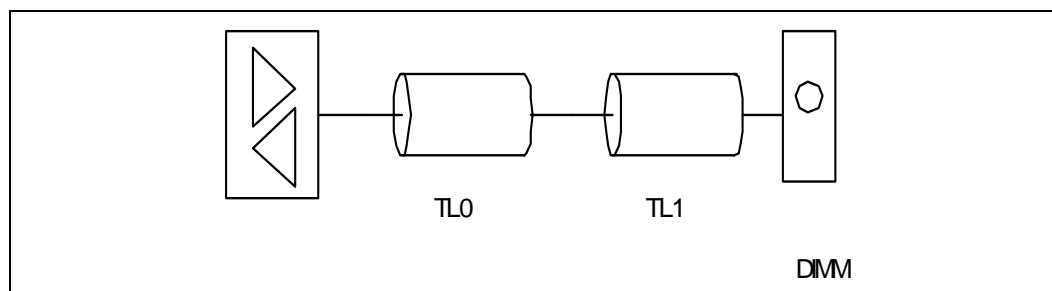
<sup>1</sup> For a right angle DDR connector consideration must be given to the lead length skew across the connector. Refer to [Table 62, "Right Angle Connector Skews \(length matching compensation\)"](#) on page 144.

**Table 8. DDR2 DIMM DQ Lengths**

Traces	Description	Layer	Min Length	Max Length	Trace Impedance	Spacing (edge to edge)	Notes
TL0	Breakout	Microstrip	0"	0.5"		5 mils	5 mils trace width OK for breakout.
TL1	Lead-in	Stripline	0.5"	8"	50 ohms $\pm$ 15% impedance for motherboards 60 ohms $\pm$ 15% impedance for Add-in cards	Within same group $\geq$ 12 mils Between other groups $\geq$ 20 mils	

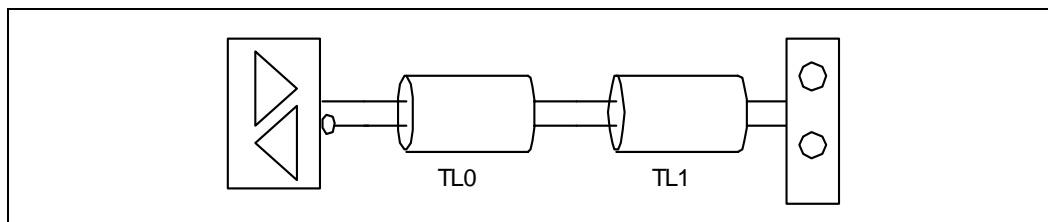


**Figure 10. DDR2 DIMM DQ Topology**



**Table 9. DDR2 DIMM DQS Lengths**

Traces	Description	Layer	Min Length	Max Length	Trace Impedance	Spacing (edge to edge)	Notes
TL0	Breakout	Microstrip	0"	0.5"		5 mils	5 mils trace width OK for breakout.
TL1	Lead-in	Stripline	0.5"	8"	Differential 85ohm +/- 15% impedance for motherboard s. Differential 100 ohm +/- 15% impedance for add-in cards	8 mils spacing (edge to edge) for 4 mil differential stripline trace. See <a href="#">Section 3.3</a> for details on differential routing. ≥ 20 mils from other signals	Route as differential pair

**Figure 11. DDR2 DIMM DQS Topology**




This section lists the recommendations for the DDR2 Clock signals.

- Refer to [Table 10](#) for the DIMM clock routing guidelines and [Table 11](#) for the DIMM clock segment lengths.
- Refer to [Figure 12](#) for the clock topology segment lengths.
- Refer to [Figure 9](#) for a block diagram of the DQ/DQS group and length matching relationship with respect to the clock signals.
- Refer to [Figure 13](#) for the Address/Command length matching relationship with respect to clock signals.

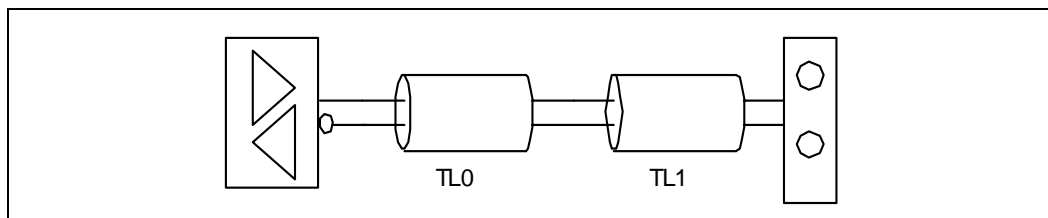
**Table 10. DDR2 DIMM Clock Routing Recommendations**

Parameter	Routing Guideline
Reference Plane	Route over unbroken ground plane preferred
Preferred Topology	Microstrip differential lines preferred
Breakout Trace Width and spacing	5 mils by 5 mils microstrip or stripline. Maximum length of breakout trace is 500 mils.
Trace Impedance	Differential impedance of 85 ohms +/- 15% motherboard Differential impedance of 100 ohms +/- 15% add-in card
Trace Spacing (edge to edge)	≥ 25 mils between other signals.
Trace Length : TL0 + TL1: signal Ball to DIMM connector	0.5" min to 8.0" max
Length Matching: <sup>1</sup> Within M_CK/M_CK# (differential clock signals)	+/- 0.0250" within pairs (intra-pair)
Length Matching: <sup>1</sup> With respect to DQS (from controller to DIMM connector):	Total Length: 0 < total length ≤ 6", matching ≤ +/- 0.5" 6" < total length ≤ 8", matching ≤ +/- 0.1"
Length Matching: With respect to address/command group (from controller to DIMM connector)	+8"/-3" <sup>1</sup> maximum for motherboard and +8"/-2" maximum for add-in card
Length Matching: With respect to CS/CKE group	+/-2" maximum for motherboard and +1"/-3" maximum for add-in card
Routing Guideline 1	Maximum of 1 via/layer change for M_CK/M_CK# clocks. (use the same number of vias between + and - signals of differential clock)

1. For a right angle, DDR connector consideration must be given to the lead length skew across the connector. Refer to [Table 62, "Right Angle Connector Skews \(length matching compensation\)"](#) on page 144.

**Table 11. DDR2 DIMM Clock Lengths**

Traces	Description	Layer	Min Length	Max Length	Trace Impedance	Spacing (edge to edge)	Notes
TL0	Breakout	Microstrip or stripline	0"	0.5"		5 mils	5 mils trace width OK for breakout.
TL1	Lead-in	Microstrip	0.5"	8"	Differential impedance of 85 ohms +/- 15% motherboard Differential impedance of 100 ohms +/- 15% add-in card	See <a href="#">Section 3.3</a> for details on differential routing. Other groups $\geq$ 25 mils	

**Figure 12. DDR2 DIMM Clock Topology**




### 4.2.3 DDR2 Address/Command/Control Routing Guidelines

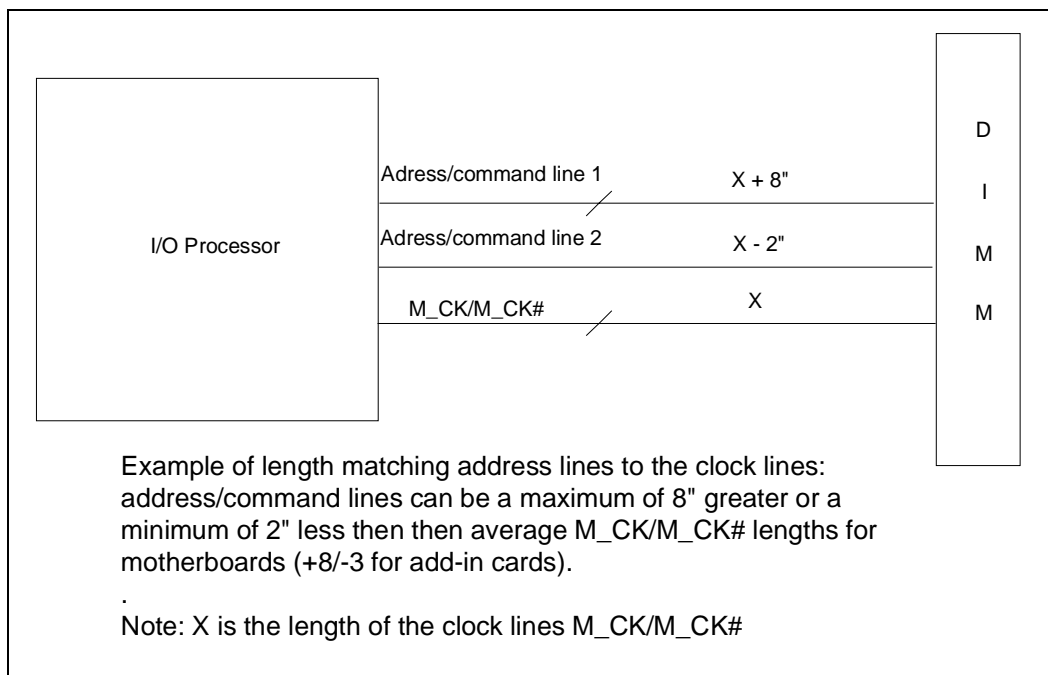
This section lists the recommendations for the DDR2 Address/Command and Control signals.

- Refer to [Figure 13](#) for the Address/Command length matching relationship with respect to clock lines.
- Refer to [Table 12](#) for a description of the Address/Command signals routing guidelines.
- Refer to [Table 13](#) for the Address/Command signals segment length guidelines.
- Refer to [Figure 14](#) for the Address/Command signals segment lengths Vtt topology and refer to [Figure 15](#) for split termination topology.

**Table 12. DDR2 DIMM Address/Command/Control Routing Recommendation**

Parameter	Routing Guideline
Reference Plane	Route over unbroken ground plane preferred
Preferred Topology	Microstrip lines
Breakout Trace Width and Spacing	5 mils x 5mils. Microstrip or stripline is acceptable. Maximum length of the breakout trace is 500 mils.
Trace Spacing (edge to edge)	5 mils acceptable between the pins and the breakout regions. $\geq 12$ mils within group $\geq 20$ mils from any other clock/DQ/DQS groups.
Trace Impedance	50 ohms +/- 15% for a motherboard 60 ohms +/- 15% for a add-in card
Trace Length: Overall length from signal Ball to DIMM Connector	0.5" min to 10" maximum Refer to <a href="#">Table 13</a> for segment lengths.
Length Matching: address/command group (except CS, ODT and CKE lines) with respect to clock (from controller to DIMM connector)	+8"/-3" <sup>1</sup> maximum for motherboard and +8"/-2" maximum for add-in card
Length Matching: CS, ODT and CKE lines with respect to clock (from controller to DIMM connector)	+/-2" maximum for motherboard and +1"/-3" maximum for add-in card
Single Parallel Termination or Split Termination	51.1 ohms +/- 1% to VTT  100 ohms +/- 1% to ground and 100 ohms +/- 1% to 1.8V
Routing Guideline 1	Place the VTT terminations in the VTT island after the DIMM with a trace length of 0.15" to 0.5"
Routing Guideline 2	For split terminations place the VTT termination in their respective power islands
Number of vias	2 Vias or less

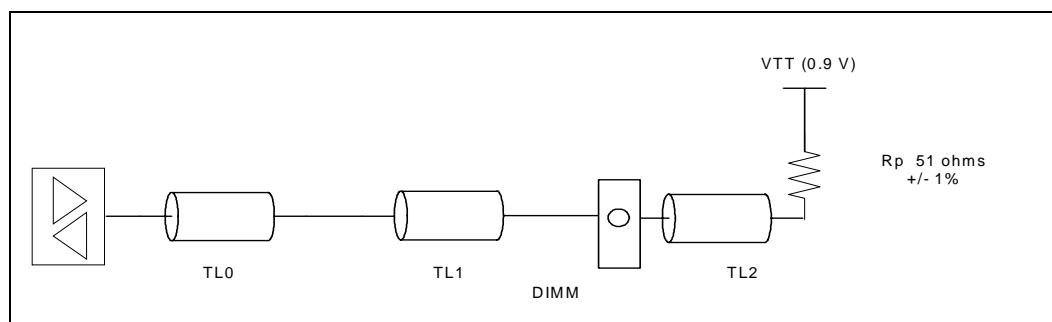
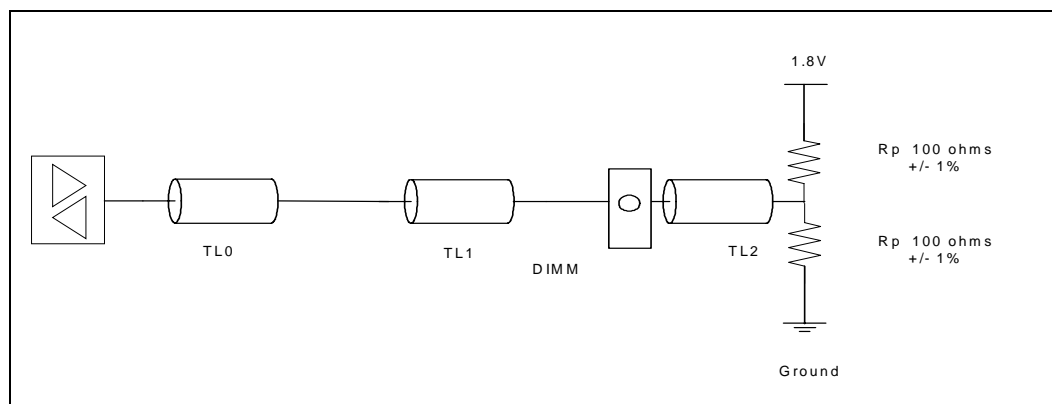
**Figure 13. DDR2 DIMM Length Matching Address/Command Group to Clocks M\_CK/M\_CK#**





**Table 13. DDR2 DIMM Address/Command Lengths**

Traces	Description	Layer	Minimum Length	Maximum Length	Trace Impedance	Spacing (edge to edge)	Notes
TL0	Breakout	Microstrip	0"	0.5"		5 mils	5 mils trace width OK for breakout.
TL1	Lead-in	Microstrip	0.5"	10"	50 +/- 15%	Other groups $\geq$ 20 mils	
TL2	Vtt	Microstrip	0.15"	0.5"			

**Figure 14. DDR2 DIMM Address/CMD Topology (Vtt Termination)****Figure 15. DDR2 DIMM Address/CMD Topology (Split Termination)**

#### 4.2.4 DDR2 533 Embedded Layout Design

This section provides the source synchronous, clock and control layout guidelines for separate DDR2 533 unbuffered memory chips placed on the board (without a DIMM). This analysis is also valid for an embedded DDR2 400 design. The topology that was simulated consisted of four memory chips x16 and one additional x8 for ECC. The simulations were based on 50 ohm +/- 15% motherboard stackup.

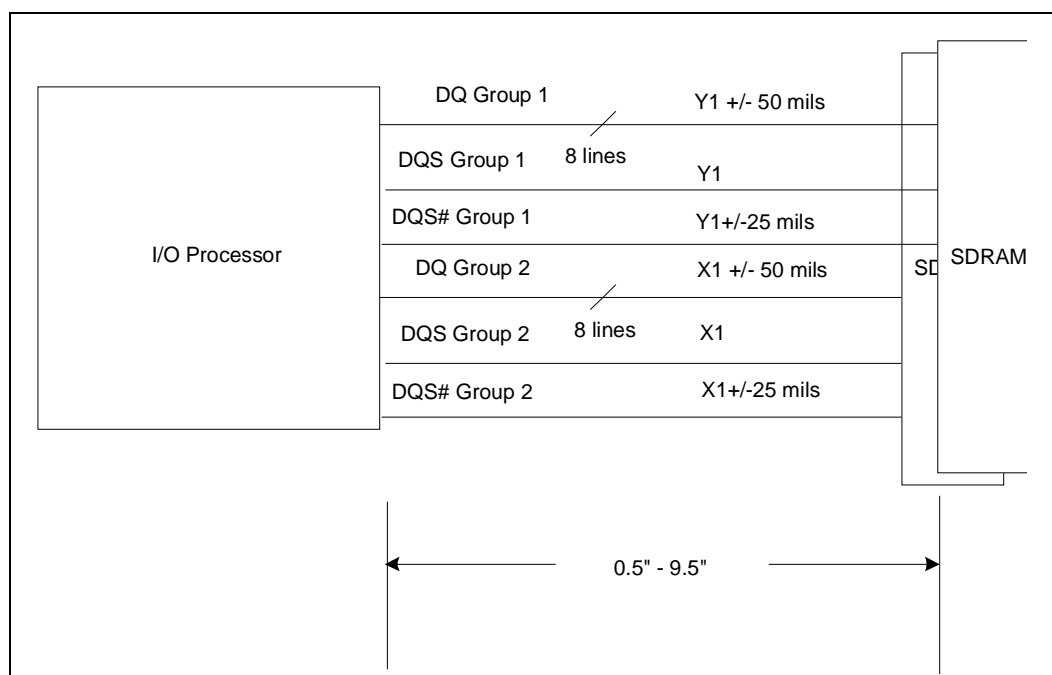
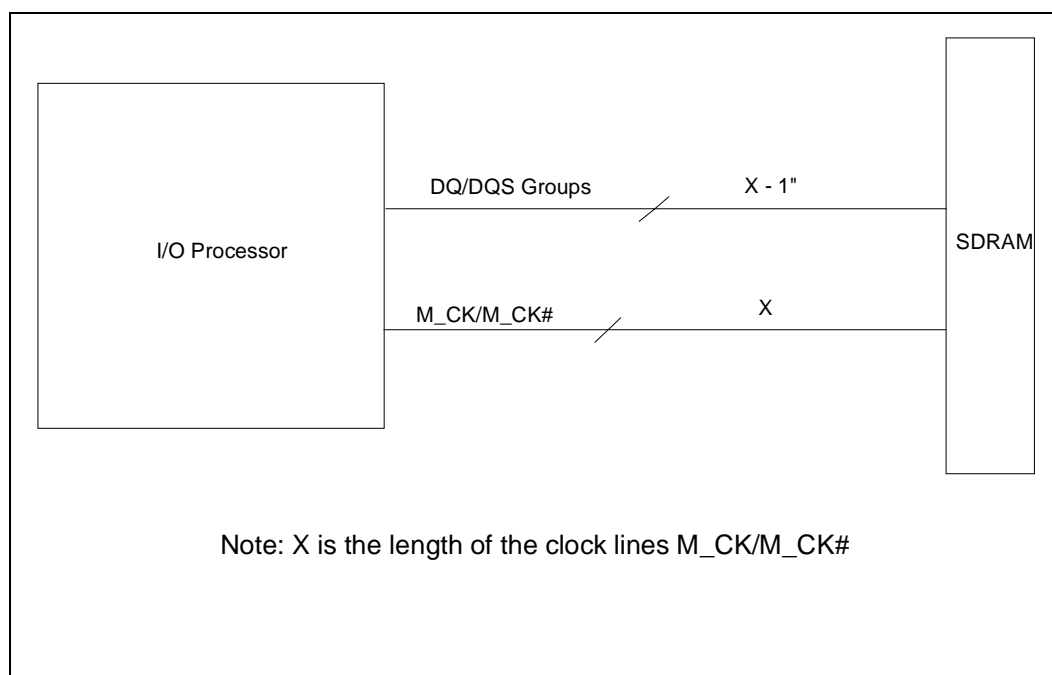
The embedded DDR2 interface is divided up into four groups that each have special routing guidelines:

1. Source synchronous signal group: DQ/DQS/DQM/CB signals, [Section 4.2.4.1](#)
2. Clocked: M\_CLK signals, 6 clocks, three positive (**M\_CLK[2:0]**) and three negative (**M\_CLK[2:0]#**), [Section](#) .
3. Control signals: Address/RAS/CAS//WE, [Section 4.2.4.2](#).
4. Control signals: CKE/CS/ODT signals, [Section](#) .

##### 4.2.4.1 DDR2 Embedded Source Synchronous Routing

This section lists the recommendations for the DDR2 Source Synchronous Routing. These signals include all the DQ/DQS/DM/CB signals.

- Refer to [Figure 16](#) for a block diagram of the DQ and DQS group length matching relationship.
- Refer to [Figure 17](#) for a block diagram of the DQ/DQS group and length matching relationship with respect to the clock M\_CLK/M\_CLK# signals.
- Refer to [Figure 18](#) for segment lengths of the DQ lines and [Figure 19](#) for the segment lengths of the DQS lines.
- [Table 14](#) lists the routing recommendations for DQ/DQS lines. [Table 15](#) lists the segment lengths for the DQ lines and [Table 16](#) lists segment lengths for the DQS lines.


**Figure 16. DDR2 Embedded Source Synchronous Routing**

**Figure 17. DDR2 Embedded Length Matching DQ/DQS Group with Clocks M\_CK/M\_CK#**


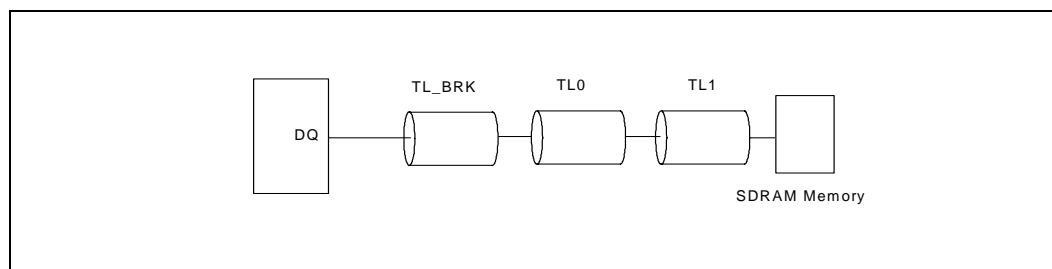


**Table 14. DDR2 Embedded Source Synchronous Routing Recommendations**

Parameter	Routing Guideline
Reference Plane	Route over unbroken ground plane or unbroken power plane.
Preferred Layer	Stripline
Breakout	5 mils width 5 mils spacing. Maximum length of breakout region $\leq$ 500 mils microstrip
DQ signals Trace Impedance	Single ended stripline lines: 50 ohms +/- 15% impedance for motherboards
DQS Signals Trace Impedance	Differential stripline: Differential 85ohm +/- 15% impedance for motherboards.
DQ Group Spacing (edge to edge) <sup>1</sup>	Spacing within the same group: 12 mils minimum Spacing from other DQ groups: 20 mils minimum For DQS from any other signals: 20 mils minimum
Overall Trace Length: signal Ball to memory ball	0.5" minimum to 9.5" maximum (correlated with the clock length from ball to memory).
DQS Length Matching: Trace Length Matching within DQS group Within one DQS pair plus and minus	+/- 0.05" within DQS group +/- 0.0250"
Length Matching:DQS group with respect to clock (from controller to memory chip)	DQS length = clock length - 1" (tolerance +/- 0.1")
Number of Vias	$\leq$ 4 (for differential signals the number of vias on + and - signals must be the same)
DQ and DQS ODT	150 ohm ODT on Intel® 81341 and Intel® 81342 I/O Processors 75 ohm ODT on SDRAM
Routing Guideline	Route all data signals and their associated strobes on the same layer.

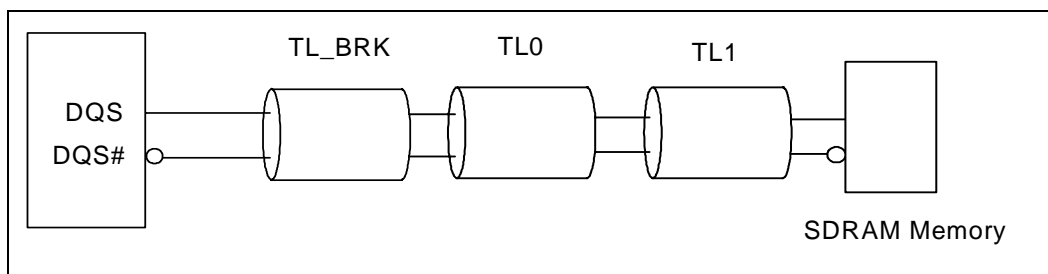
**Table 15. DDR2 Embedded DQ Lengths**

Traces	Description	Layer	Min Length	Max Length	Trace Impedance	Spacing (edge to edge)	Notes
TL_BRK	Breakout	Microstrip	0"	0.5"		5 mils	5 mils trace width OK for breakout.
TL0	Lead-in	Microstrip	0.5"	8"	50 ohms +/- 15% impedance for motherboards	Within same group $\geq 12$ mils Between other groups $\geq 20$ mils	
TL1	SDRAM Lead-in	Microstrip	0.2"	0.75"	"	"	

**Figure 18. DDR2 Embedded DQ Topology**

**Table 16. DDR2 Embedded DQS Lengths**

Traces	Description	Layer	Minimum Length	Maximum Length	Trace Impedance	Spacing (edge to edge)	Notes
TL_BRK	Breakout	Micro-strip	0"	0.5"		5 mils	5 mils trace width OK for breakout.
TL0	Lead-in	Microstrip	0.5"	8"	Differential 85ohm +/- 15% impedance for motherboards.	8 mils spacing (edge to edge) for 4 mil differential stripline trace. See <a href="#">Section 3.3</a> for details on differential routing. ≥ 20 mils from other signals Route as differential pair	Length tolerance +/- 0.1"
TL1	SDRAM Lead-in	Microstrip	0.2"	0.75"	"	"	"

**Figure 19. DDR2 Embedded DQS Topology**




This section lists the recommendations for the DDR2 Clock signals.

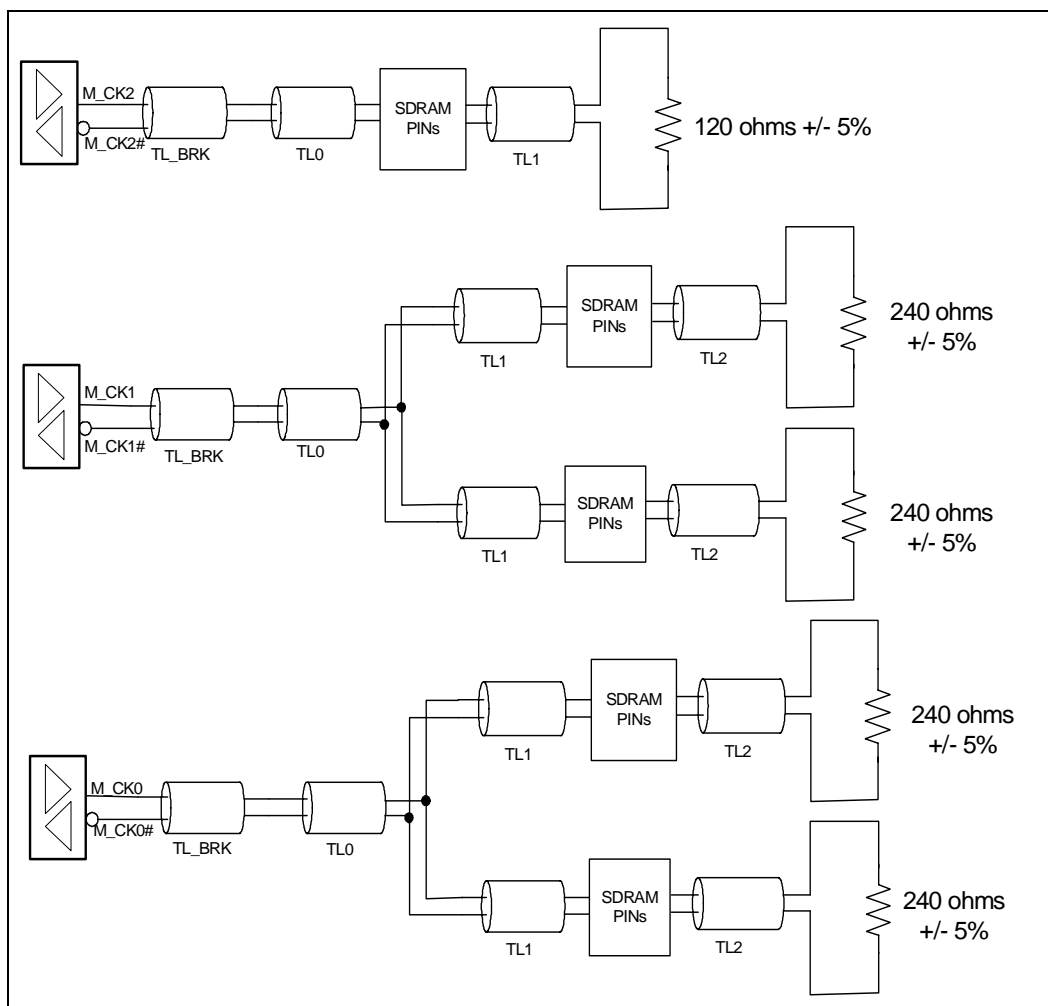
- Refer to [Table 17](#) for the embedded clock routing guidelines and [Table 18](#) for the DIMM clock segment lengths.
- Refer to [Figure 20](#) for the clock topology segment lengths.
- Refer to [Figure 17](#) for the DQ/DQS group length matching relationship with respect to the clock signals.

**Table 17. DDR2 Embedded Clock Routing Recommendations**

Parameter	Routing Guideline
Reference Plane	Route over unbroken ground plane preferred
Preferred Topology	Microstrip differential lines preferred
Breakout Trace Width and spacing	5 mils by 5 mils microstrip or stripline. Maximum length of breakout trace is 500 mils.
Trace Impedance	Differential impedance of 85 ohms +/- 15% motherboard
Trace Spacing (edge to edge)	$\geq 25$ mils between other signals.
Trace Length 1: signal Ball to memory ball	0.5" min to 10.5" max
Length Matching: Within M_CK/M_CK# (differential clock signals)	+/- 0.0250" within pairs (intra-pair)
Length Matching: With respect to DQ/DQS group (from controller to memory ball)	DQ/DQS length = clock length - 1"
Length Matching: With respect to address/command group (except CS, CKE, ODT) from controller to memory ball	ADDR/CMD $\leq$ clock length + 2" ADDR/CMD $\geq$ clock length - 1"
Length Matching: With respect to CS/CKE group	For daisy chain topology: when CS/CKE group length is $\leq 4$ "; clock length + 1" when CS/CKE group length is $> 4$ "; clock length + 3" For balanced segment topology: when CS/CKE group length is $\leq 2$ "; clock length + 1" when CS/CKE group length is $> 2$ "; clock length +/- 0.5"
Routing Guideline 1	Maximum of 2 via/layer change for M_CK/M_CK# clocks. (use the same number of vias between + and - signals of differential clock)

**Table 18. DDR2 Embedded Clock Lengths**

Traces	Description	Layer	Min Length	Max Length	Trace Impedance	Spacing (edge to edge)	Notes
TL_BRK	Breakout	Microstrip or stripline	0"	0.5"		5 mils	5 mils trace width OK for breakout.
TL0	Lead-in	Microstrip	0.5"	10"	Differential impedance of 85 ohms +/- 15% motherboard Differential impedance of 100 ohms +/- 15% add-in card	See <a href="#">Section 3.3</a> for details on differential routing. Other groups ≥ 25 mils	Length Tolerance +/- 0.1"
TL1	Lead-in SDRAM	Microstrip	0.05"	0.2"	"	"	"
TL2	Lead-in resistor	Microstrip	0.05"	0.2"	"	"	"

**Figure 20. DDR2 Embedded Clock Topology With Five SDRAMs**






#### 4.2.4.2 DDR2 Address/Command/Control Routing Guidelines

This section lists the recommendations for the DDR2 Address/Command and Control signals. (except for CS, ODT and CKE signals).

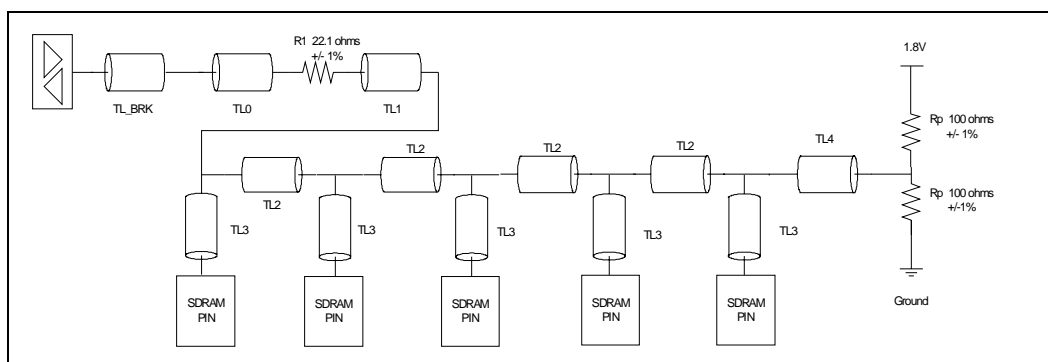
- Refer to [Table 19](#) for a description of the Address/Command signals routing guidelines.
- Refer to [Figure 14](#) for the Address/Command for split termination topology.

**Table 19. DDR2 Embedded Address/Command/Control Routing Recommendation**

Parameter	Routing Guideline
Reference Plane	Route over unbroken ground plane preferred
Preferred Topology	Microstrip lines
Breakout Trace Width and Spacing	5 mils x 5mils. Microstrip is preferred. Maximum length of the breakout trace is 500 mils.
Trace Spacing (edge to edge) <sup>1</sup>	5 mils acceptable between the pins and the breakout regions. $\geq 12$ mils within group $\geq 20$ mils from any other clock/DQ/DQS groups.
Trace Impedance	50 ohms +/- 15% for a motherboard
Trace Length: Overall length from signal Ball to SDRAM ball	1" min to 12" maximum Refer to <a href="#">Table 20</a> for segment lengths.
Length Matching: address/command group (except CS, ODT and CKE lines) with respect to clock (from controller to SDRAM ball)	ADDR/CMD $\leq$ clock length + 2" ADDR/CMD $\geq$ clock length - 1"
Split Termination	100 ohms +/- 1% to ground and 100 ohms +/- 1% to 1.8V
Routing Guideline 1	Place the VTT terminations in the VTT island after the DIMM with a trace length of 0.15" to 0.5"
Routing Guideline 2	For split terminations place the VTT termination in their respective power islands
Number of vias	6 Vias or less

**Table 20. DDR2 Embedded Address/CMD Lengths Topology**

Traces	Description	Layer	Min Length	Max Length	Trace Impedance	Spacing (edge to edge)	Notes
TL_BRK	Breakout	Microstrip	0.05"	0.5"		5 mils	5 mils trace width OK for breakout.
TL0	Lead-in Resistor	Microstrip	0.5"	9"	50 +/- 15% motherboard	≥ 12 mils within group, Other groups ≥ 20 mils	Length Tolerance +/- 0.1
TL1	Segment	Microstrip	0.2"	0.75"	"	"	"
TL2	Segment	Microstrip	0.2"	0.75"	"	"	"
TL3	Lead-in SDRAM	Microstrip	0.05"	0.2"	"	"	"
TL4	Lead-in VTT	Microstrip	0.05"	0.2"	"	"	"

**Figure 21. DDR2 Embedded Address/CMD Topology (Split Termination)**




#### 4.2.4.3 DDR2 CS, ODT and CKE Routing Guidelines

This section lists the recommendations for the layout of the DDR2 CS, ODT and CKE signals.

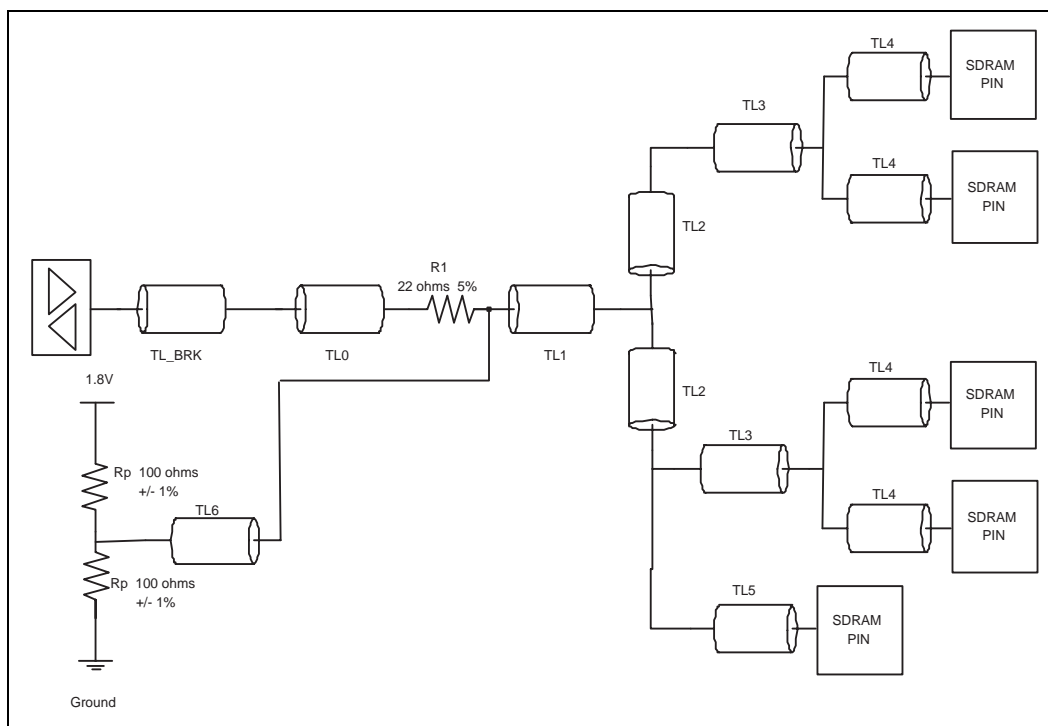
- Refer to [Table 21](#) for the segment lengths and [Figure 22](#) for the CS, ODT and CKE balanced topology. This topology requires matching each of the branches going to the SDRAM chips. This topology is the preferred topology.
- Refer to [Table 22](#) for the segment lengths and [Figure 23](#) for the CS, ODT and CKE daisy chain topology. This topology is provided for easier routing as an alternate for board designers that has difficulty with the balanced segment topology shown in [Figure 22](#).

**Table 21. DDR2 Embedded CS, ODT and CKE Routing Recommendation**

Parameter	Routing Guideline
Reference Plane	Route over unbroken ground plane preferred
Preferred Topology	Microstrip lines
Breakout Trace Width and Spacing	5 mils x 5mils. Microstrip preferred. Maximum length of the breakout trace is 500 mils.
Trace Spacing (edge to edge)	5 mils acceptable between the pins and the breakout regions. <ul style="list-style-type: none"> <li>• <math>\geq 12</math> mils within group</li> <li>• <math>\geq 20</math> mils from any other clock/DQ/DQS groups.</li> </ul>
Trace Impedance	50 ohms +/- 15% for a motherboard
Trace Lengths	Refer to <a href="#">Table 22</a> for segment lengths.
Length Matching: With respect to CS/CKE group	For daisy chain topology: when CS/CKE group length is $< 4"$ : CK length + 1" when CS/CKE group length is $> 4"$ : CK length + 3" For balanced segment topology: when CS/CKE group length is $< 2"$ : CK length + 1" when CS/CKE group length is $> 2"$ : CK length +/- 0.5"
Series Termination R1	22 +/- 5%
Split Termination Rp	100 ohms +/- 1% to ground and 100 ohms +/- 1% to 1.8V
Routing Guideline 1	For split terminations place the VTT termination in their respective power islands
Number of vias	5 Vias or less

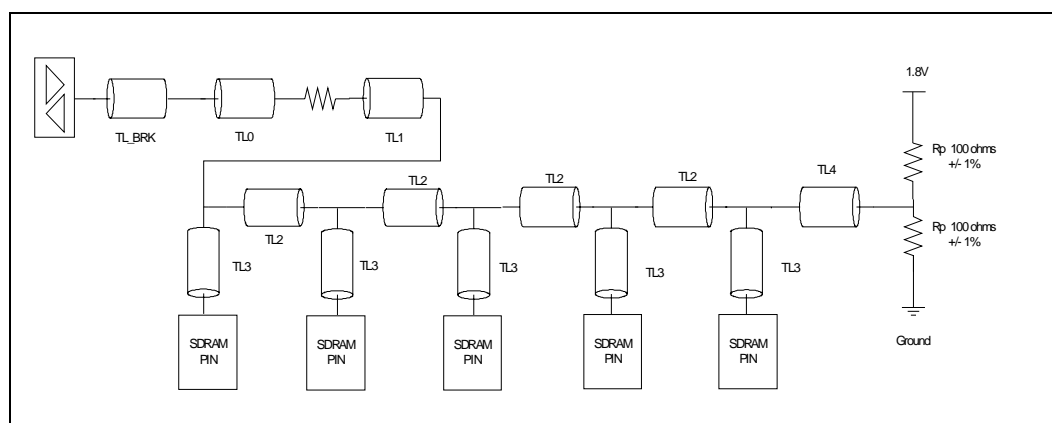
**Table 22. DDR2 Embedded CS, ODT and CKE Lengths Balanced Topology**

Traces	Description	Layer	Min Length	Max Length	Trace Impedance	Spacing (edge to edge)	Notes
TL_BRK	Breakout	Microstrip	0.05"	0.5"		5 mils	5 mils trace width OK for breakout.
TL0	Lead-in Resistor	Microstrip	0.5"	8"	50 +/- 15% motherboard	≥ 12 mils within group, Other groups ≥ 20 mils	Length Tolerance +/- 0.050
TL1	Segment	Microstrip	0.2"	0.75"	"	"	"
TL2	Segment	Microstrip	0.2"	0.2"	"	"	"
TL3	Segment	Microstrip	0.2"	0.2"	"	"	"
TL4	Lead-in SDRAM	Microstrip	0.2"	0.2"	"	"	"
TL5	Lead-in SDRAM	Microstrip	0.4	0.4	"	"	"
TL6	Lead-in Vtt	Microstrip	0.05"	0.2"	"	"	"

**Figure 22. DDR2 Embedded CS, ODT and CKE Balanced Topology**


**Equation 1. DDR2 Embedded CS, ODT and CKE Lengths Daisy Chain Topology**

Traces	Description	Layer	Min Length	Max Length	Trace Impedance	Spacing (edge to edge)	Notes
TL_BRK	Breakout	Microstrip	0.05"	0.5"		5 mils	5 mils trace width OK for breakout.
TL0	Lead-in Resistor	Microstrip	0.5"	8"	50 +/- 15% motherboard	≥ 12 mils within group, Other groups ≥ 20 mils	Length Tolerance +/- 0.05"
TL1	Segment	Microstrip	0.2"	0.75"	"	"	"
TL3	Lead-in SDRAM	Microstrip	0.05"	0.2"	"	"	"
TL4	Lead-in VTT	Microstrip	0.05"	0.2"	"	"	"

**Figure 23. DDR2 Embedded CS, ODT and CKE Daisy Chain Topology**

## 4.3 DDR2 Signal Termination

This section provides details on layout for DDR2 signal termination.

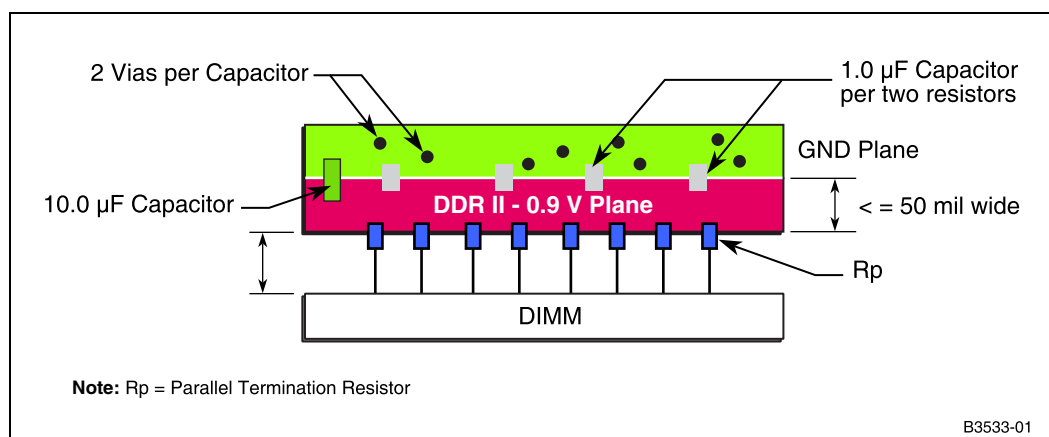
- Refer to [Section 4.3.1](#) for details on laying out the VTT for a DIMM design.
- Refer to [Section 4.3.3](#) for details on Vref.

### 4.3.1 DDR2 DIMM VTT Details

This section provides the suggested guidelines for layout for placement of the DDR2 termination resistors used for a 0.9V VTT DIMM configuration:

- Place a 0.9 V termination plane on the top layer or one of the inner layers, just beyond the DIMM connector.
- The  $V_{TT}$  island must be at least 50 mils wide.
- Use this termination plane to terminate all DIMM signals, using one termination resistor per signal.
- Decouple the  $V_{TT}$  plane using one 0.1 mF decoupling capacitor per two termination resistors.
- Each decoupling capacitor must have at least two vias between the top layer ground fill and the internal ground plane.
- In addition, place one 10  $\mu$ F or larger (100  $\mu$ F suggested) Tantalum capacitor on each end of the termination island for bulk decoupling.
- [Figure 24](#) provides an example of how to route the termination resistors.

**Figure 24. Routing Termination Resistors (Top View)**





### 4.3.2 DDR2 Termination Voltage

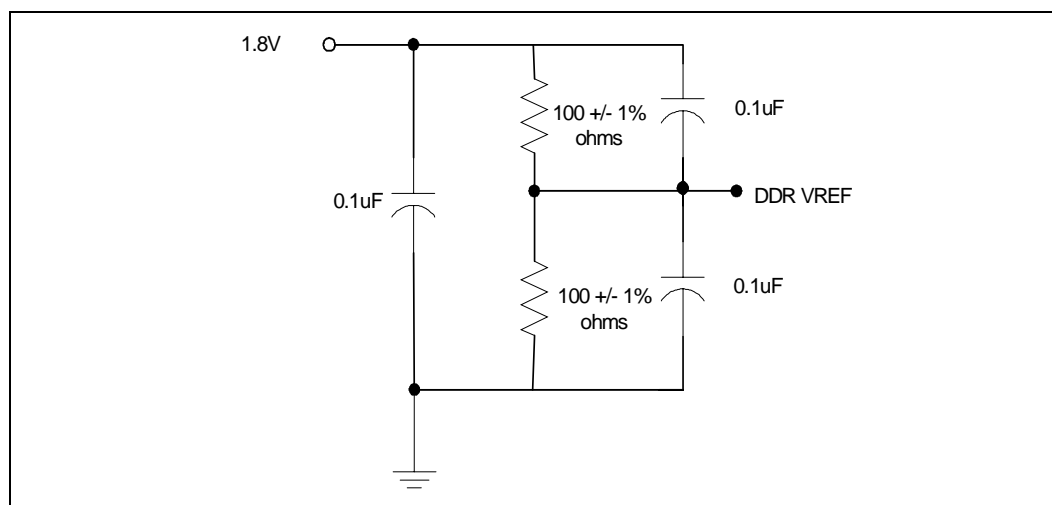
The VTT DDR termination voltage must track the VDDQ (voltage for the DDR SDRAM DQ signals) and provide the termination voltage to the termination resistors. This tracking must be 50 percent of (VDDQ - VSSQ) over voltage, temperature, and noise. It must maintain less than 40 mV offset from VREF over these conditions. This voltage must be low-impedance and source-significant current. The source and sink DC current for signal termination is at its absolute maximum current of 2.6 A-2.9 A for a 64/72-bit DIMM.

### 4.3.3 DDR V<sub>REF</sub> Voltage

The Figure 25 shows the DDR Vref voltage. The DDR VREF is a low-current source (supplying input leakage and small transients). It must track 50 percent of (VDDQ - VSSQ) over voltage, temperature, and noise. Use a single source for VREF to eliminate variation and tracking of multiple generators. Maintain 15-20 mils clearance around other nets. Use a distributed decoupling scheme. Use a simple resistor divider with 1% or better accuracy.

**Note:** The 100 ohm resistors are replaced with 1K +/- 1% resistors to minimize leakage current during battery backup mode.

**Figure 25. DDR V<sub>REF</sub> Circuit**





## 5.0 PCI Express Layout

---

This section provides an overview of the PCI Express layout recommendation based on simulation results. PCI Express is a serial differential low-voltage point-to-point interconnect. The PCI Express was designed to support 20 inches between components with standard FR4.

For more information on the PCI Express standard refer to PCI Express Base Specification 1.0a and the *PCI Express Card Electromechanical Specification*, revision 1.0a, found on the <http://www.pcisig.com/home> website.

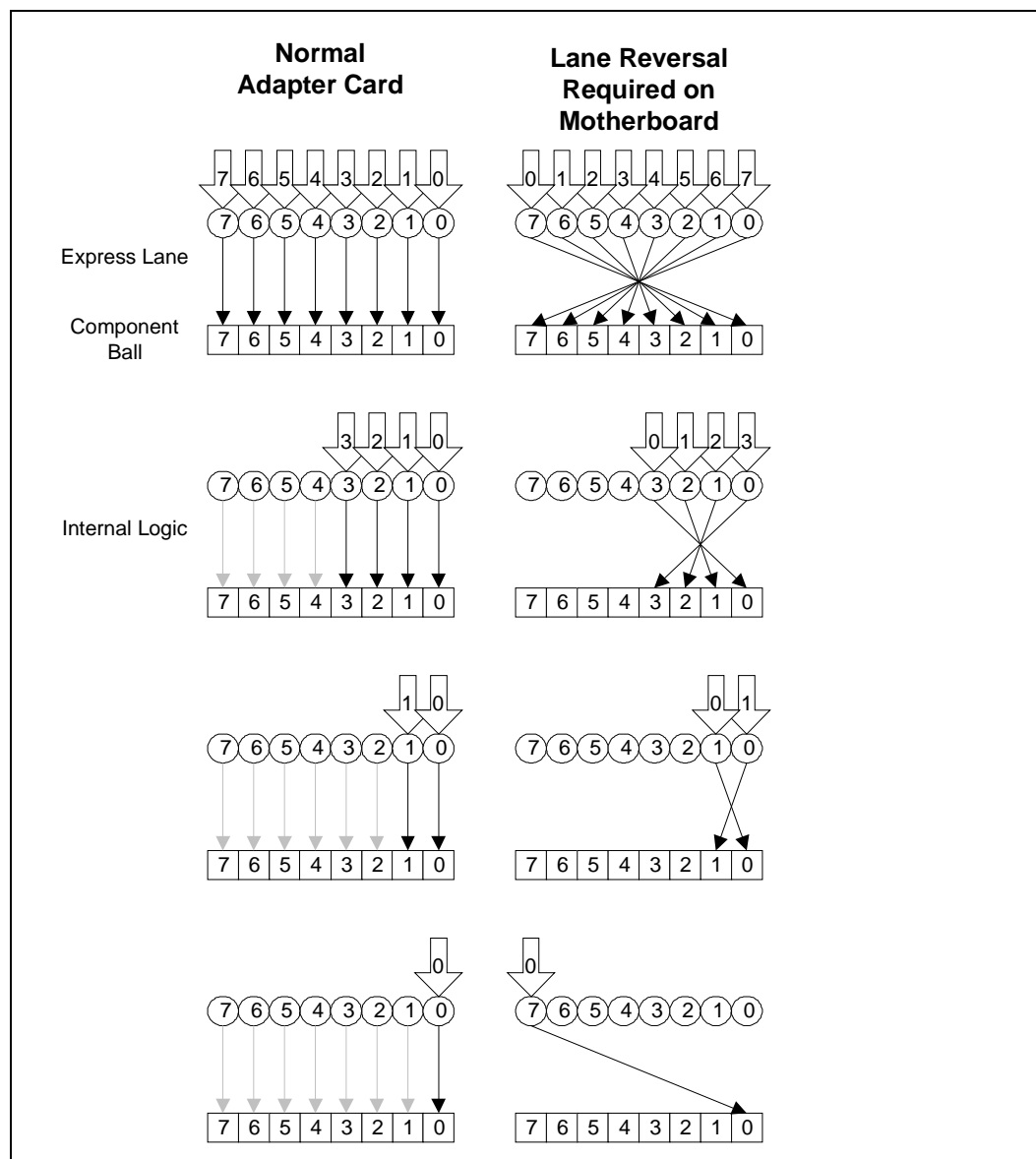




## 5.1 Optional PCI Express Lane Reversal

Figure 26 describes the lane reversal which is considered when designing a PCI-E x8 motherboard slot or an adapter card to improve PCB routing. Note that the adapter card PCI-E pins map with a straight through connection but the motherboard implements lane reversal in x8, x4, x2 and x1 configurations as shown in Figure 26.

**Figure 26. PCI Express Lane Reversal To Improve PCB Routing**



## 5.2 PCI Express Layout recommendations

The following recommendations are summarized based on presilicon simulation results for the following topologies:

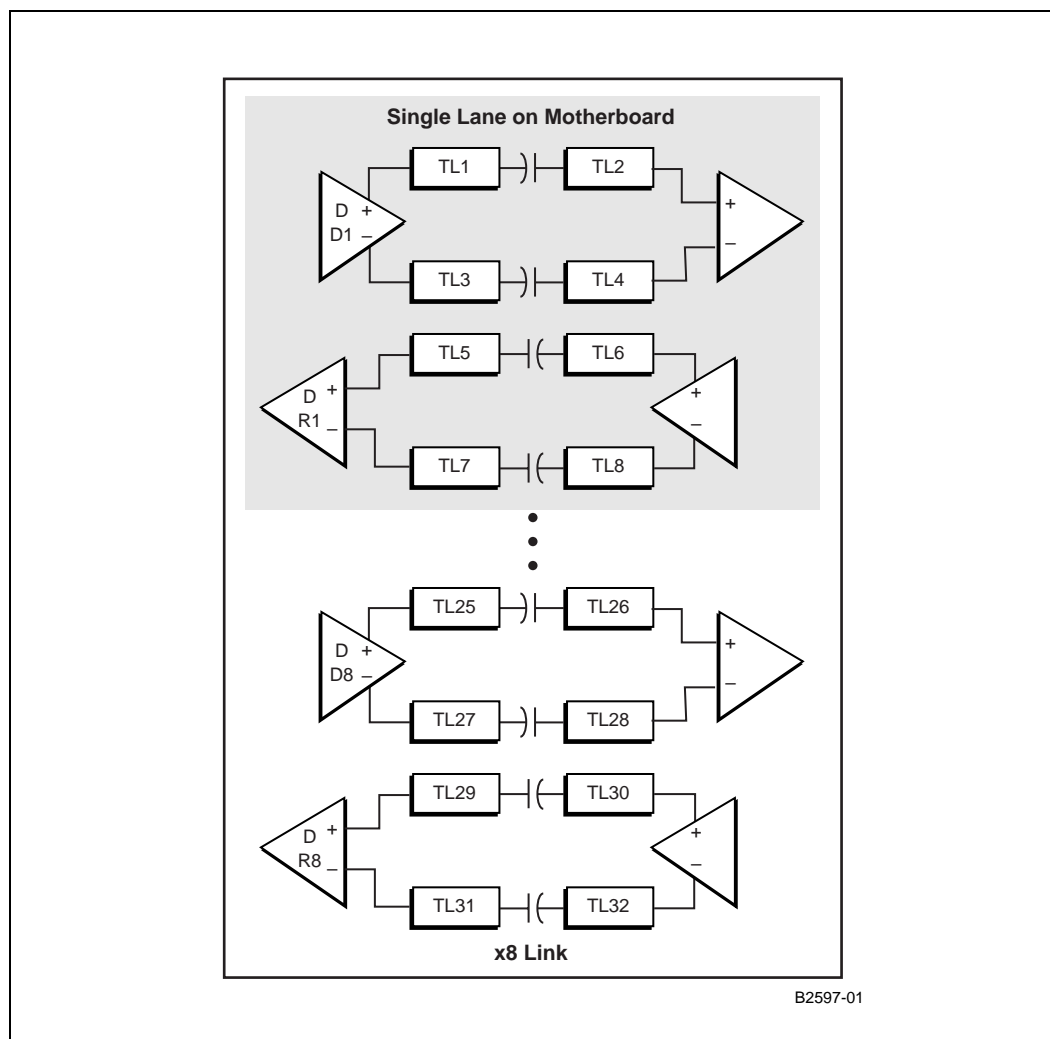
1. motherboard topology with the PCI Express device on the board [Section 5.2.1](#).
2. motherboard topology with a PCI Express connector and an adapter card topology with the device on the card [Section 5.2.2](#).

The PCI Express clock layout recommendations are listed in [Section 5.2.3](#).

### 5.2.1 PCI Express Motherboard Layout Guidelines

The following layout recommendations were determined for a motherboard application with the PCI Express device on the board.

**Figure 27. Motherboard Topology**



**Table 23. PCI Express Layout for a Motherboard**

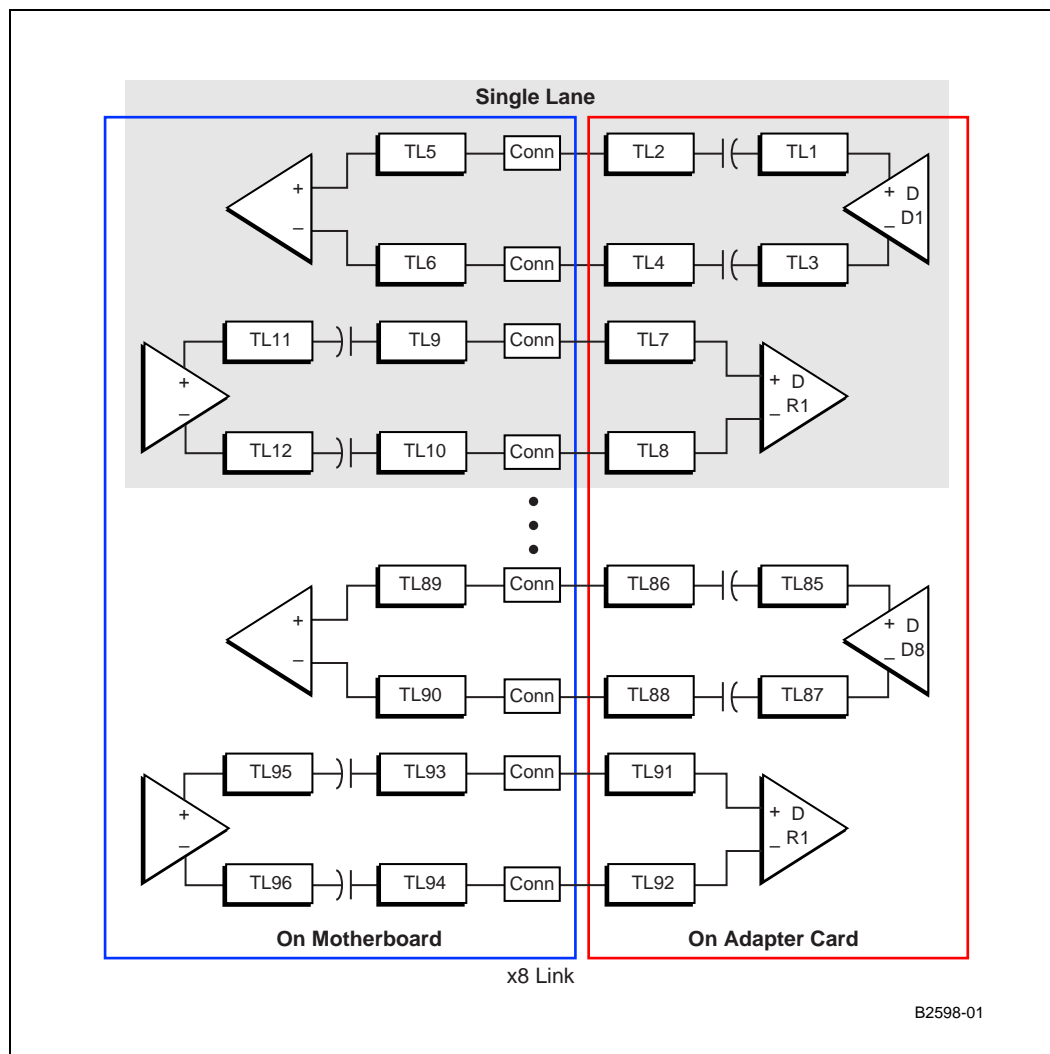
Parameter	Routing Guidelines
Signal Group	Transmit and receive differential pairs
Reference Plane	Routing over unbroken ground plane is preferred. When unbroken ground plane is not available, route over unbroken voltage plane.
Characteristic Trace Impedance:	Single-ended: 50 ohms +/- 15% Differential: 85 ohms nominal +/-15%
Microstrip Trace Width	5 mils (Refer to Table Note)
Microstrip Trace Spacing	Between + (P) and - (N) of pair: 7 mils edge to edge Between other signals: $\geq 25$ mils edge to edge Transmit and receive pairs are interleaved. When interleaving is not possible, then the spacing between pairs (inter pair) are increased to $\geq 45$ mils (edge to edge). Edge to Edge of inter pair is defined as edge of Positive of one pair to edge of Negative of the next pair or vice versa.
Stripline Trace Width	5 mils (Refer to Table Note below)
Stripline Trace Spacing	Between + (P) and - (N) of pair: 7 mils edge to edge Between other pairs: $\geq 25$ mils edge to edge Transmit and receive pairs are interleaved. When interleaving is not possible, then inter pair spacing are increased to 45 mils (edge to edge). Edge to Edge of inter pair is defined as edge of Positive of one pair to edge of Negative of the next pair or vice versa.
Group Spacing	Spacing from other groups: $> 25$ mils minimum from edge to edge for microstrip or stripline.
AC Coupling	AC Coupling capacitors must be located at the transmitter. Required values of 75 nF to 200 nF.
Total Trace Length - (Transmitter/Receiver) from device signal pin to AC coupling capacitor and AC coupling capacitor to PCI Express device pin	1.0" min. - 30.0" max
Length Matching Requirements	Total allowable between pair (length skew between + and - signals of the pair) length mismatch on a system board must not exceed 10 mils. Length are matched on a segment by segment basis. Each routing segment to be matched as close as possible. Total skew across all lanes must be less than 20 ns.
Number of Vias	4 max

**Note:** Width and Intra Pair (length skew between + and - signals of the pair) spacing recommendations need not be strictly adhered to, but it is very important to meet the given differential target impedance and specified tolerance. It is also very important to follow the inter pair spacing recommendations.

## 5.2.2 PCI Express Layout Motherboard-Adapter Card Guidelines

This section provides the routing guidelines for the motherboard-adapter card topology as shown in Figure 28. Table 24 provides the routing guidelines for a motherboard with a PCI Express connector on it and the routing guidelines for an adapter card.

**Figure 28. Motherboard-Adapter Card Topology**



**Table 24. PCI Express Layout for Motherboard-Adapter Card Topology**

Parameter	Routing Guidelines
Signal Group	Transmit and Receive differential pairs
Reference Plane	Routing over unbroken ground plane is preferred. When unbroken ground plane is not available route over unbroken voltage plane.
Characteristic Trace Impedance motherboard	Single Ended: 50 +/- 15% ohms nominal Differential: 85 +/- 15% ohms nominal
Characteristic Trace Impedance adapter card	Single Ended: 60 +/- 15% ohms nominal Differential: 100 +/- 15% ohms nominal
Microstrip Trace Width	5 mils
Microstrip Trace Spacing	Between intra-pair (between + (P) and - (N) of pair): 7 mils edge to edge (see Table Note) Between other pairs: $\geq 25$ mils edge to edge Transmit and receive pairs are interleaved. When interleaving is not possible, then the spacing between pairs (inter pair) are increased to $\geq 45$ mils (edge to edge). Edge to Edge of inter pair is defined as edge of the positive of one pair to edge of negative of the next pair or vice versa
Stripline Trace Width	5 mils (see Table Note)
Stripline Trace Spacing	Between + (P) and - (N) of pair: 7 mils edge to edge Between other pairs: $\geq 25$ mils edge to edge Transmit and Receive pairs are interleaved. When interleaving is not possible, then inter pair spacing are increased to 45 mils (edge to edge). Edge to Edge of inter pair is defined as edge of the positive of one pair to edge of negative of the next pair or vice versa
Group Spacing	Spacing from other groups: $> 20$ mils minimum from edge to edge for microstrip or stripline.
AC Coupling	AC Coupling capacitors must be located at the transmitter. Required value of 75 nF to 200 nF.
Total Length: Topology 1: from device signal pin transmitter on motherboard with PCI-E device receiver on adapter card	1.0" min. - 27" max
Total Length: Topology 2: from device signal pin transmitter on adapter card and the PCI-E device receiver on motherboard.	1.0" min. - 25" max
Length Matching Requirements	Total allowable intra-pair (length skew between + and - signals of the pair) trace mismatch for a lane that must not exceed 15 mils for the motherboard-adapter card combination (10 mils for the motherboard, 5 mils for the adapter card). Length are matched on a segment by segment basis. Total skew across all lanes must be less than 20 ns.
Number of Vias	4 max

**Note:** Width and Intra Pair Spacing (between + (P) and - (N) of pair) recommendations need not be strictly adhered to, but it is very important to meet the given differential target impedance and specified tolerance. It is also very important to follow the inter pair spacing recommendations.

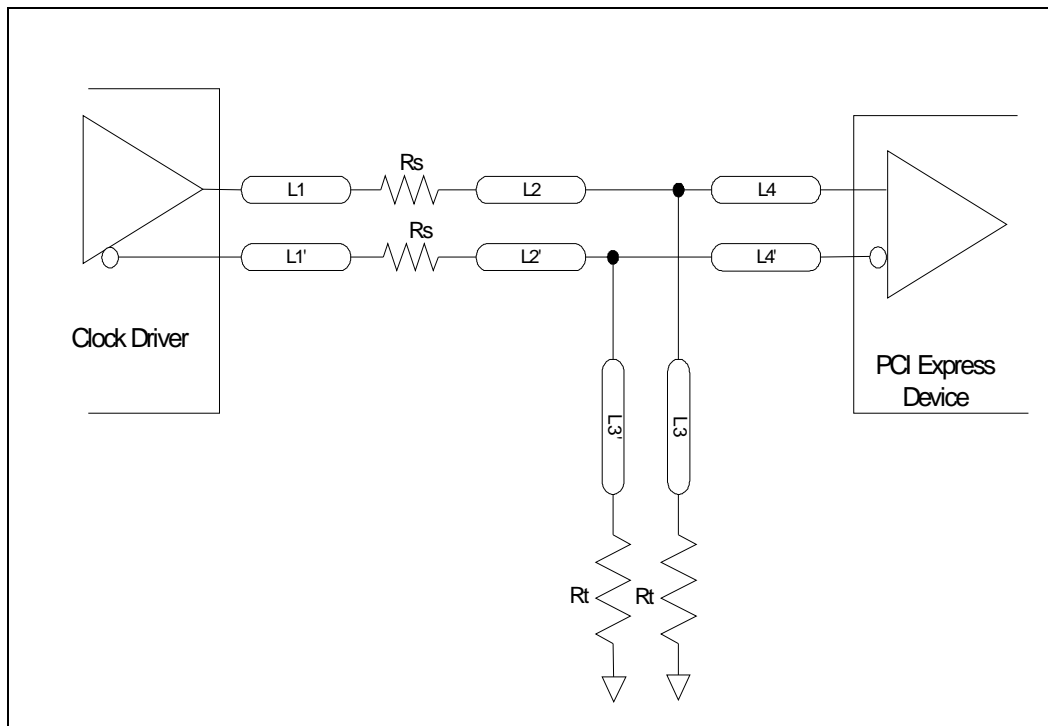
### 5.2.3 Clock Routing Guidelines

This section provides routing guidelines for the PCI Express Clocks in an application. The *PCI Express Card Electromechanical Specification Rev 1.0a* states in that any terminations required by the clock are to be on the system board.

The termination in Figure 29 is only required on the system board when these resistors were not already provided.

- PCI Express adapter cards do not have to add  $R_s$  and  $R_t$  termination resistors.

**Figure 29. PCI Express Clock Routing Topology**



**Table 25. PCI Express Layout for Clock Routing**

Parameter	Routing Guidelines
Signal Group	REFCLKP, REFCLKN differential pairs
Reference Plane	Routing over unbroken ground plane is preferred. When unbroken ground plane is not available route over unbroken voltage plane.
Characteristic Trace Impedance	Single Ended: 50 +/- 15% ohms nominal Differential: 100 +/- 15% ohms nominal
Trace Width <sup>1</sup>	5 mils (see Table Note 2)
REFCLKP, REFCLKN differential clock Pair Spacing	< 1.4 x Space Width
Serpentine Spacing (spacing of a clock lines from itself)	> 25 mils
Clock to Other Signal Spacing	> 25 mils
Trace Lengths <sup>2</sup>	L1, L1: 0.5" max
	L2, L2: 0.2" max
	L3, L3: 0.2" max
	L4, L4 Device down: 2" to 15.3" or Connector: 2" to 11.3
	Total Length = L1+L2+L4 Device Down: 3" to 16" or Connector: 3" to 12"
Length Matching Requirements within differential pair	+/- 5 mils
Rs Series Resistors	33 +/- 5%
Rt Shunt Resistors	49.9 +/- 1%
Number of Vias	4 max

**Notes:**

1. Termination resistors are only required on system boards when not already present. Adapter cards do not require Rs and Rt resistors)
2. Width and Intra Pair Spacing recommendations need not be strictly adhered to, but it is very important to meet the given differential target impedance and specified tolerance. It is also very important to follow the inter pair spacing recommendations.



## 6.0 PCI-X Layout Guidelines

---

This section provides an overview of the PCI-X layout recommendations based on Intel simulation results. The results were compiled for a motherboard with 50 ohm impedance and an adapter card with 60 ohm impedance.

- [Section 6.1](#) provides details on the central resource mode details including: PCI-X Frequency control, interrupt routing and arbitration.
- [Section 6.2](#) provides the layout recommendations for each of the topologies and PCI-X speeds.

For more information on the PCI-X standard refer to *PCI-X Addendum to the PCI Local Bus Specification*, Revision 1.0a on the [www.pcisig.com](http://www.pcisig.com) website.





## 6.1 Central Resource/Endpoint Mode Details

Intel® 81341 and Intel® 81342 I/O Processors are enabled as a central resource or an endpoint with the external strapping signal **PCIX\_EP#**. For the central resource mode **PCIX\_EP#** = 1 is set by default with an internal pull-up. For the endpoint mode **PCIX\_EP#** = 0 is set with a pull-down. The central resource dependent functions described in this section include:

- [Section 6.1.1](#) PCI-X Frequency Control
- [Section 6.1.2](#) Interrupt Routing
- [Section 6.1.3](#) Internal Arbitration
- [Section 6.1.4](#) External Arbitration

### 6.1.1 PCI/PCI-X Frequency Selection

When the central resource is enabled, the resultant mode and frequency is dependent upon the device capabilities reported as well as any system specific loading information. [Table 26](#) lists the encoding of M66EN and PCIXCAP to determine the capability speed of the PCI/PCI-X bus.

**Note:** The **PCIXCAP** signal has been defeatured. Refer the Non-Core Erratum in the *Intel® 81341 and 81342 I/O Processors Specification Update* for more information. This erratum overrides references throughout this document.

**Table 26. PCI/PCI-X Device Capability Reporting**

M66EN	PCIXCAP	PCI Device Frequency Capability	PCI-X Device Frequency Capability
Ground	Ground	33 MHz	Not capable
8.2K pull-up <sup>1</sup>	Ground	66MHz	Not Capable
Ground	10K pull-down	33 MHz	PCI-X 66MHz
8.2 K pull-up <sup>1</sup>	10K pull-down	66 MHz	PCI-X 66MHz
Ground	NC	33MHz	PCI-X 133 MHz
8.2K pull-up <sup>1</sup>	NC	66 MHz	PCI-X 133MHz

<sup>1</sup> M66EN pulled high on the motherboard.

[Table 27](#) describes the PCI-X bus mode and frequency initialization pattern that this part initiates on the PCI bus when coming out of reset as a central resource. Intel® 81341 and Intel® 81342 I/O Processors decode this initialization pattern to determine the bus frequency when it is set as an endpoint.

**Table 27. PCI-X Initialization Pattern**

DEVSEL#	STOP#	TRDY#	Mode	Clock Period (ns)		Clock Frequency (MHz)	
				Max	Min	Min	Max
Deasserted	Deasserted	Deasserted	PCI 33	60	30	16	33
			PCI 66	30	15	33	66
Deasserted	Deasserted	Asserted	PCI-X	20	15	50	66
Deasserted	Asserted	Deasserted	PCI-X	15	10	66	100
Deasserted	Asserted	Asserted	PCI-X	10	7.5	100	133

The ATU is additionally limit the frequency of the output clocks. This is useful when in an application where the PCI bus is connected to individual devices or bus slots and the PCI bus system speed needs to be limited. In this case the designer terminates the M66EN, PCIXCAP and PCIXM1\_100 # (reset strap) to set the PCI clock frequency.

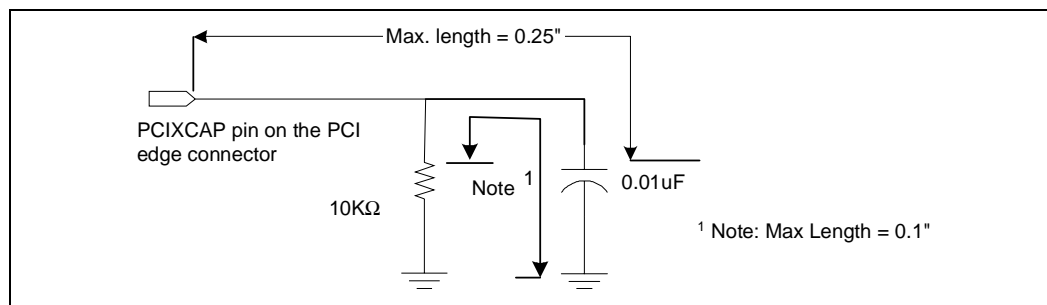
**Table 28. PCI Bus Frequency Encoding**

M66EN	PCIXCAP	PCIXM1_100#	PCI Device Frequency Capability	PCI-X Device Frequency Capability
Ground	Ground	-	33 MHz	Not capable
8.2K pull-up	Ground	-	66MHz	Not Capable
_1	10K pull-down	-		PCI-X 66MHz
-	8.2K pull-up	GND		PCI-X 100MHz
-	8.2K pull-up	NC (internal pull-up)		PCI-X 133MHz

1. '-' value is a do not care for computing the bus mode/frequency.

Figure 29 provides layout guidelines for locating the connections from the PCIXCAP pin on the card edge connector for an adapter card. With Intel® 81341 and Intel® 81342 I/O Processors on an adapter card the P\_PCIXCAP pin are pulled-up with an 8.2K resistor.

**Table 29. P\_PCIXCAP Layout Guidelines with a Intel® 81341 and Intel® 81342 I/O Processors Adapter card**

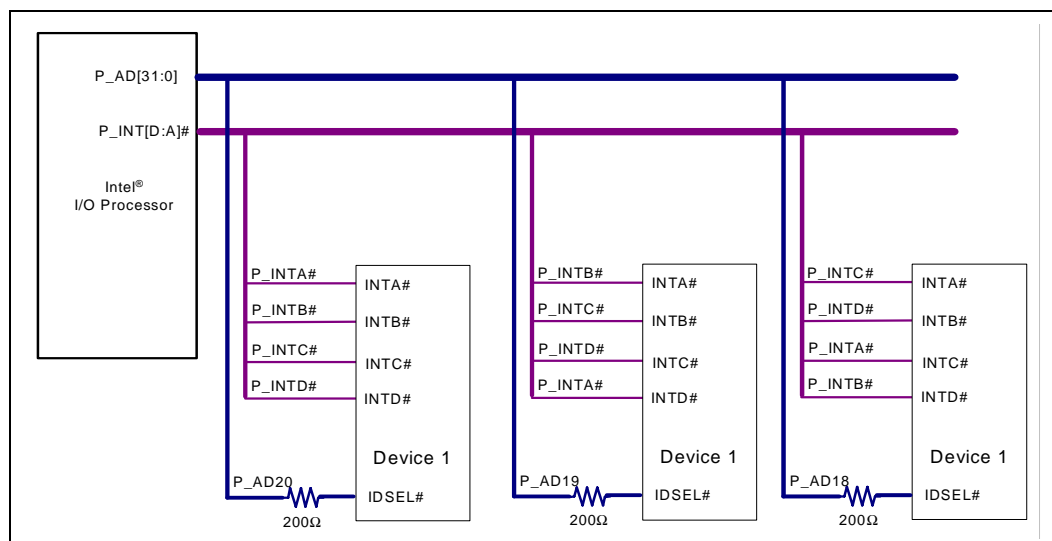




### 6.1.2 Interrupt Routing in Central Resource Mode

Figure 30 shows the device running in central resource mode connected to three multifunction PCI devices. Notice that the interrupts are rotated for each device. The practice of rotating interrupts are also used when connecting to PCI slots. The IDSEL lines acts as chip selects during the configuration cycles. The IDSEL lines are mapped to upper address lines which are unused during the configuration cycles. Each IDSEL line requires a 200 ohm series resistor on it as shown in Figure 30.

**Figure 30. Interrupt and IDSEL Mapping**



### 6.1.3 Internal Arbitration

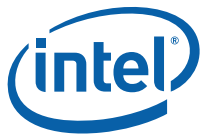
Intel® 81341 and Intel® 81342 I/O Processors have an internal PCI arbiter that supports up to four external masters. A hardware bootstrap method has been provided to enable or disable the internal arbiter at boot-up time. The internal arbiter is enabled when  $EXT\_ARB\# = '1'$  at the rising edge of  $P\_RST\#$  signal. The request inputs into the internal arbiter include: 4 external request inputs  $P\_REQ[3:0]\#$ , and the Intel® 81341 and Intel® 81342 I/O Processors Address Translation Unit.

### 6.1.4 External Arbitration

When the reset strap  $EXT\_ARB\# = '0'$ , then the internal arbiter in Intel® 81341 and Intel® 81342 I/O Processors are disabled and an external arbiter is used instead for PCI bus arbitration.

When operating in the external arbiter mode, Intel® 81341 and Intel® 81342 I/O Processors produces one  $P\_REQ\#$  output and receives one  $P\_GNT\#$  input. The  $P\_REQ\#$ / $P\_GNT\#$  pair is for ATU transactions.

The Intel® 81341 and Intel® 81342 I/O Processors arbitration pins switch modes between internal and external arbitration.  $P\_GNT[0]\#$  pin becomes the ATU request output  $P\_REQ\#$  to the external arbiter and  $P\_REQ[0]\#$  pin becomes the ATU grant input  $P\_GNT\#$  from the external arbiter.



## 6.2 PCI-X Layout Recommendations

This section provides the layout recommendations for PCI-X topologies in the following subsections:

- [Section 6.2.1, "PCI-X Clock Routing Guidelines"](#)
- [Section 6.2.2, "Point-to-Point Signals \(REQ#/GNT#\)"](#)
- [Section 6.2.3, "133 MHz One Slot Topology"](#)
- [Section 6.2.4, "Embedded 133 MHz Topology"](#)
- [Section 6.2.5, "Mixed 133 MHz Topology"](#)
- [Section 6.2.6, "100 MHz Two Slot Topology"](#)
- [Section 6.2.7, "Embedded 100 MHz Topology"](#)
- [Section 6.2.8, "Mixed 100 MHz Topology"](#)
- [Section 6.2.9, "66 MHz PCI-X Four Slot Topology"](#)
- [Section 6.2.10, "Embedded 66 MHz Topology"](#)
- [Section 6.2.11, "Mixed 66 MHz Topology"](#)
- [Section 6.2.12, "Additional PCI Layout Notes"](#)



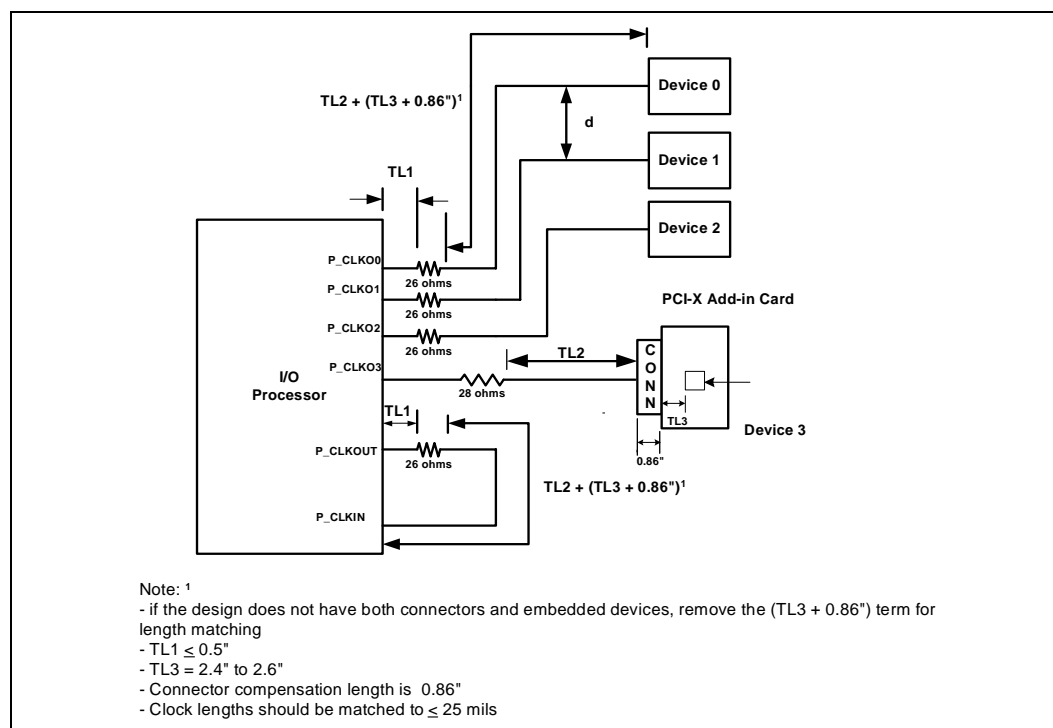
## 6.2.1 PCI-X Clock Routing Guidelines

The Intel® 81341 and Intel® 81342 I/O Processors provides a clock buffer for up to four PCI-X devices when operating in central resource mode. Note that when the **P\_CLKIN** is the primary clock source (**CLK\_SRC\_PCIE#** = 1), the PCI Clock outputs are unavailable and not used as a clock source for any device.

The *PCI-X Addendum to the PCI Local Bus Specification*, Revision 1.0b, allows a maximum of 0.5 ns clock skew timing for each of the PCI-X frequencies: 66 MHz, 100 MHz, and 133 MHz.

Figure 31 shows four clocks connected to individual PCI-X devices with P\_CLKOUT fed back into P\_CLKIN.

**Figure 31. PCI Clock Distribution and Matching Requirements**



**Table 30. PCI-X Clock Layout Guidelines**

Parameter	Routing Guidelines
Reference Plane	Route over unbroken ground plane.
Recommended Layer	Stripline
Trace Impedance: Motherboard	Microstrip: 50 ohm +/- 15%, stripline: 50 ohm +/- 10%
Trace Impedance: Adapter Card	Microstrip or stripline: 60 ohm +/- 15%
Trace Spacing (edge to edge)	between two different clock lines $\geq 25$ mils between two segments of the same clock line $\geq 25$ mils between clock and other signals $\geq 50$ mils
Series Resistors	28 ohms 1% for connectors 26 ohms 1% for embedded
Trace Length TL1 from buffer to the resistor	1.0" max
Total Trace Length: from device ball to device (including resistor segment)	11" max
Length Matching:	All clock lines including PCLKOUT to PCLKIN (feedback clock) must be matched to within 25 mils. Refer to <a href="#">Figure 31</a> .
Topologies with only embedded devices	Match clocks to within 25 mils
Topologies with only connectors	Match clocks to within 25 mils. Rout feedback clock longer to compensate for the adapter card length (2.4" to 2.6") + 0.85" (for the connector delay)
Topologies with both slots and devices used in the design	Match Clocks to within 25 mils Rout feedback clock longer to compensate for the adapter card length (2.4" to 2.6") + 0.85" (for the connector delay) PCLK's going to the embedded devices must be compensate for the adapter card length (2.4" to 2.6") + 0.85" (for the connector delay)
Vias	$\leq 2$ vias



## 6.2.2 Point-to-Point Signals (REQ#/GNT#)

This section provides the layout guidelines for REQ# and GNT# lines. Topology in Figure 32 for 133 MHz slot design is the same as the one used for point-to-point signals.

**Table 31. PCI-X REQ#/GNT# Layout Guidelines**

Parameter	Routing Guidelines
Signal Group	REQ# and GNT# lines
Reference Plane	Route over unbroken reference plane.
Motherboard Impedance (microstrip)	50 ohm $\pm$ 15% microstrip and 50 ohm $\pm$ 10% stripline
Motherboard Trace Spacing	14 mils microstrip and 12 mils stripline
Add-in Card Impedance	60 ohm $\pm$ 15% microstrip and stripline
Add-in Card Trace Spacing	18 mils microstrip and 14 mils stripline
Group Spacing: Spacing from other groups	25 mils minimum, edge to edge
Trace Length TL1 - from buffer to the connector	0.5" min to 4.5" max for 133MHz 0.5" min to 12.0" max for 100MHz 0.5" min to 15.0" max for 66MHz
Trace Length TL2 - from connector to the receiver	2.4" - 2.6" max
Vias	$\leq$ 3 vias

### 6.2.2.1 Point-to-Point Signals (REQ#/GNT#)

This section provides the layout guidelines for REQ# and GNT# lines. Topology in Figure 32 for 133MHz slot design is the same as the one used for point-to-point signals.

**Table 32. PCI-X REQ#/GNT# Layout Guidelines**

Parameter	Routing Guidelines
Signal Group	REQ# and GNT# lines
Reference Plane	Route over unbroken reference plane.
Motherboard Impedance (microstrip)	50 ohm $\pm$ 15% microstrip and 50 ohm $\pm$ 10% stripline
Motherboard Trace Spacing	14 mils microstrip and 12 mils stripline
Add-in Card Impedance	60 ohm $\pm$ 15% microstrip and stripline
Add-in Card Trace Spacing	18 mils microstrip and 14 mils stripline
Group Spacing: Spacing from other groups	25 mils minimum, edge to edge
Trace Length TL1 - from buffer to the connector	0.5" min to 4.5" max for 133MHz 0.5" min to 12.0" max for 100MHz 0.5" min to 15.0" max for 66MHz
Trace Length TL2 - from connector to the receiver	2.4" - 2.6" max
Vias	$\leq$ 3 vias

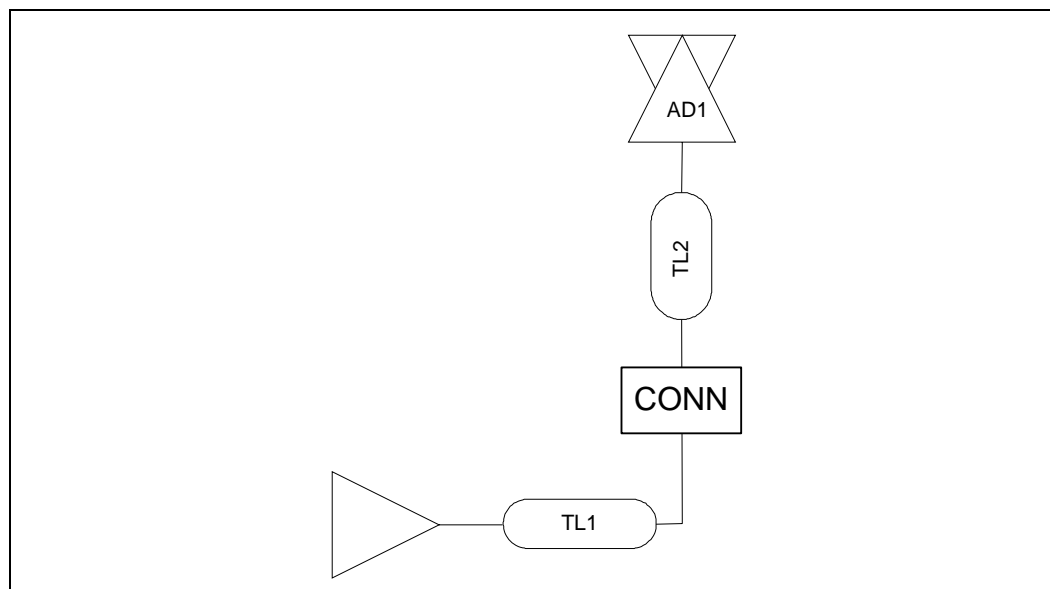
### 6.2.3 133 MHz One Slot Topology

This section lists the parameters used for the address/data and control lines for 133 MHz single slot design.

**Table 33. 133 MHz Single-Slot Topology**

Parameter	Routing Guidelines	
	Lower AD	Upper AD
Signal Group	Address, data and control lines	
Reference Plane	Route over unbroken reference plane.	
Motherboard Impedance (microstrip)	50 ohm +/- 15% microstrip and 50 ohm +/- 10% stripline	
Motherboard Trace Spacing	14 mils microstrip 12 mils stripline	
Add-in Card Impedance	60 ohm +/- 15% microstrip and stripline	
Add-in Card Trace Spacing	14 mils microstrip and stripline	
Group Spacing	Spacing from other groups: 25 mils minimum, edge to edge	
Trace Length TL1 - from SL ball to the connector	1.0" - 6.0" max	0.5" - 5.0" max
Trace Length TL2 - from connector to the receiver	0.75" - 1.5" Max	1.75" - 2.75" Max
Vias	< 3 vias	

**Figure 32. 133 MHz One Slot Topology**



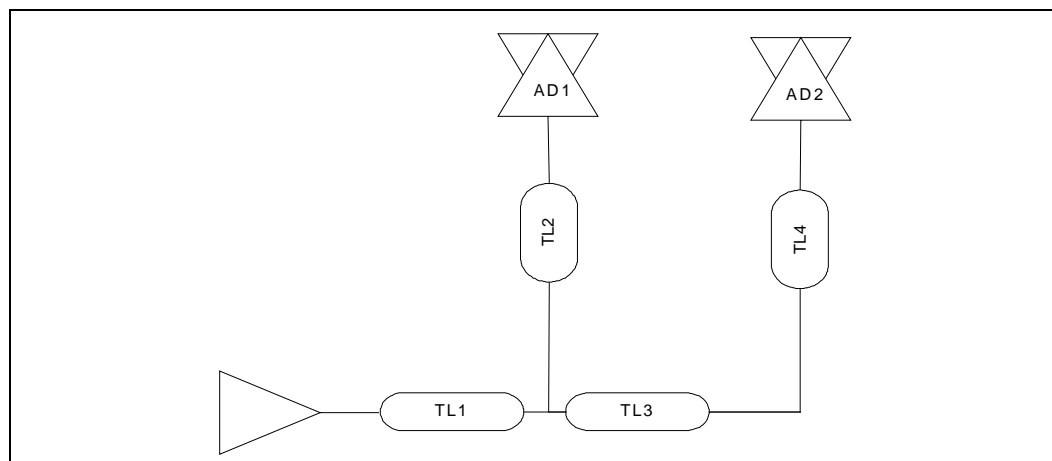




## 6.2.4 Embedded 133 MHz Topology

This section lists the parameters used for the address, data and control signals for 133 MHz embedded design with two embedded devices.

**Figure 33. Embedded 133 MHz Topology**



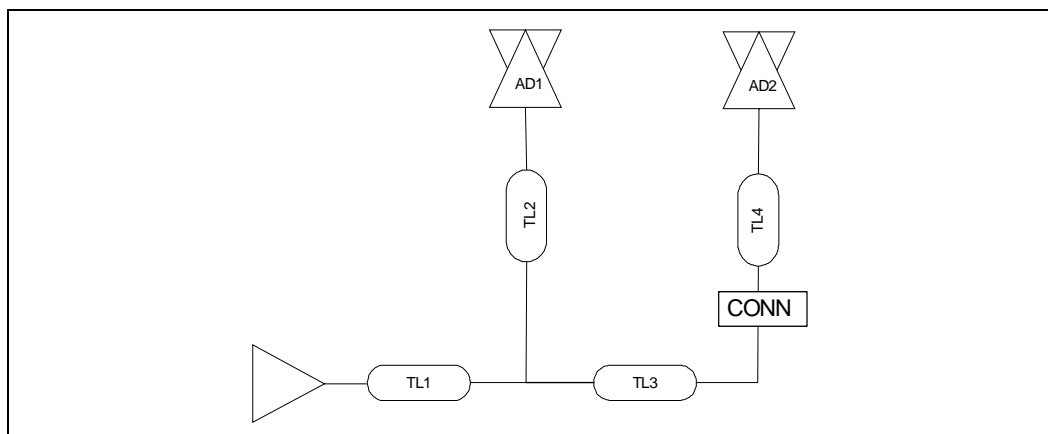
**Table 34. Embedded 133 MHz Topology**

Parameter	Routing Guidelines	
	Lower AD	Upper AD
Signal Group	Address, Data and control line	
Reference Plane	Route over unbroken reference plane.	
Motherboard Impedance (microstrip)	50 ohm +/- 15%	
Motherboard Impedance (Stripline)	50 ohm +/- 10%	
Motherboard Trace Spacing	14 mils microstrip, 12 mils stripline	
Group Spacing	Spacing from other groups: 25 mils minimum, edge to edge	
Trace Length TL1 - from ball to the junction	0.75" min. to 2.5" max	
Trace Length TL3 - from junction to junction	0.75" min. to 2.5" max	
Trace Length TL2, TL4, from junction to receiver	0.75" min. to 2.5" max	
Vias	< 3 vias	

## 6.2.5 Mixed 133 MHz Topology

This section lists the parameters used for the address, data and control signals for 133 MHz embedded design with one embedded load and one connector.

**Figure 34. Mixed 133 MHz Topology**



**Table 35. Mixed 133 MHz Topology**

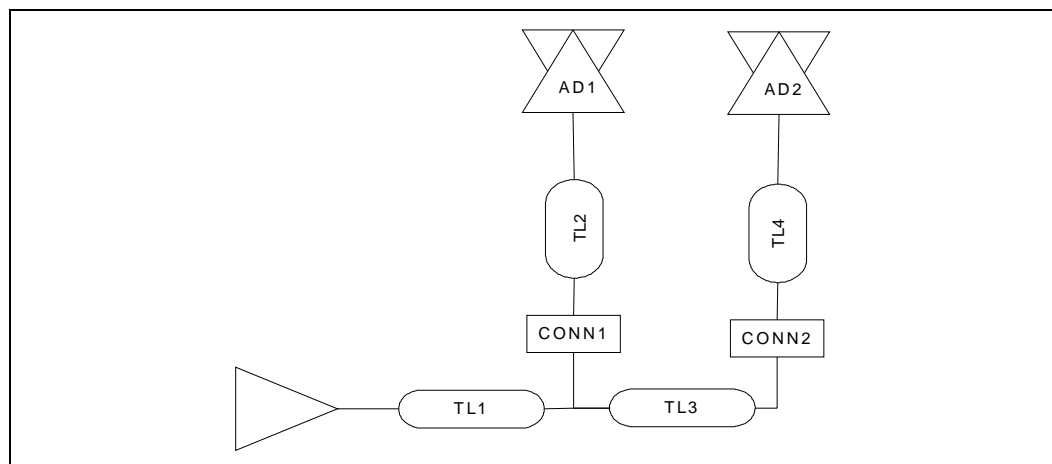
Parameter	Routing Guidelines	
	Lower AD	Upper AD
Signal Group	Data and control line	
Reference Plane	Route over unbroken reference plane.	
Motherboard Impedance (microstrip)	50 ohm +/- 15% microstrip and 50 ohm +/- 10% stripline	
Motherboard Trace Spacing	18 mils microstrip 14 mils stripline	
Add-in Card Impedance	60 ohm +/- 15% microstrip and stripline	
Add-in Card Trace Spacing	18 mils microstrip and 14 mils stripline	
Group Spacing	Spacing from other groups: 25 mils minimum, edge to edge	
Trace Length TL1 - from SL ball to the junction	0.5" min. to 2.0" max	0.5" min. to 2.0" max
Trace Length TL2 - from junction to AD1	0.5" min. to 2.0" max	0.5" min. to 2.0" max
Trace Length TL3, from junction to CONN	0.5" min. to 3.5" max	0.5" min. to 2.25" max
Trace Length TL4, from CONN to adapter	0.75" min. to 1.5" max	1.75" min. to 2.75" max
Vias	< 3 vias	



## 6.2.6 100 MHz Two Slot Topology

This section lists the parameters used for the address, data and control signals for 100 MHz. This topology is shown in [Figure 35](#).

**Figure 35. 100 MHz Dual Slot Topology**



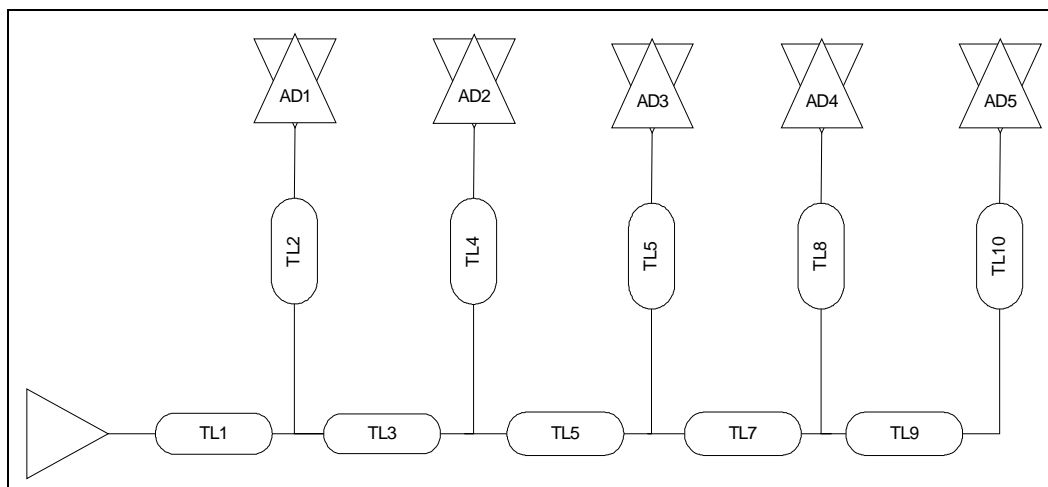
**Table 36. 100 MHz Two Slot Topology**

Parameter	Routing Guidelines	
	Lower AD	Upper AD
Signal Group	Data and control line	
Reference Plane	Route over unbroken reference plane.	
Motherboard Impedance (microstrip)	50 ohm +/- 15% microstrip and 50 ohm +/- 10% stripline	
Motherboard Trace Spacing	18 mils microstrip 14 mils stripline	
Add-in Card Impedance	60 ohm +/- 15% microstrip and stripline	
Add-in Card Trace Spacing	18 mils microstrip and 14 mils stripline	
Group Spacing	Spacing from other groups: 25 mils minimum, edge to edge	
Trace Length TL1 - from ball to the connector	0.5" - 12.0" max	0.5" - 10.0" max
Trace Lengths TL3 - Between connectors	0.5" - 3.0" max	0.5" - 3.0" max
Trace Lengths TL2 - from connector to the first receiver, TL4 - from connector to the second receiver	0.75" - 1.50" max	1.75" - 2.75" max
Vias	< 3 vias	

## 6.2.7 Embedded 100 MHz Topology

This section lists the parameters used for the address, data and control signals for 100 MHz embedded design with five embedded loads.

**Figure 36. Embedded 100 MHz Topology**



**Table 37. Embedded 100 MHz Topology**

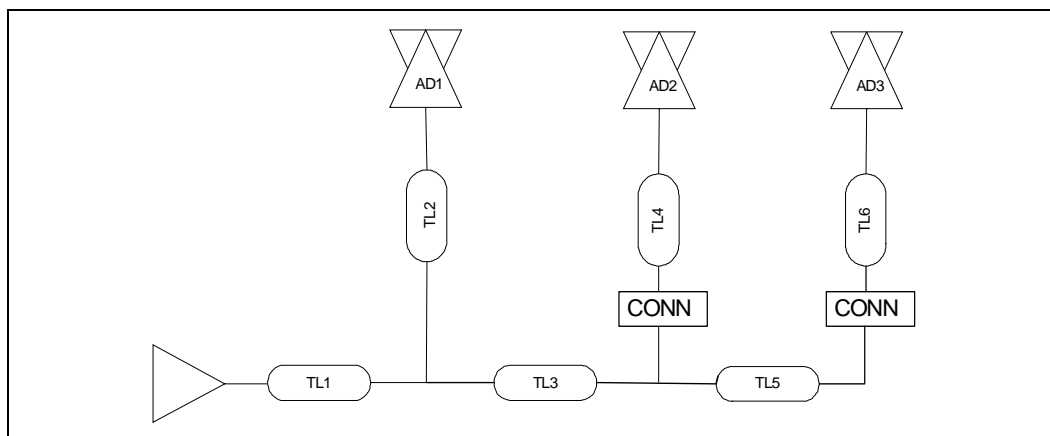
Parameter	Routing Guidelines	
	Lower AD	Upper AD
Signal Group	Address, data and control line	
Reference Plane	Route over unbroken reference plane.	
Motherboard Impedance (microstrip)	50 ohm +/- 15% microstrip and 50 ohm +/- 10% stripline	
Motherboard Trace Spacing	18 mils microstrip 14 mils stripline	
Group Spacing	spacing from other groups: 25 mils minimum, edge to edge	
Trace Length TL1 - from SL ball to the junction	0.5" min. to 3.0" max (3 loads, 5 loads)	
Trace Length TL3, TL5, TL7, TL9: from junction to junction	0.5" min. to 2.0" max (3 loads) 0.5" min. to 1.0" max (5 loads)	
Trace Length TL2, TL4, TL6, TL8, TL10: from junction to receiver	0.5" min. to 3.0" max (3 loads) 0.5" min to 2.0" max (5 loads)	
Vias	≤ 4 vias	



## 6.2.8 Mixed 100 MHz Topology

This section lists the parameters used for the address, data and control signals for 100 MHz embedded design with one embedded load and two connectors.

**Figure 37. Mixed 100 MHz Topology**



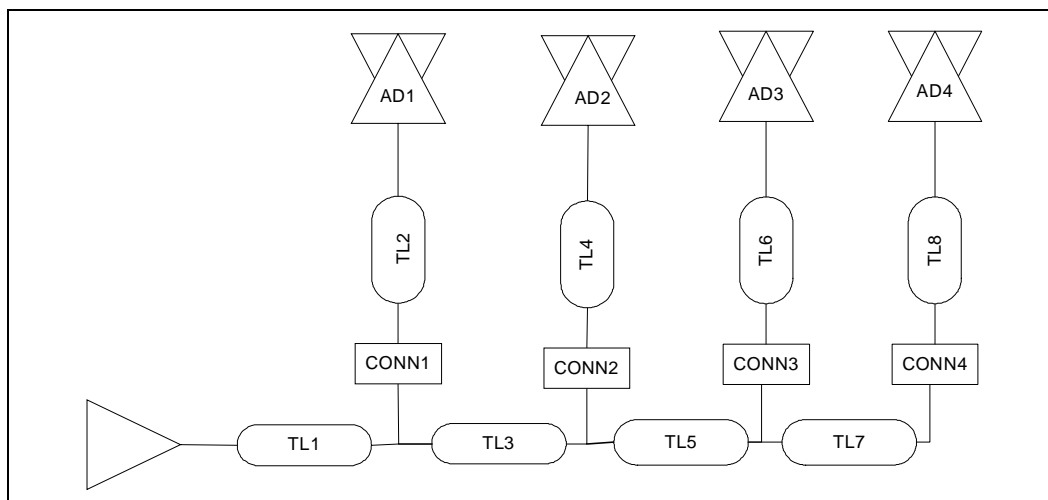
**Table 38. Mixed 100 MHz Topology**

Parameter	Routing Guidelines	
	Lower AD	Upper AD
Signal Group	Address, data and control line	
Reference Plane	Route over unbroken reference plane.	
Motherboard Impedance (microstrip)	50 ohm +/- 15% microstrip and 50 ohm +/- 10% stripline	
Motherboard Trace Spacing	18 mils microstrip 14 mils stripline	
Add-in Card Impedance	60 ohm +/- 15% microstrip and stripline	
Add-in Card Trace Spacing	18 mils microstrip and 14 stripline	
Group Spacing	Spacing from other groups: 25 mils minimum, edge to edge	
Trace Length TL1 - from SL ball to the junction	0.5" min. to 2.5" max	0.5" min. to 2.5" max
Trace Length TL2 - from junction to AD1	0.5" min. to 2.0" max	0.5" min. to 2.0" max
Trace Length TL3, from junction to first CONN	0.5" min. to 3.5" max	0.5" min. to 3.0" max
Trace Length TL5, from 1st CONN to 2nd CONN	0.5" min. to 3.5" max	0.5" min. to 3.5" max
Trace Length TL4, from 1st CONN to AD2 Trace Length TL6, from 2nd CONN to AD3	0.75" min. to 1.5" max	1.75" min. to 2.75" max
Vias	≤ 3 vias	

## 6.2.9 66 MHz PCI-X Four Slot Topology

This section lists the parameters used for the address, data and control signals for 66 MHz. This topology is shown in Figure 38.

**Figure 38. 66 MHz Four Slot Topology**



**Table 39. 66 MHz Four Slot Topology**

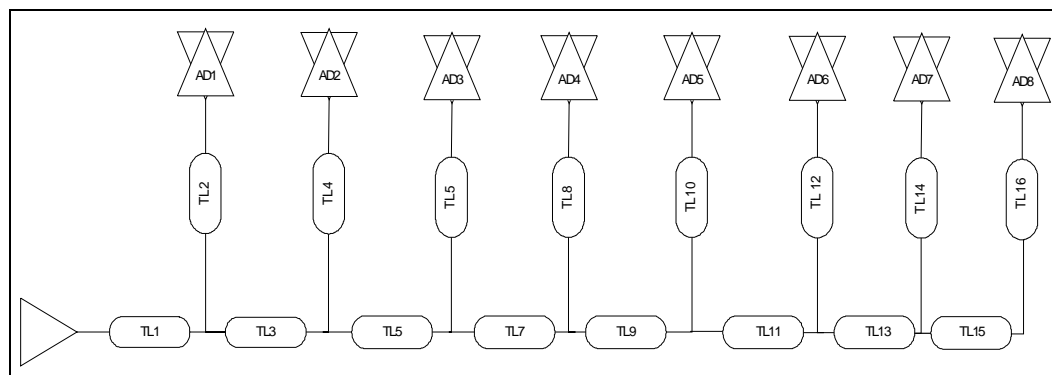
Parameter	Routing Guidelines	
	Lower AD	Upper AD
Signal Group	Data and control line	
Reference Plane	Route over unbroken reference plane.	
Motherboard Impedance (microstrip)	50 ohm +/- 15% microstrip and 50 ohm +/- 10% stripline	
Motherboard Trace Spacing	18 mils microstrip 14 mils stripline	
Add-in Card Impedance	60 ohm +/- 15% microstrip and stripline	
Add-in Card Trace Spacing	12 mils microstrip and 12 mils stripline	
Group Spacing	Spacing from other groups: 25 mils minimum, edge to edge	
Trace Length TL1 - from ball to the connector	0.5" - 12.0" max	0.5" - 9.0" max
Trace Lengths TL3, TL5, TL7 - Between connectors	0.5" - 2.0" max	0.5" - 2.0" max
Trace Lengths TL2, TL4, TL6, TL8- from connector to the first receivers	0.75" - 1.50" max	1.75" - 2.75" max
Vias	≤ 4 vias	



## 6.2.10 Embedded 66 MHz Topology

This section lists the parameters used for the address, data and control signals for 66 MHz embedded design with 8 embedded loads.

**Figure 39. Embedded 66 MHz Topology**



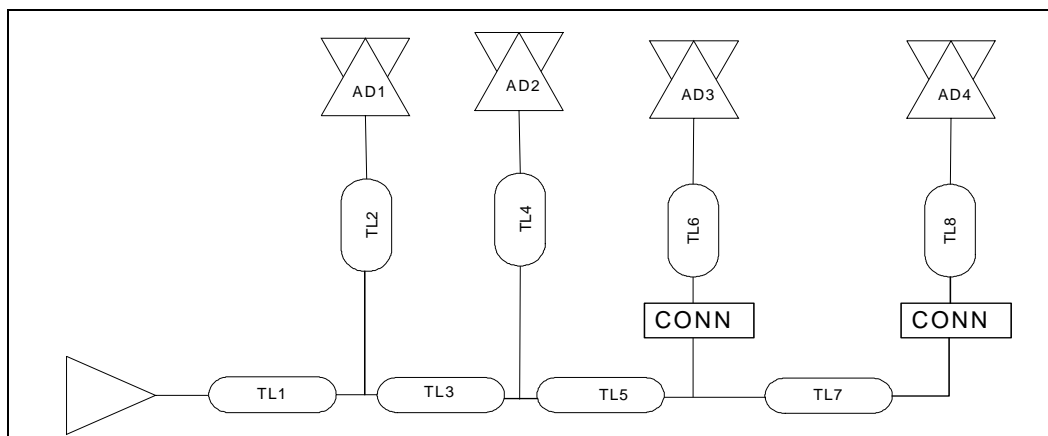
**Table 40. Embedded 66 MHz Topology**

Parameter	Routing Guidelines	
	Lower AD	Upper AD
Signal Group	Address, data and control line	
Reference Plane	Route over unbroken reference plane.	
Motherboard Impedance (microstrip)	50 ohm +/- 15% microstrip and 50 ohm +/- 10% stripline	
Motherboard Trace Spacing	18 mils microstrip 14 mils stripline	
Group Spacing	Spacing from other groups: 25 mils minimum, edge to edge	
Trace Length TL1 - from SL ball to the junction	0.5"min. to 3.0" max (8 loads) 0.5"min. to 3.5" max (6 loads)	
Trace Length TL3, TL5, TL7, TL9, TL11, TL13, TL15: from junction to junction	0.5"min. to 1.5" max (8 loads) 0.5"min. to 2.5" max (6 loads)	
Trace Length TL2, TL4, TL6, TL8, TL10, TL12, TL14, TL16: from junction to receiver	0.5"min. to 1.5" max (8 loads) 0.5" min to 2.0" max (6 loads)	
Vias	≤ 4 vias	

### 6.2.11 Mixed 66 MHz Topology

This section lists the parameters used for the address, data and control signals for 66 MHz embedded design with one embedded load and two connectors.

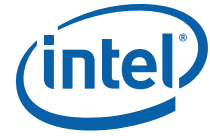
**Figure 40. Mixed 66 MHz Topology**



**Table 41. Mixed 66 MHz Topology**

Parameter	Routing Guidelines	
	Lower AD	Upper AD
Signal Group	Address, data and control line	
Reference Plane	Route over unbroken reference plane.	
Motherboard Impedance (microstrip)	50 ohm +/- 15% microstrip and 50 ohm +/- 10% stripline	
Motherboard Trace Spacing	18 mils microstrip 14 mils stripline	
Add-in Card Impedance	60 ohm +/- 15% microstrip and stripline	
Add-in Card Trace Spacing	12 mils microstrip and 12 mils stripline	
Group Spacing	Spacing from other groups: 25 mils minimum, edge to edge	
Trace Length TL1 - from SL ball to the junction	0.5" min. to 11" max	0.5" min. to 10" max
Trace Length TL2, TL4 - from junction to AD1, AD2	0.5" min. to 4.5" max	0.5" min. to 4.0" max
Trace Length TL3, TL5, TL7 from junction to junction	0.5" min. to 4.0" max	0.5" min. to 4.0" max
Trace Length TL6 from 1st CONN to AD3, TL8: from 2nd CONN to AD4	0.75" min. to 1.5" max	1.75" min. to 2.75" max
Vias	≤ 4 vias	





## 6.2.12 Additional PCI Layout Notes

- The P\_INT[D:A]# signals do not have any length restrictions.
- When PCIX\_PULLUP# is pulled-low, it enables internal pull-ups on the following PCI signals: P\_AD[63:32], P\_C/BE[7:4]#, P\_PAR64, P\_REQ64#, P\_ACK64#, P\_FRAME#, P\_IRDY#, P\_TRDY#, P\_STOP#, P\_DEVSEL#, P\_SERR#, P\_INT[D:A]#, and P\_PERR#.
- When application requires external pull-ups on the upper P\_AD bus make sure that the location of the pull-up is less than  $\leq 1$ " from the ball.



## 7.0 Peripheral Local Bus

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This section provides the layout guidelines for the Peripheral Bus Interface Unit (PBI) of Intel® 81341 and Intel® 81342 I/O Processors. The PBI bus is commonly used to interface flash components to the Intel® 81341 and Intel® 81342 I/O Processors Peripheral Bus.

The PBI unit includes two chip enables. The PBI chip enables activate the appropriate peripheral device when the address falls within one of the PBIs two programmable address ranges. Each chip enable supports up to 32 MBytes of addressability.

### 7.1 Peripheral Bus Signals

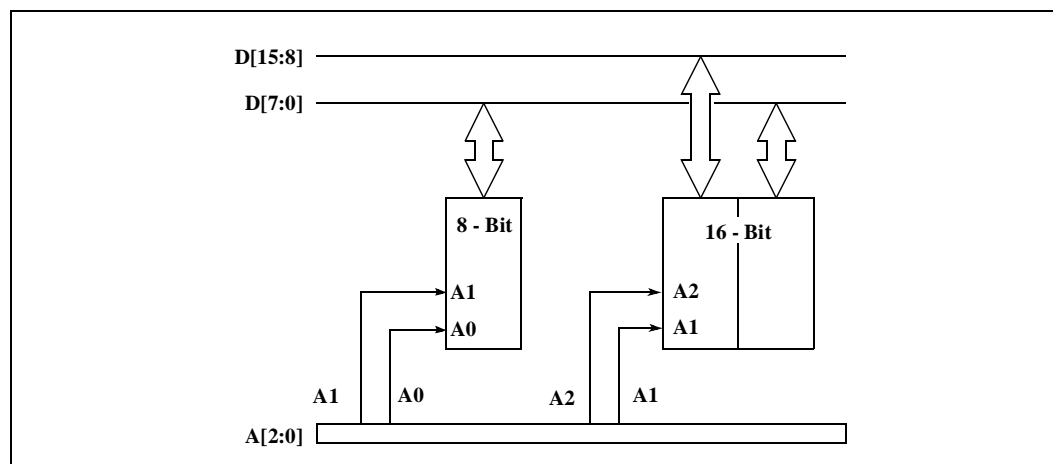
Bus signals consist of three groups: address A[24:0], data D[15:0], and control/status lines POE#, PWE#, PCE[1:0], PB\_RSTOUT#.



## 7.2 PBI Bus Width

The PBI allows an 8-, or 16-bit data bus width for each range. The PBI places 8- and 16-bit data on low-order data signals, simplifying the interface to narrow bus external devices. As shown in [Figure 41](#), 8-bit data is placed on lines D[7:0]; 16-bit data is placed on lines D[15:0].

**Figure 41. Data Width and Low Order Address Lines**



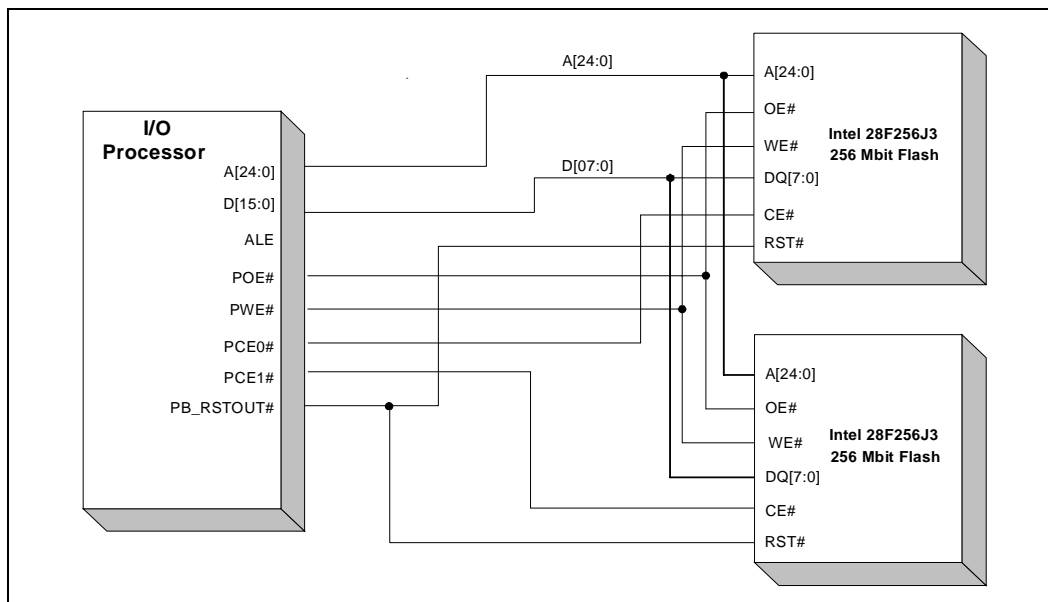
The user needs to wire up the flash memories in a manner consistent with the programmed bus width:

- 8-bit region: A[1:0] provide the demultiplexed byte address for a read burst.
- 16-bit region: A[2:1] provide the demultiplexed short-word address for a read burst.

## 7.3 Flash Memory Support

PBI peripheral bus interface supports 8-, or 16- bit Flash devices. Figure 42 shows two 8-bit flash devices connect with the Intel® 81341 and Intel® 81342 I/O Processors through the PBI Interface.

**Figure 42. Sixty-Four Mbyte Flash Memory System**

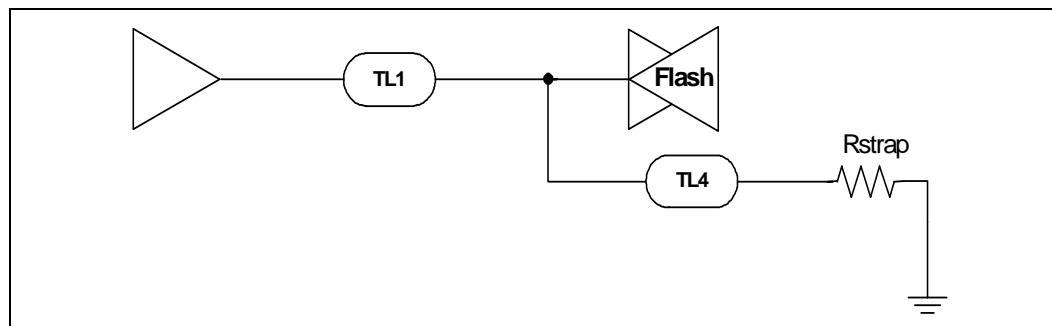




## 7.4 PBI Topology Layout Guidelines

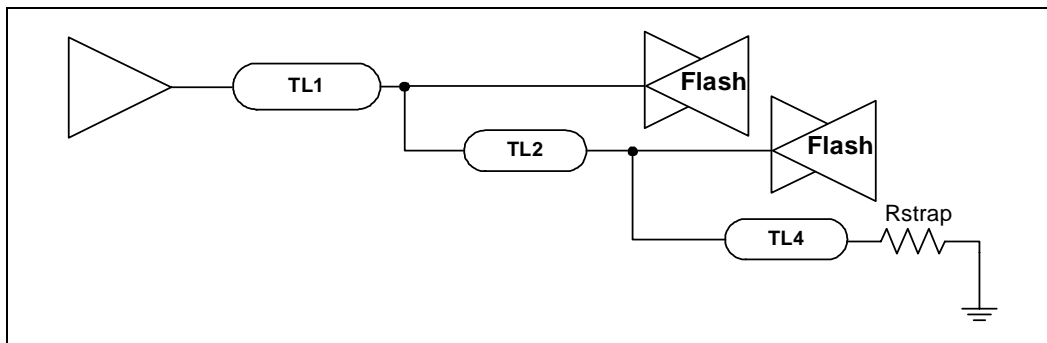
This section provides the topologies for routing the Address and Data bus for single load, double load and three load topologies. Note that no length matching is required between the Address and Data lines.

**Figure 43. Peripheral Bus Single Load Topology**

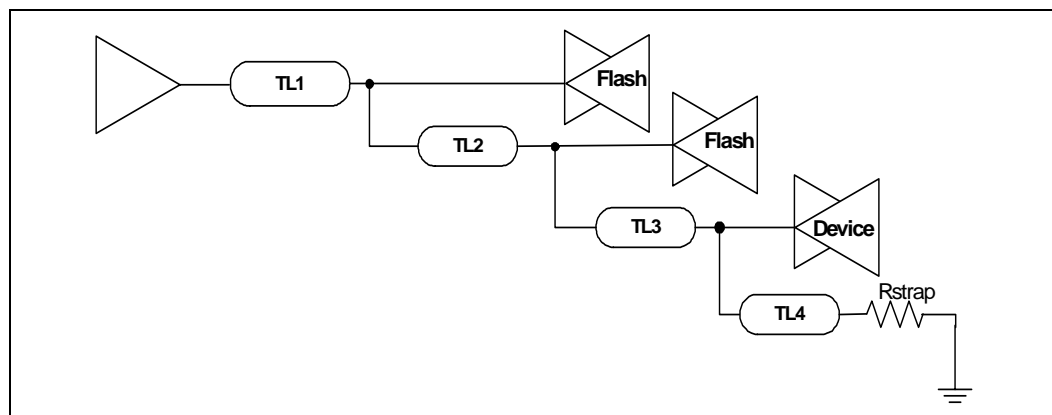


**Table 42. PBI Routing Guideline Single Load**

Parameter	Routing Guidelines
Reference Plane	Route over unbroken ground plane or unbroken power plane. When routing over power plane maintain this consistency throughout the topology.
Routing	Microstrip or stripline or combination of microstrip and stripline.
Motherboard Impedance (for both microstrip and stripline)	50 ohms +/- 15%
Add-in card Impedance (for both microstrip and stripline)	60 ohms +/- 15%
Trace Spacing (edge to edge)	≥ 5 mils between all Address and Data lines ≥ 20 mils must be maintained from all other signals or vias (for 5 mils trace width)
Breakout	5 mils on 5 mils spacing. Maximum length of breakout region is 500mils.
Trace Length TL1	0" to 20.0"
Trace Length to strapping resistors TL4	0.5" to 3.0" from the last device on the bus.
Routing Recommendations	Number of vias ≤ 8
Routing Recommendations	Route as Daisy Chain

**Figure 44. Peripheral Bus Dual Load Topology**

**Table 43. PBI Routing Guidelines for Dual Loads**

Parameter	Routing Guidelines
Reference Plane	Route over unbroken ground plane or unbroken power plane. When routing over power plane maintain this consistency throughout the topology.
Routing	Microstrip or stripline or combination of microstrip and stripline
Motherboard Impedance (for both microstrip and stripline)	50 ohms +/- 15%
Add-in card Impedance (for both microstrip and stripline)	60 ohms +/- 15%
Trace Spacing (edge to edge)	$\geq 5$ mils between all Address and Data lines $\geq 20$ mils must be maintained from all other signals or vias (for 5 mils trace width)
Breakout	5 mils on 5 mils spacing. Maximum length of breakout region is 500mils
Trace Length TL1	2.0" to 20.0"
Trace Length to TL2	0.5" to 2.0"
Trace Length to strapping resistor TL4	0.5" to 3.0" from the last device on the bus
Routing Recommendations	Number of vias for microstrip $\leq 8$
	Route as daisy-chain only

**Figure 45. Peripheral Bus Three Load Topology****Table 44. PBI Routing Guideline for Three Loads**

Parameter	Routing Guidelines
Reference Plane	Route over unbroken ground plane or unbroken power plane. When routing over power plane maintain this consistency throughout the topology.
Breakout	5 mils on 5 mils spacing. Maximum length of breakout region is 500mils.
Routing	Microstrip or stripline minimize the layer changes.
Motherboard Impedance (for both microstrip and stripline)	50 ohms +/- 15%
Add-in card Impedance (for both microstrip and stripline)	60 ohms +/- 15%
Trace Spacing (edge to edge)	$\geq 5$ mils between all Address and Data lines $\geq 20$ mils must be maintained from all other signals or vias (for 5 mils trace width)
Breakout	5 mils on 5 mils spacing. Maximum length of breakout region is 500mils.
Trace Length TL1	2.0" to 20.0"
Trace Length TL2, TL3	0.5" to 2.0"
Trace Length to strapping resistor TL4	0.5" to 3.0" from the last device on the bus.
Routing Recommendations	Number of vias for microstrip $\leq 8$
	Route as daisy-chain only



## 8.0 JTAG Circuitry for Debug

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Certain restrictions exist in order to use JTAG based debuggers with the Intel XScale® microarchitecture. This is primarily due to the Tap Controller reset requirements of the Intel XScale® processor and the reset requirements of specific JTAG debuggers. The following outlines these requirements along with suggestions for circuitry to alleviate potential problems

### 8.1 Requirements

The Intel® 81341 and Intel® 81342 I/O Processors, requires that TRST# (Tap Reset) is asserted during power-up. This is to ensure a fully initialized boundary scan chain. Failure to comply with this requirement results in spurious behavior of the application.

The ARM\* Multi-ICE\* JTAG debugger requires that TRST# is always weakly pulled high. This requirement stems from the fact that the debugger only asserts TRST# (drive low). Both reset signals coming from the Multi-ICE™ (TRST# and SRST#) are open collector and must be weakly pulled high in order to avoid unintentional resets (System or TAP).

JTAG Board Layout Tips:

- Make the connector easily accessible with a debugger by positioning it near the edge of the board.
- Label the debug connector and pin 1 on the silk-screen of the PCB.
- End the debug connector at the end of the JTAG chain nets, not in the center of the nets.
- TCK, TDI, TDO, TRST# and TMS signals do not have length restrictions but these signals are kept as short as possible and close to equal in length.



## 8.2 JTAG Signals / Header

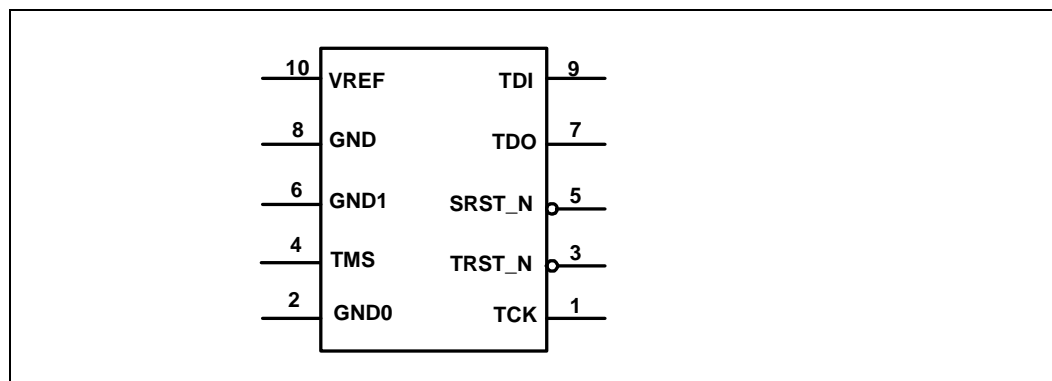
Figure 46 provides the pin definition (20-pin standard ARM connector) for JTAG. Figure 47 is the pin out for a smaller profile 10 pin connector. Note that the nTRST is equivalent to the TRST# and the nSRST is equivalent to the SRST#. The connector in Figure 47 is provided as an alternative to the 20 pin to save on board space. This connector is implemented on customer reference board. This part is a 10 pin, 2 x 5, surface mounted header with 2 mm spacing.

**Figure 46. JTAG Header Pin Out**

VTref	1	2	Vsupply
nTRST	3	4	GND
TDI	5	6	GND
TMS	7	8	GND
TCK	9	10	GND
RTCK	11	12	GND
TD0	13	14	GND
nSRST	15	16	GND
DBGRRQ	17	18	GND
DGBACK	19	20	GND

A8982-01

**Figure 47. Mini JTAG Header Pin Out**



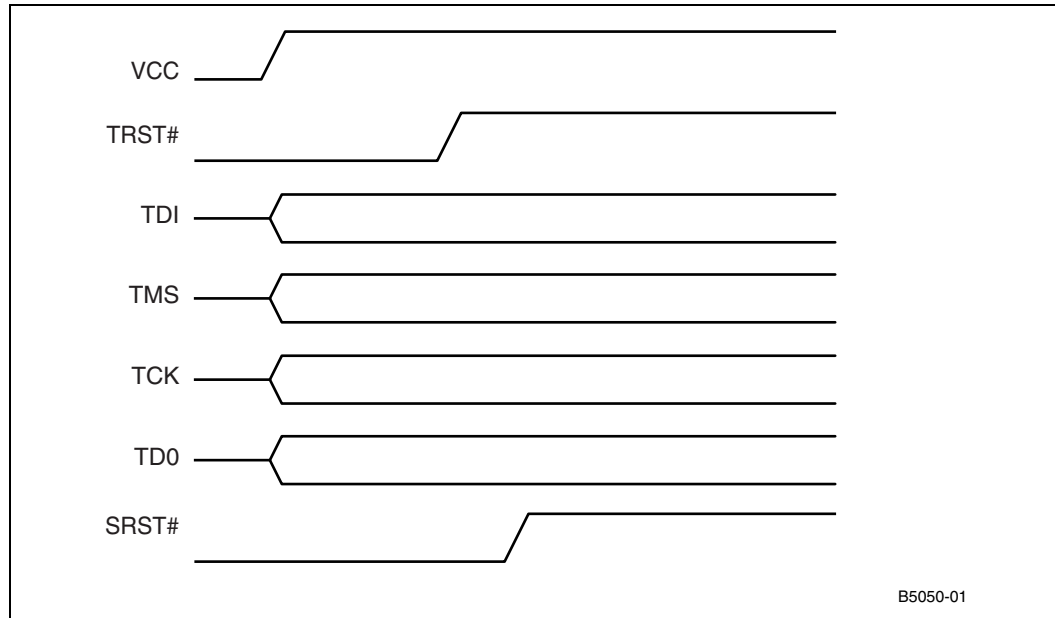
The ARM Multi-ICE debugger along with the Macraigor Raven\* and WindRiver Systems\* visionPROBE / visionICE utilize this connector. The main difference is the specific implementation of TRST# for each debugger. The Macraigor Raven implementation actively drives TRST# (high and low). The WindRiver Systems\* visionPROBE / visionICE configures TRST# active or open collector (only drive low). ARM Multi-ICE is configured as open collector only.

## 8.3 System Requirements

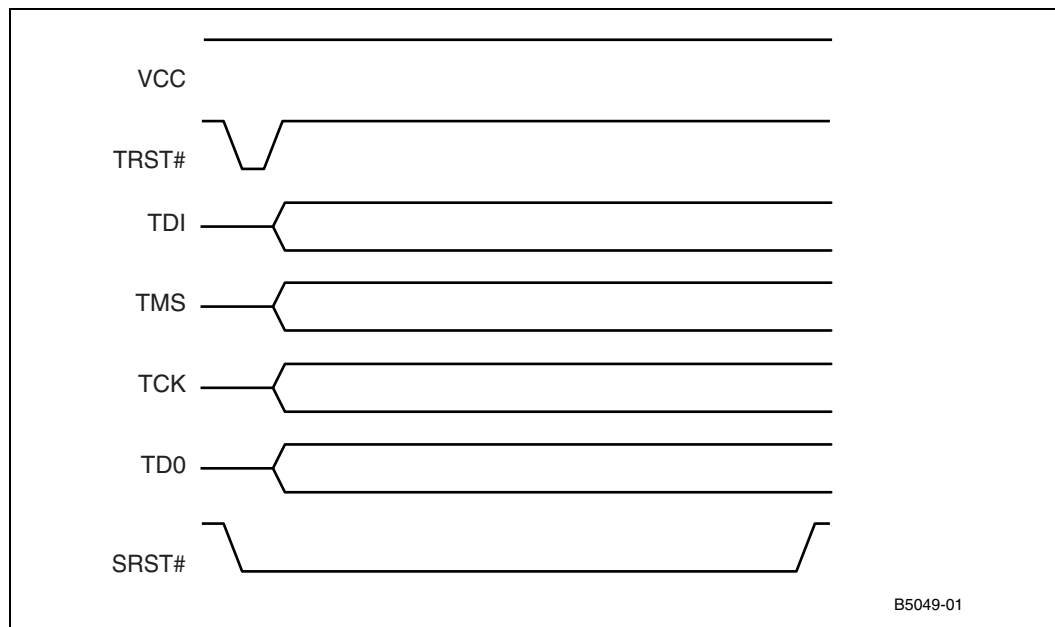
In order to successfully invoke a debug session, the JTAG debug unit must be able to control TRST# and SRST# independently. The TRST# signal allows the debugger to get the TAP controller in a known state. The SRST# signal allows the debugger to control system/processor reset in order to download the debug handler via the JTAG interface.

- Figure 48 and Figure 49 are used as examples without reflecting actual signal timings.

**Figure 48. JTAG Signals at Powerup**



**Figure 49. JTAG Signals at Debug Startup**



## 8.4 JTAG Hardware Requirements

Due to the conflicting requirements of Multi-ICE\* and the Intel XScale® microarchitecture, it is necessary to incorporate a circuit that drives TRST# low at power-up and weakly pull it high at all other times. The following section details the circuits required for the Macraigor Raven\*, WindRiver Systems\* visionPROBE\* / visionICE\*, and ARM\* Multi-ICE\*. [Figure 50](#) provides the JTAG section from the customer reference board. Note that for JTAG debuggers that actively drive the JTAG signals resistor MR\_N must be installed. For debuggers that have open collector outputs this resistor MR\_N is removed.

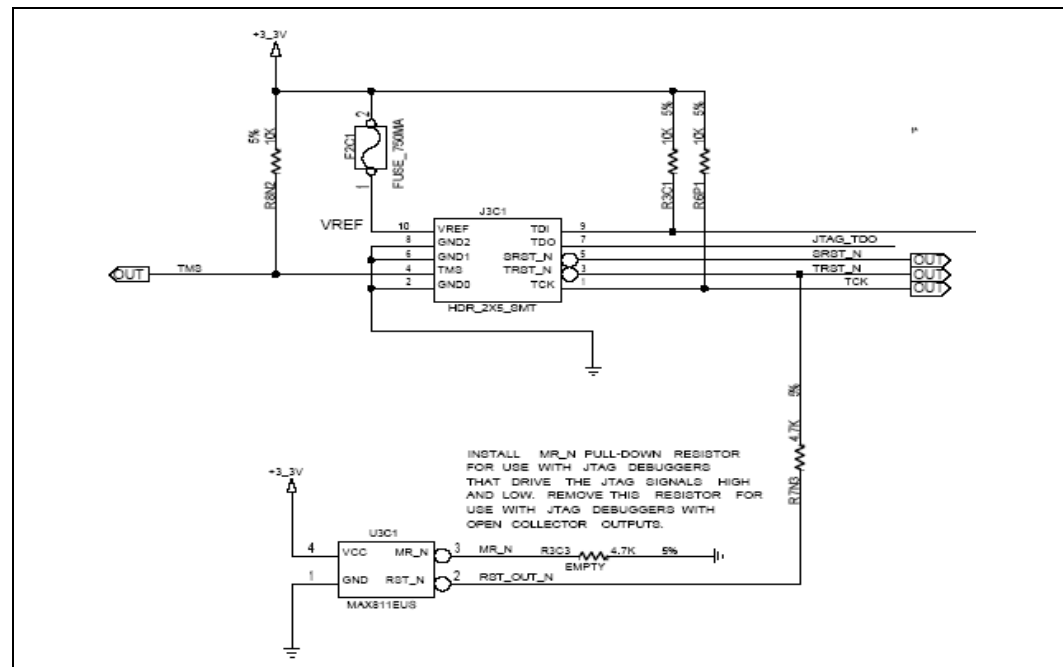
#### 8.4.1 Macraigor Raven and WindRiver Systems visionPROBE / visionICE

Both the Macraigor Raven and WindRiver Systems visionPROBE / visionICE (when configured as active) do not require any special power-up circuitry. The requirement is that TRST# is weakly pulled down at the processor. It is suggested that the value of the pull-down resistor is 10 K $\Omega$  or greater. The value of this resistor needs to be confirmed with the JTAG debugger manufacturer to ensure optimal performance.

### 8.4.2 ARM Multi-ICE

The ARM Multi-ICE debugger requires special power-up circuitry due to the open collector implementation of the TRST# signal. This power-up circuit must ensure that TRST# is asserted (low) at power on and weakly pulled high thereafter. Refer to [Figure 50](#) for an example of the Power-Up Circuit for TRST# based on the customer reference board.

**Figure 50. Example Power-Up Circuit for TRST#**



## 9.0 Debug and Test

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This chapter provides information on test equipment that is used to test the PCI-X and PCI Express interfaces of this part. It is recommended to check the bus interface and manufacturer websites for the latest test techniques and test equipment.

### 9.1 PCI-X Debugging

There are several tools available that aid in the debug and development of PCI-X bus based systems and cards. Agilent Technologies\*, VMetro\* and Catalyst Enterprises\* make analyzer/exercisers cards for capturing and generating PCI-X transactions. These cards also provide capability to trigger on errors, emulate an initiator or target, invoke errors, measure performance, and check for protocol and compliance issues. For pure analysis of the PCI-X bus, both Tektronix and Agilent make passive interposer probe cards that plug into the PCI-X slot of the device under test to capture PCI-X traffic. An example of an interposer card that works with the Agilent logic analyzers is the FuturePlus Systems\* FS2007. Another method to capture the PCI-X bus signals with a logic analyzer is to place AMP\* Mictor-38 connectors or Agilent Soft Touch Connectorless Probes on the PCB. For the pinout of the connectors that work with the Agilent logic analyzer refer to the FuturePlus Systems\* website [www.futureplus.com](http://www.futureplus.com).



## 9.2 PCI Express Debugging

Debugging a PCI Express design requires analysis at the physical layer to verify the layout and the data link/transaction layer to ensure that the read and write request packets are being transmitted correctly.

### 9.2.1 Physical Layer Debugging

For PCI Express, the fundamental signaling frequency is 1.25GHz (half the bit rate) and the specified 20-80% rise-time is 100 ps. The Tektronix<sup>TM</sup> TDS6604 Real-Time Digital Storage Oscilloscope and the Agilent Technologies<sup>TM</sup> 54855A provides an analog bandwidth of 6 GHz (with a 20GSa/sec. sampling rate) sufficient to measure the PCI Express differential signals with their respective differential probes.

The alternative equipment to the high speed oscilloscopes include Vector Network Analyzers or Time Domain Reflectometry (TDR) scopes which help pinpoint signal integrity issues with the PCBs and connectors. This test equipment allows checking the lane-to-lane skew, analyzing jitter and measuring drive strength and receiver tolerance for verification of the physical layer. For more information on using TDR analysis, the application note from Tektronix is useful:

[TDR Impedance Measurements: A Foundation for Signal Integrity.](#)

### 9.2.2 Data Link and Transaction Layer Testing

The Data Link/Transaction layer is debugged and validated with PCI Express protocol analyzers or PCI Express analyzer/exerciser tools. Companies that make protocol analyzers for PCI Express include: Catalyst Enterprises, LeCroy (formerly CATC), Agilent, Tektronix and Finisar (formerly DataTransit). For more information on the PCI Express test equipment refer to Intel's PCI Express Developer's website <http://www.pciexpressdevnet.org/kshowcase/>. The probing solutions for the PCI Express bus include an interposer card and a mid-bus probing solution.

Agilent Technology has a PCI Express Packet Analysis Probe N4220B which works in conjunction with their 16700 family of logic analyzers. The Agilent slot interposer part numbers that work with the 16700 logic analyzer include: N4224A for a x8 slot, N4225A for a x4 slot and N4227A for a x1 slot. The Tektronix slot interposer solution that works with their TLA700 logic analyzer is the TMS817.

### 9.2.3 PCI Express Analyzer/Exercisers

Agilent E2960A, Catalyst Enterprise SPX-8E and LeCroy PETRacer/PETTrainer provide the ability to capture and exercise the PCI Express bus.

### 9.2.4 Mid-bus Probing

The mid-bus probe provides probing between two devices without PCI Express connector. Catalyst Enterprises, Agilent and Tektronix support mid-bus PCI Express probing. Agilent makes a protocol analyzer/exerciser, E2960A, which uses the Soft touch mid-bus probe e2941A. The Agilent solution that works with the 16700 analyzer is the N4221A. Tektronix solution is the TMSIC6. The PCB must be designed with the PCI Express mid-bus footprint to allow probing between two devices. Refer to the following paper for more information on PCI Express mid-bus probing and the layout of the mid-bus probe:

[http://www.tek.com/Measurement/logic\\_analyzers/contact/\\_notes/probe\\_design\\_guide\\_pci.pdf](http://www.tek.com/Measurement/logic_analyzers/contact/_notes/probe_design_guide_pci.pdf).



## 10.0 Power Delivery

This section provides information on the power delivery for this chip including:

- the different voltage domains that are required on the Intel® 81341 and Intel® 81342 I/O Processors are provided in [Table 45](#)
- an example of the power plane layout used on the eight layer customer reference board [Section 10.1](#)
- decoupling recommendations [Section 10.2](#)
- required power sequencing [Section 10.3](#)
- the power failure recommendations [Section 10.4](#)

**Table 45. Supply Voltages**

Voltage Supply	Voltage	Minimum	Maximum
V <sub>CC3P3</sub>	3.3 V supply voltage for PCI-X interface and general purpose I/Os	3.0	3.6
V <sub>CC1P8E</sub>	1.8 V supply voltage for PCI Express* interface	1.71	1.89
V <sub>CC1P8</sub>	1.8 V supply voltage for DDR2 SDRAM memory interface I/Os	1.71	1.89
V <sub>CCVIO</sub>	3.3 V supply voltage for PCI-X interface	3.0	3.6
V <sub>CC1P2X</sub>	1.2 V supply voltage for Intel XScale® processors	1.164	1.236
V <sub>CC1P2</sub>	1.2 V supply voltage for most digital logic	1.164	1.236
V <sub>CC1P2E</sub>	1.2 V supply voltage for PCI Express* interface digital logic	1.164	1.236
V <sub>CC1P2AE</sub>	1.2 V supply voltage for PCI Express* interface analog logic	1.164	1.236
V <sub>CC1P2PLL</sub>	1.2 V supply voltage for PCI-X PLL	1.164	1.236
V <sub>CC1P2PLLD</sub>	1.2 V supply voltage for DDR2 SDRAM PLL	1.164	1.236
V <sub>CC3P3PLLX</sub>	3.3 V supply voltage for core logic PLL	3.0	3.6
M_VREF	Memory I/O reference voltage	0.49V <sub>CC1P8</sub>	0.51V <sub>CC1P8</sub>

- ESL for a 0603 package is 150pH, divide this by 6 = 25nH
- Total ESL: 25nH || 19 pH = ~ 18.9pH

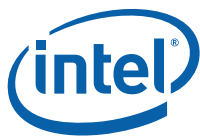


## 10.1 Power Plane Layout

This section provides the layout of the power planes around the 81348 package on the eight layer customer reference board (CRB). While these figures provide additional supplies required for the 81348 storage interface are included in the Intel® 81341 and Intel® 81342 I/O Processors design guide for reference purposes. The voltage plane descriptions are listed in Table 46 and the stackup for the customer reference board is listed in Table 47. Figure 51 provides the voltage layout for layer 3, Figure 52 provides the voltage layout for layer 5, Figure 48 provides the voltage layout for layer 6 and Figure 49 provides the voltage layout for layer 8. Note that with careful power supply layout 1.2V and 1.8V switching regulators are used to generate each of the 1.2V and 1.8V supplies. It is important to connect the +1\_2V and +1\_2VA supplies at a single point such as the 1.2V switching regulator output capacitor. This same recommendation applies to connecting the +1\_8V and +1\_8VA at a single point such as the 1.8V switching regulator output capacitor.

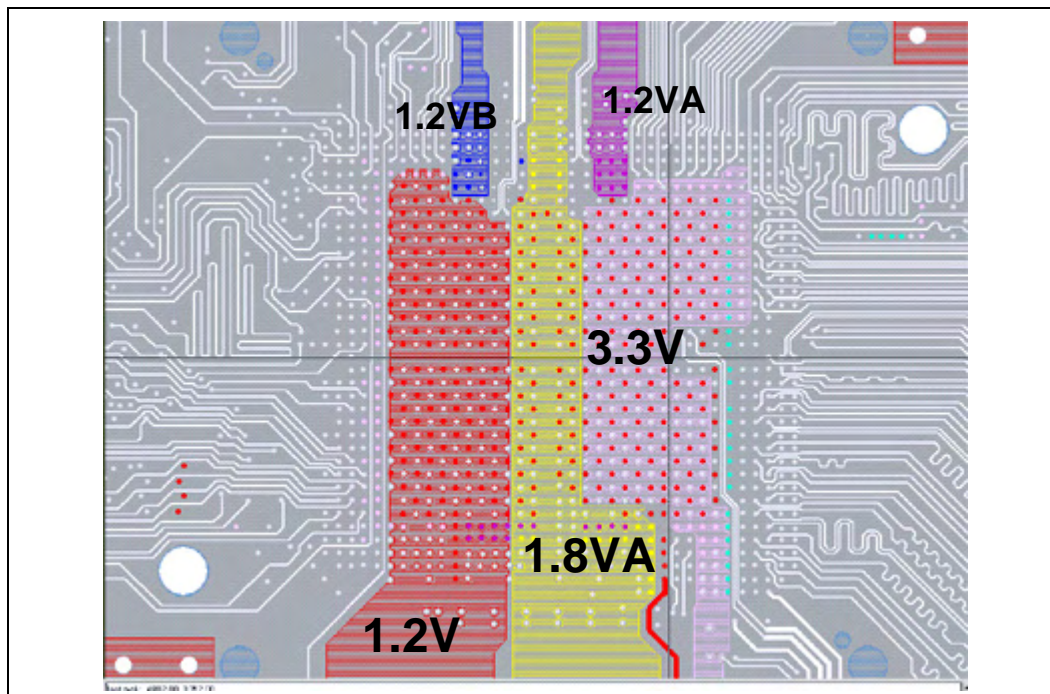
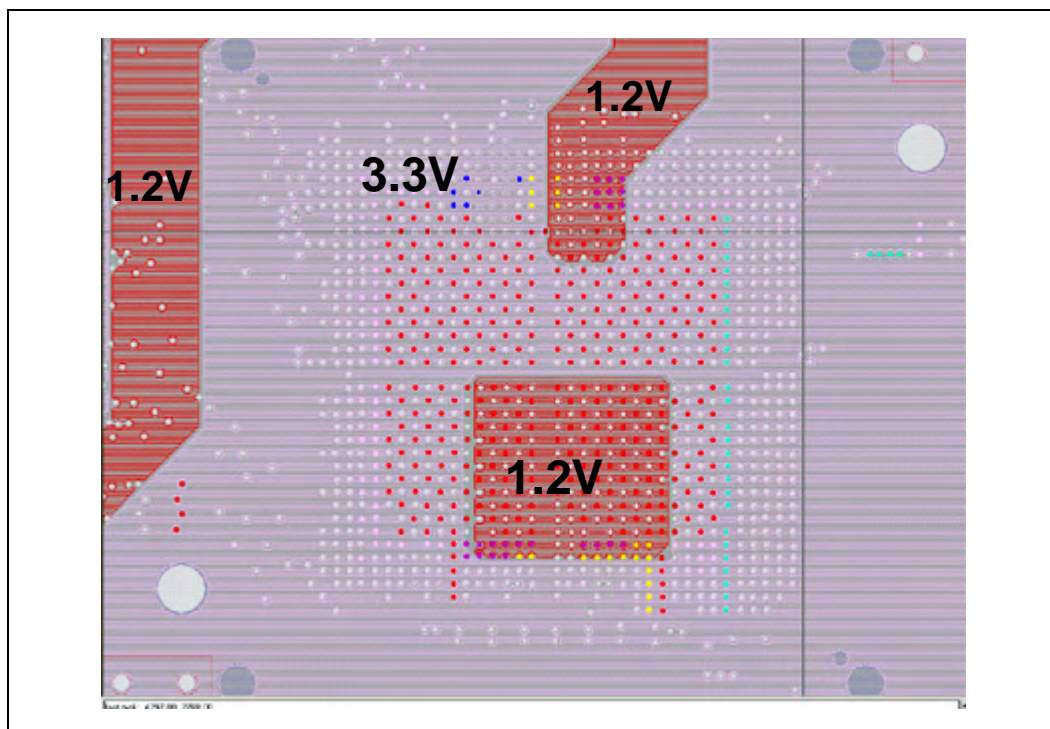
**Table 46. Customer Reference Board Voltage Planes**

CRB Voltage Plane	Package Voltage Planes	Voltage Source	Voltage Description	Notes
+1_2V	VCC1P2X, VCC1P2	1.2V Switching Regulator	1.2V digital voltage for core logic	Connect +1_2V and +1_2VA only at a single point.
+1_2VA	VCC1P2AE, VCC1P2AS, VCC1P2E	1.2V Switching Regulator	1.2V analog voltage for PCI-E and storage interfaces	
+1_8VA	VCC1P8E, VCC1P8S	1.8V Switching Regulator	1.8V analog voltage for PCI-E and storage interfaces	Connect +1_8V and +1_8VA only at a single point.
+1_8V	VCC1P8	1.8V Switching Regulator	1.8V digital voltage for DRAM interface.	
+3_3V	VCC3P3, VCCVIO	System power	3.3V digital voltage for PCI-X and peripheral bus interfaces	

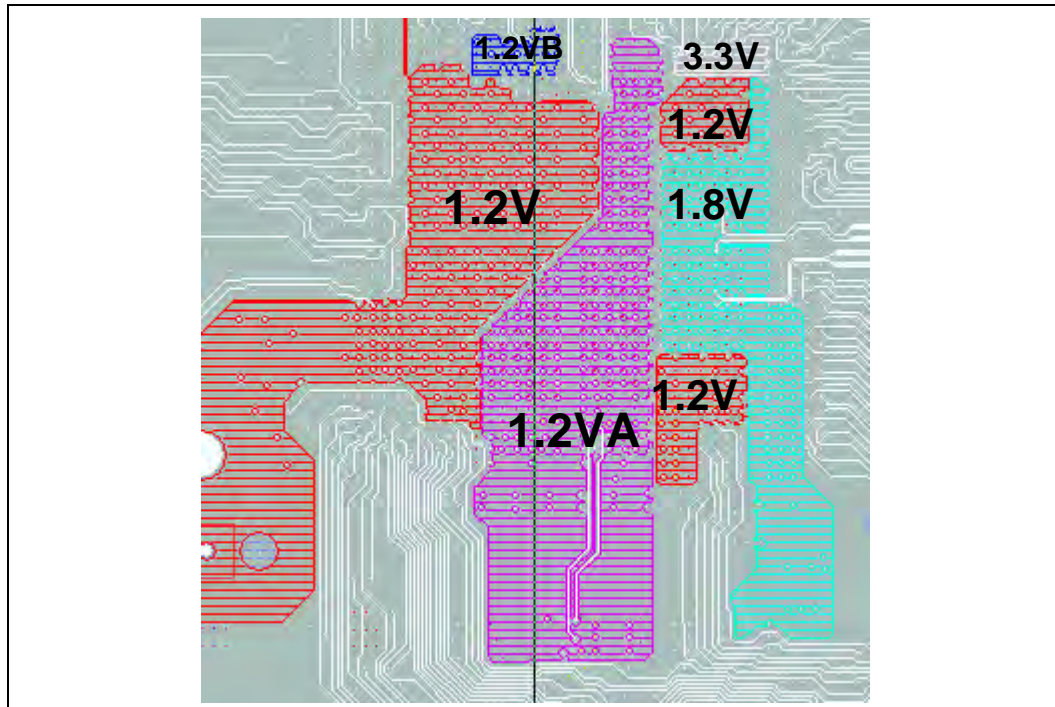
**Table 47. Customer Reference Board Layer Stackup**

Layer	Layer Description	Voltage Planes	Color Code
1	Primary side	none	
2	Ground plane 1		
3	Internal routing layer 1	+1_2V	Red
		+1_2VB	Blue
		+1_8VA	Yellow
		+1_2VA	Purple
		+3_3V	Pink
4	VCC split plane	+1_2V	Red
		+3_3V	Pink
5	Ground plane 2		
6	Internal routing layer 2	+1_2V	Red
		+1_2VB	Blue
		+1_2VA	Purple
		+3_3V	Pink
		+1_8V	Green
7	Ground plane 3		
8	Secondary layer	+1_8VA	Yellow
		+1_8V	Green

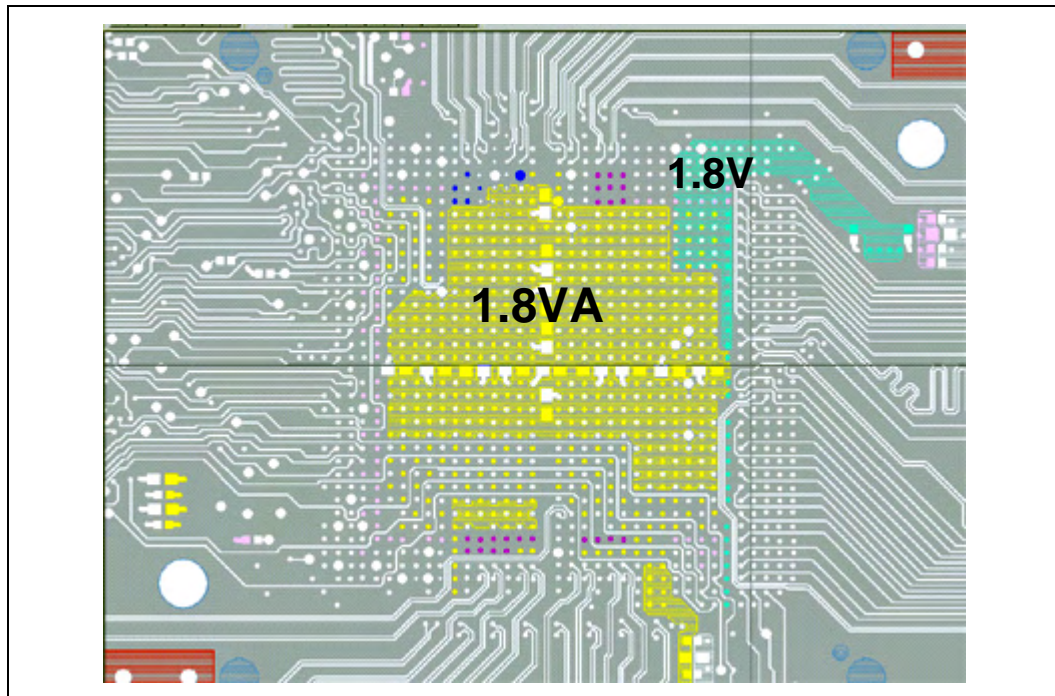


**Figure 51. Split Voltage Planes for Layer 3 (Top View)****Figure 52. Split Voltage Planes for Layer 4 (Top View)**

**Table 48. Split Voltage Planes for Layer 6 (Top View)**



**Table 49. Split Voltage Planes for Layer 8 (Top View)**





## 10.2 Decoupling Recommendations

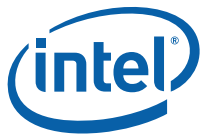
Table 50 contains the decoupling recommendations for Intel® 81341 and Intel® 81342 I/O Processors. Note that the recommendations provide the total minimum capacitance for each voltage plane. The recommended decoupling capacitance, ESR and ESL for each voltage plane is an minimum aggregate value that is achieved with adding multiple decoupling capacitors in parallel.

Each decoupling capacitor is placed with a single via to a voltage plane (or plane fill area) and solid ground plane, such that copper loss and inductance between the capacitor and nearby ball via is negligible. Distribute the capacitors so that all power ball vias have decoupling nearby. It is recommended that the distance from ball vias to decoupling be minimized.

**Table 50. Decoupling Recommendations**

Voltage	Interface	Capacitors
1.2V Digital	Intel XScale® processor 1 Voltage	1 x 20uF min with < 150pH ESL, ~1mohm ESR
	Intel XScale® processor 2 Voltage	1 x 20uF min with < 150pH ESL, ~1mohm ESR
1.2V High Speed Serial	PCI Express	1 x 5 uF min with < 150pH ESL, ~3mohm ESR
1.8V Digital	DDR2	1 x 10uF with < 100pH ESL, ~1mohm ESR
1.8V Analog	PCI Express	1 x 5 uF min with <150pH ESL, ~3mohm ESR
3.3V	PCI-X	1 x 10 uF min with <150pH ESL, ~1mohm ESR

**Note:** The symbol “||” represents inductors or resistors in parallel. The equivalent inductance is calculated as the product of the two inductances divided by the sum of the two inductances. The equivalent resistance is calculated as the product of the two resistances divided by the sum of the two resistances.



## 10.3 Power Sequencing

Intel® 81341 and Intel® 81342 I/O Processors requires the following power sequence:

Power-up: 1.8V  $\leq$  1.2V

- 1.8V supply must not turn on before or any faster than the 1.2V supplies.

Power Down: 1.8V  $\leq$  1.2V

- 1.8V turns off first, or, 1.8V must reach 1.2V before 1.2V begins ramping down.

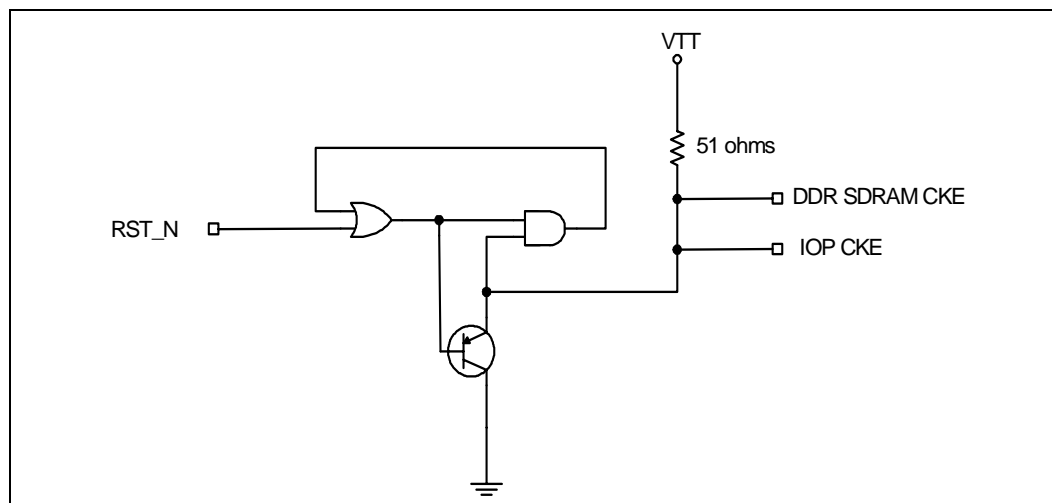
## 10.4 Power Failure

The following list provides details on the power failure mode:

- Intel® 81341 and Intel® 81342 I/O Processors always attempts to the put the DRAM in the self refresh mode whenever power is lost.
- During battery backup mode it is recommended that power to Intel® 81341 and Intel® 81342 I/O Processors be isolated and only the DRAM be powered in order to reduce battery power drain.
- CKE[1:0] must remain deasserted regardless of the state of Vcc powering the Intel® 81341 and Intel® 81342 I/O Processors during the battery backup mode. [Figure 53](#) shows an implementation of the CKE latching circuit to maintain the DRAM in self refresh. Note that the 51 ohms pull-up to VTT is shown as the typical termination for the CKE lines. Additional information on terminating the CKE lines is detailed in [Section 4.2.2, “DDR2 533 DIMM Layout Design” on page 24](#).

VTT is turned off battery backup needs to maintain power on DDR voltages **V<sub>DD</sub>**, **V<sub>DDQ</sub>** and **V<sub>REF</sub>** to prevent data loss.

### Figure 53. SCKE Circuit



### 10.4.1 Non-Battery Backup Circuits

For applications not supporting battery back-up, the circuit in [Figure 53](#) is not required. Instead the following is recommended:

- Connect CKE pins directly from Intel® 81341 and Intel® 81342 I/O Processors to the CKE pins on the SDRAM.



## 11.0 Terminations

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This chapter provides details on the configuration modes that are implemented with strapping options and terminations for each of the signals required for a Intel® 81341 and Intel® 81342 I/O Processors layout.

- Refer to [Section 11.1](#) for configuration details for operational modes.
- Refer to [Section 11.2](#) and [Table 52](#) for the table of terminations for each of the signals.
- Refer to [Table 53](#) for the table of configuration reset straps.
- Refer to [Section 11.4](#) for the PLL filters.
- Refer to [Section 11.5](#) for the PCI Resistor Calibration resistors.
- Refer to [Section 11.6](#) for the PCI Express Calibration resistors.
- Refer to [Section 11.7](#) for the Memory Calibration resistors.



## 11.1 Configuration Details

Table 51 provides the reset strap configuration for valid operational modes of the chip: PCI Express root complex or endpoint and PCI-X endpoint or central resource. Note that PCI Express endpoint and PCI-X endpoint mode simultaneously is not supported.

**Note:** The **PCIXCAP** signal has been defeatured. Refer the Non-Core Erratum in the *Intel® 81341 and 81342 I/O Processors Specification Update* for more information. This erratum overrides references throughout this document.

**Table 51. PCI Express/PCI-X Strapping Configuration Table**

Application	Endpoint Configuration	Strapping Settings		
		INTERFACE_SEL_PCIX#	PCIE_RC# (PCI Express root Complex strap)	PCIX_EP# (PCI-X endpoint strap)
HBA or Motherboard	PCI Express endpoint with PCI-X Central Resource (default)	1	1 (PCIE Endpoint)	1 (Central Resource)
HBA or Motherboard	PCI-X endpoint with PCI Express Root Complex	0	0 (PCI Root Complex)	0 (PCIX Endpoint)
HBA or Motherboard	PCI Express endpoint	1	1	X
HBA or Motherboard	PCI-X endpoint	0	X	0
Motherboard	PCI Express Root Complex and PCI-X Central Resource	0 (ATU-X is function 0, ATU-E is function 5)	0	1
		1 (ATU-E is function 0, ATU-X is function 5)		
Motherboard	PCI-X Central Resource	0	X	1
Motherboard	PCI Express Root Complex	1	0	X





### 11.1.1 PCI Express Mode Only

When PCI Express interface is used and PCI-X is not used the following provisions must be made:

- **P\_CLKIN** - **GND**
- **P\_PCIXCAP** - **GND**
- **P\_INT[D:A]/XINT[3:0]** - these pins have 8.2K pull-ups.
- **CLK\_SRC\_PCIE#** strapping = 1 (**NC** default).
- **REFCLKP**, **REFCLKN** differential pins must be connected to 100MHz oscillator.
- **PETP[7:0]**, **PETN[7:0]** differential transmit pair pins lanes 0 through 7 connect to series capacitors with value of 75nF to 200nF and then to corresponding RX lane pins on device or connector. Unused lanes are **NCs**.
- **PERP[7:0]**, **PERN[7:0]** differential receiver pairs lanes 0 through 7 these connect to the corresponding TX lane pins on device or connector. Unused lanes are **NCs**.
- **PE\_CALP**, **PE\_CALN** - Connect **PE\_CALP** ball through 1.4K 1% resistor to the **PE\_CALN** ball.
- **VCCVIO** pins are grounded.
- **VCC1P2PLL** and **VSSPLL** filter pins are grounded.

Any other PCI-X pins are no connects. Make sure configuration strapping options are set correctly for operation mode shown in [Table 51](#). Refer to [Table 52](#) for additional details.





### 11.1.2 PCI-X Mode Only

When PCI-X interface is used and PCI Express is not used the following provisions must be made:

- **REFCLKP/REFCLKN** must have the 100 MHz differential signal on it to generate the P\_CLKOs.
- **PE\_CALP, PE\_CALN** - Connect **PE\_CALP** ball through 1.4K 1% resistor to the **PE\_CALN** ball.
- **CLK\_SRC\_PCIE#** strap: Make sure strapping reflects whether the clock source is the **REFCLKP/REFCLKN** (**CLK\_SRC\_PCIE#** = 0) or PCI clock in (**CLK\_SRC\_PCIE#** = 1 default).
- All the other PCI Express pins are no connects. Make sure configuration strapping options are set correctly for operation mode shown in [Table 51](#). Refer to [Table 52](#) for additional details.

### 11.1.2.1 Central Resource Mode: (PCIX\_EP# = 1)

#### 1. P\_PCIXCAP:

- **CLK\_SRC\_PCIE#** = 0, (using **P\_CLK[3:0]** outputs): **P\_PCIXCAP** connect signal with 3.3 K ohms pull-up to 3.3 V.
- **CLK\_SRC\_PCIE#** = 0, (using **P\_CLK[3:0]** outputs but limiting the PCI clock frequency): refer to [Table 25, "PCI Express Layout for Clock Routing" on page 55](#). Note that strapping **PCIXM1\_100#** is pulled low to limit frequency to 100MHz.
- **CLK\_SRC\_PCIE#** = 1, (**P\_CLKIN** primary clock source): refer to [Section 6.1.1, "PCI/PCI-X Frequency Selection" on page 57](#) for PCI frequency selection information.

#### 2. P\_M66EN:

- **CLK\_SRC\_PCIE#** = 0, (using **P\_CLK[3:0]** outputs): pull-up the signal 8.2K ohms pull-up to 3.3 V.
- **CLK\_SRC\_PCIE#** = 0, (using **P\_CLK[3:0]** outputs but limiting the PCI clock frequency): refer to [Table 25, "PCI Express Layout for Clock Routing" on page 55](#). Note that strapping **PCIXM1\_100#** is pulled low to limit frequency to 100MHz.
- **CLK\_SRC\_PCIE#** = 1, (**P\_CLKIN** primary clock source): Refer to [Section 6.1.1, "PCI/PCI-X Frequency Selection" on page 57](#) for PCI frequency selection information.

#### 3. P\_IDSEL: pull-down the signal 1K ohm resistor.

#### 4. P\_REQ[0]#/P\_GNT#: This is an input request signal and have a 8.2K pull-up resistor.

#### 5. P\_GNT[0]#/P\_REQ#: (internal arbiter): This is an output grant signal.

#### 6. P\_GNT[3:1]#: (internal arbiter) - These are output grant signals and unused signals are NCs.

#### 7. P\_REQ[3:1]#: (internal arbiter) - These are input request signals and unused signals are NCs.

#### 8. P\_CLKIN:

- **CLK\_SRC\_PCIE#** = 0, Connect to **P\_CLKOUT** through a 26 ohm +/- 1% resistor.
- **CLK\_SRC\_PCIE#** = 1, (**P\_CLKIN** primary clock source) connect to system clock.

#### 9. P\_CLKOUT:

- **CLK\_SRC\_PCIE#** = 0, Connect to **P\_CLKIN** through a 26 ohm +/- 1% resistor.
- **CLK\_SRC\_PCIE#** = 1, (**P\_CLKIN** primary clock source) signal is left unconnected.



### 11.1.2.2 PCI-X Endpoint Mode (PCIX\_EP# = 0)

1. **P\_PCIXCAP**:
  - Pull-up signal with 8.2K resistor and refer to [Section 6.1.1, “PCI/PCI-X Frequency Selection” on page 57](#) for information on termination for the PCIXCAP pin on the edge connector.
  - **CLK\_SRC\_PCIE#** = 1, **P\_CLKIN** primary clock source: Refer to [Section 6.1.1, “PCI/PCI-X Frequency Selection” on page 57](#) for PCI frequency selection information
2. **P\_M66EN**: connect to the M66EN on the board and refer to [Section 6.1.1, “PCI/PCI-X Frequency Selection” on page 57](#) for PCI frequency selection information.
3. **P\_IDSEL**: connect to one of the AD lines
4. **P\_REQ[0]#/P\_GNT#**: This is an input grant signal and have a 8.2K pull-up resistor.
5. **P\_GNT[0]#/P\_REQ#**: (external arbiter): This is an output request signal and connect to the external arbiter **P\_REQ#** line.
6. **P\_GNT[3:1]#**: (external arbiter): These signals are unused signals are **NCs**.
7. **P\_REQ[3:1]#**: (external arbiter): These signals are unused signals are **NCs**.
8. **P\_CLKIN**: Connect to system PCI clock.
9. **P\_CLKOUT**: this signal is left unconnected.
10. **REFCLKP** connect to a resistor divider such that the **REFCLKP** node is connected to both a 17.4K to VCC3P3 and a 4.7K connected to **GND**. **REFCLKN** must be connected to **GND**.

### 11.1.3 Dual Interface Mode

For dual interface mode with PCI-E and PCI-X interfaces active refer to the [Table 51](#) for **INTERFACE\_SEL\_PCIX#**, **PCIE\_RC#** and **PCIX\_EP#** straps for the following modes:

1. PCI-E root complex with PCI-X endpoint mode in [Section 11.1.3.1](#)
2. PCI-X central resource with PCI-E endpoint mode in [Section 11.1.3.2](#)
3. PCI-E root complex with PCI-X central resource mode in [Section 11.1.3.3](#)

#### 11.1.3.1 PCI-E Root Complex with PCI-X Endpoint Mode

- PCI-E Root complex:
  - **REFCLKP**, **REFCLKN** differential pins must be connected to 100MHz oscillator.
  - **PETP[7:0]**, **PETN[7:0]** differential transmit pair pins lanes 0 through 7 connect to series capacitors with value of 75nF to 200nF and then to corresponding RX lane pins on device or connector. Unused lanes are **NCs**.
  - **PERP[7:0]**, **PERN[7:0]** differential receiver pairs lanes 0 through 7 these connect to the corresponding TX lane pins on device or connector.
  - **PE\_CALP**, **PE\_CALN** - Connect **PE\_CALP** ball through 1.4K 1% resistor to the **PE\_CALN** ball.
- PCI-X endpoint mode follow recommendations in [Section 11.1.2.2](#)

#### 11.1.3.2 PCI-E endpoint with PCI-X Central Resource

- PCI-E Endpoint
  - **REFCLKP**, **REFCLKN** differential pins must be connected to 100MHz oscillator.
  - **PETP[7:0]**, **PETN[7:0]** differential transmit pair pins lanes 0 through 7 connect to series capacitors with value of 75nF to 200nF and then to corresponding RX lane pins on device or connector. Unused lanes are **NCs**.
  - **PERP[7:0]**, **PERN[7:0]** differential receiver pairs lanes 0 through 7 these connect to the corresponding TX lane pins on device or connector. Unused lanes are **NCs**.
  - **PE\_CALP**, **PE\_CALN** - Connect **PE\_CALP** ball through 1.4K 1% resistor to the **PE\_CALN** ball.
- PCI-X central resource mode follow recommendations in [Section 11.1.2.1](#)

#### 11.1.3.3 PCI-E Root Complex with PCI-X Central Resource

- PCI-E Root complex:
  - **REFCLKP**, **REFCLKN** differential pins must be connected to 100MHz oscillator.
  - **PETP[7:0]**, **PETN[7:0]** differential transmit pair pins lanes 0 through 7 connect to series capacitors with value of 75nF to 200nF and then to corresponding RX lane pins on device or connector. Unused lanes are **NCs**.
  - **PERP[7:0]**, **PERN[7:0]** differential receiver pairs lanes 0 through 7 these connect to the corresponding TX lane pins on device or connector. Unused lanes are **NCs**.
  - **PE\_CALP**, **PE\_CALN** - connect **PE\_CALP** ball through 1.4K 1% resistor to the **PE\_CALN** ball.
- PCI-X central resource mode follow recommendations in [Section 11.1.2.1](#)



## 11.2 Termination Resistors

Table 52 lists these Intel® 81341 and Intel® 81342 I/O Processors termination values.

**Table 52. Terminations: Pull-up/Pull-down (Sheet 1 of 7)**

Signal	Recommendations	Comments
REFCLKP, REFCLKN	<ul style="list-style-type: none"> <li>For PCI Express interface: connect to a 100MHz oscillator.</li> <li>For PCI-X with central resource: connect to a 100MHz oscillator.</li> <li>For PCI-X end point: connect the <b>REFCLKP</b> to a resistor divider such that the <b>REFCLKP</b> node is connected to both a 17.4K to VCC3P3 and a 4.7K connected to <b>GND</b>. <b>REFCLKN</b> must be connected to <b>GND</b>.</li> </ul>	<b>Note:</b> 100 MHz oscillator is required for the PCI Express differential clock and to generate the P_CLKs.
PETP[7:0], PETN[7:0]	Connect to series capacitors with value of 75nF to 200nF (low ESR) on each of the lines and then to corresponding RX lane pins on device or connector. <b>NC</b> when not used	
PERP[7:0], PERN[7:0]	<ul style="list-style-type: none"> <li>No series capacitor needed</li> <li>Connect to the corresponding TX lane pins on device or connector.</li> <li><b>NC</b> when not used</li> </ul>	
P_AD[63:32], P_CBE[7:4]#, P_PAR64	<ul style="list-style-type: none"> <li>When only PCI Express interface active these signals are internally pulled-up and left as a <b>NCs</b>.</li> <li>When the <b>PCIX_PULLUP#</b> is enabled (pulled to 0), these signals are internally pulled-up.</li> <li>When the <b>PCIX_32BIT#</b> is enabled (32 bit bus width), these signals are internally pulled-up and left as a <b>NCs</b>.</li> </ul>	
P_AD[31:0], P_CBE[3:0]#	When only PCI Express interface active these signals are internally pulled-up and left as a <b>NCs</b> .	
P_GNT[0]# / P_REQ#	<ul style="list-style-type: none"> <li>PCI Express: <b>P_GNT[0]#</b> / <b>P_REQ#</b> has an internal pull-up and left as a <b>NC</b>.</li> <li>In Central Resource mode (internal arbiter) with <b>PCIX_EP#</b> = 1: <b>P_GNT[0]#</b> is output grant signal 0.</li> <li>PCI Endpoint mode (external arbiter) <b>PCIX_EP#</b> = 0: This is the output request signal for the ATU and needs to connect to the external arbiter <b>P_REQ#</b> lines.</li> </ul>	
P_REQ[0]# / P_GNT#	<ul style="list-style-type: none"> <li>PCI Express: <b>P_REQ[0]#</b> / <b>P_GNT#</b> has an internal pull-up and left as a <b>NC</b>.</li> <li>In Central Resource mode (internal arbiter) with <b>PCIX_EP#</b> = 1: <b>P_REQ[0]#</b> is input request signal to the ATU.</li> <li>PCI Endpoint mode (external arbiter) <b>PCIX_EP#</b> = 0: <b>P_GNT#</b> is input grant signal for the ATU. Pull this pin up to VCC3P3 with an 8.2K resistor.</li> </ul>	
P_GNT[3:1]#	<ul style="list-style-type: none"> <li>When PCI Express interface only: <b>P_GNT[3:1]#</b> is left as a <b>NC</b>.</li> <li>In Central Resource mode (internal arbiter) with <b>PCIX_EP#</b> = 1: These are three output grant signals. Unused signals are left as <b>NCs</b>.</li> <li>In endpoint mode (external arbiter) with <b>PCIX_EP#</b> = 0: These signals are not used and left as <b>NCs</b>.</li> </ul>	

Table 52. Terminations: Pull-up/Pull-down (Sheet 2 of 7)

Signal	Recommendations	Comments
P_REQ[3:1]#	<ul style="list-style-type: none"> <li>When PCI Express interface only: P_REQ[3:1]# is left as a NC.</li> <li>In Central Resource mode (internal arbiter) with PCIX_EP# = 1: These are three input request signals to the internal arbiter. Unused signals are left as NCs.</li> <li>In endpoint mode (external arbiter) with PCIX_EP# = 0: These signals are not used and left as NCs</li> </ul>	
P_REQ64#	<ul style="list-style-type: none"> <li>When only PCI Express interface is active these signals are internally pulled-up and left as a NC.</li> <li>When the PCIX_PULLUP# is enabled (pulled to 0), these signals are internally pulled-up.</li> <li>When the device is PCI endpoint then the width of the bus is indicated by the state of REQ64# at the rising edge of RST#.</li> </ul>	
P_ACK64#, P_PAR, P_SERR#, P_PERR#, P_INT[D:A]#	<ul style="list-style-type: none"> <li>When only PCI Express interface is active these signals are internally pulled-up and left as a NC.</li> <li>When the PCIX_PULLUP# is enabled (pulled to 0), these signals are internally pulled-up.</li> </ul>	
P_FRAME#, P_IRDY#, P_TRDY#, P_STOP#, P_DEVSEL#	<ul style="list-style-type: none"> <li>When only PCI Express interface is active these signals are internally pulled-up and left as a NC.</li> <li>When the PCIX_PULLUP# is enabled (pulled to 0), these signals are internally pulled-up.</li> <li>When the device is PCI endpoint PCIX_EP# = 0 state of these signals are used as for PCI-X initialization pattern at the rising edge of RST#.</li> </ul>	Refer to the <i>PCI-X Specification 1.0b</i> for more information on the PCI-X Initialization pattern.
P_M66EN	<p>PCI Express: P_M66EN has an internal pull-up and left as a NC.</p> <p>PCI-X Central Resource mode (PCIX_EP# = 1): Connect to the M66EN signal on the board</p> <ul style="list-style-type: none"> <li>CLK_SRC_PCIE# = 0 (using the P_CLK[3:0] clock outputs): pull-up signal with 8.2 KW.</li> <li>CLK_SRC_PCIE# = 0 (using the P_CLK[3:0] clock outputs but limiting PCI clock frequency): Refer to Table 25, "PCI Express Layout for Clock Routing" on page 55.</li> <li>CLK_SRC_PCIE# = 1: (P_CLKIN primary clock source): refer to Section 6.1.1, "PCI/PCI-X Frequency Selection" on page 57 for PCI frequency selection information.</li> </ul> <p>PCI Endpoint mode (PCIX_EP# = 0): Connect to the M66EN signal on the board and refer to Section 6.1.1, "PCI/PCI-X Frequency Selection" on page 57 for PCI frequency selection information.</p>	
P_IDSEL	<ul style="list-style-type: none"> <li>PCI Express: P_IDSEL has an internal pull-up and left as a NC.</li> <li>Central Resource mode PCIX_EP# = 1: pull down with 1K resistor.</li> <li>PCI Endpoint mode PCIX_EP# = 0: connect to AD lines Section 6.1.1, "PCI/PCI-X Frequency Selection" on page 57</li> </ul>	



Table 52. Terminations: Pull-up/Pull-down (Sheet 3 of 7)

Signal	Recommendations	Comments
P_CLKIN	<p>For PCI Express only this signal is connected to <b>GND</b>.</p> <p>PCI Central Resource mode (<b>PCIX_EP# = 1</b>):</p> <ul style="list-style-type: none"> <li><b>CLK_SRC_PCIE# = 0</b>: Connect to <b>P_CLKOUT</b> through a 26 ohm +/- 1% resistor. Refer to the PCI-X chapter for length match details.</li> <li><b>CLK_SRC_PCIE# = 1</b> (<b>P_CLKIN</b> is the primary clock source): connect to the system PCI clock</li> </ul> <p>PCI Endpoint mode (<b>PCIX_EP# = 0</b>): connect to the system PCI clock.</p>	<p><b>Note:</b></p> <ul style="list-style-type: none"> <li><b>REFCLKP, REFCLKN</b> must have 100 MHz clock to generate the <b>P_CLKO[3:0]</b> outputs.</li> <li>When the <b>P_CLKIN</b> is the primary clock source (<b>CLK_SRC_PCIE# = 1</b>), the PCI Clock outputs are unavailable and are not used as a clock source for any device.</li> </ul>
P_CLKOUT	<p>For PCI Express only these signals are unconnected.</p> <p>PCI Central Resource mode (<b>PCIX_EP# = 1</b>):</p> <ul style="list-style-type: none"> <li><b>CLK_SRC_PCIE# = 0</b> (using the <b>P_CLKO[3:0]</b> outputs): Connect to the <b>P_CLKIN</b> through a 26 ohm +/- 1% resistor (see PCI-X chapter of the design guide for length match details)</li> <li><b>CLK_SRC_PCIE# = 1</b>: this signal is left unconnected.</li> </ul> <p>PCI Endpoint mode (<b>PCIX_EP# = 0</b>): this signal is left unconnected.</p>	<p><b>Note:</b> <b>REFCLKP, REFCLKN</b> must have 100 MHz clock to generate the <b>P_CLKO[3:0]</b> outputs.</p>
P_CLKO[3:0]	<p>Connect to PCI device <b>P_CLK</b> inputs through a 28 ohm +/- 1% series resistor for each slot and a 26 ohm +/- 1% for each embedded device. Refer to the PCI-X chapter of the design guide for length match details.</p> <p>Any unused <b>P_CLKOs</b> are left unconnected.</p>	<p><b>Note:</b> <b>REFCLKP, REFCLKN</b> must be have 100 MHz clock to generate the <b>P_CLKO[3:0]</b> outputs.</p>
P_PCIXCAP	<p>When PCI Express only:</p> <ul style="list-style-type: none"> <li><b>GND</b> this pin.</li> </ul> <p>When PCI Central Resource mode is enabled <b>PCIX_EP# = 1</b>:</p> <ul style="list-style-type: none"> <li><b>CLK_SRC_PCIE# = 0</b> using <b>P_CLK[3:0]</b> outputs: connect signal with 3.3 K<math>\Omega</math> pull-up to 3.3 V.</li> <li><b>CLK_SRC_PCIE# = 0</b> (using the <b>P_CLK[3:0]</b> clock outputs but limiting clock frequency): Refer to <a href="#">Table 25, "PCI Express Layout for Clock Routing" on page 55</a>.</li> <li><b>CLK_SRC_PCIE# = 1</b> (<b>P_CLKIN</b> primary clock source): refer to <a href="#">Section 6.1.1, "PCI/PCI-X Frequency Selection" on page 57</a> for PCI frequency selection information.</li> </ul> <p>When PCI Endpoint mode: <b>PCIX_EP# = 0</b>:</p> <ul style="list-style-type: none"> <li>Pull-up signal with 8.2K resistor.</li> </ul> <p>Refer to <a href="#">Section 6.1.1, "PCI/PCI-X Frequency Selection" on page 57</a> for PCI frequency selection information and for termination for the PCIXCAP pin on the edge connector.</p>	<p>Refer to PCI-X Specification 1.0b and <a href="#">Section 6.1.1, "PCI/PCI-X Frequency Selection" on page 57</a></p> <p><b>Note:</b> This signal has been defeatured. Refer to non-core Erratum 18 in the Intel® 81341 and Intel® 81342 I/O Processors <a href="#">Specification Update</a> for more information.</p>
P_BMI	<ul style="list-style-type: none"> <li>When PCI Express only: this signal is left as a no connect.</li> <li>For PCI-X: no connect when not used.</li> </ul>	
P_CAL[0], P_CAL[2]	<ul style="list-style-type: none"> <li>When PCI-X interface is used: This pin is connected to a separate 22.1 <math>\Omega</math> 1% resistor to <b>GND</b>. See <a href="#">Section 11.5</a> for more information.</li> <li>When PCI-X interface is not used: These pins are left as <b>NCs</b></li> </ul>	<p><b>PCI Calibration:</b> is connected to an external calibration resistor to dynamically adjust their slew rate and drive strength to compensate for voltage and temperature variations.</p>

Table 52. Terminations: Pull-up/Pull-down (Sheet 4 of 7)

Signal	Recommendations	Comments
P_CAL[1]	<ul style="list-style-type: none"> <li>When PCI-X is used: This pin is connected to a separate 121Ω 1% resistor to <b>GND</b>. See <a href="#">Section 11.5</a> for more information.</li> <li>When PCI-X interface is not used: These pins are left as <b>NCs</b></li> </ul>	<b>PCI Calibration:</b> is connected to an external calibration resistor such that the output drivers reference the resistor to dynamically adjust the ODT resistance to compensate for voltage and temperature variations.
PE_CALP PE_CALN	Connect <b>PE_CALP</b> ball through 1.4K 1% resistor to the <b>PE_CALN</b> ball. Refer to <a href="#">Figure 57</a> . Note: this is required even when the PCI-Express interface is not used.	
M_CAL[0]	Connect to 24.9 ohm 1% resistor ground. Refer to <a href="#">Figure 58</a>	
M_CAL[1]	Connect to 301 ohm 1% resistor to ground. Refer to <a href="#">Figure 58</a> .	
ODT[1:0]	<b>NC</b> when not used When On-Die DDR2 termination used connect to the ODT inputs on the DDR2 SRAM.	Follow the same layout guidelines for CS# signals.
M_CK[2:0], M_CK[2:0]#	<ul style="list-style-type: none"> <li>Unused M_CK/M_CK#s are left unconnected</li> <li>When used with Registered DIMMs: connect <b>M_CK[0]/M_CK[0]#</b> pair, <b>M_CK[1]/M_CK[1]#</b>, <b>M_CK[2]/M_CK[2]#</b> are left unconnected</li> <li>When used with unbuffered DIMMs: Connect M_CK[2:0]/M_CK[2:0]# to the DDR2 CK/CK# inputs.</li> </ul>	These DDR2 clock signals are used to provide the three differential clock pairs. Refer to <a href="#">Section 6.1.1, "PCI/PCI-X Frequency Selection"</a> on <a href="#">page 57</a> for more details.
M_RST#	<b>NC</b> when not used	This Reset signal asynchronously forces all registered outputs LOW on the registered DDR2 DIMM
MA[13:0]	Unused address lines are left unconnected. When used refer to <a href="#">Section 4.2.1</a> for DDR2 533 termination recommendations.	DDR2 address signals
RAS#, CAS#, WE#, CS[1:0]#, CKE[1:0]	Unused lines are left unconnected. When used refer to <a href="#">Section 4.2.1</a> for DDR2 533 termination recommendations.	DDR2 control signals
DQ[63:0], DM[8:0], CB[7:0], DQS[8:0], DQS[8:0]#	Unused pins are left unconnected. When used refer to <a href="#">Section 4.2.1</a> for DDR2 533 termination recommendations.	Source Synchronous signals
M_VREF	Connect to the memory VREF voltage 0.9V refer to <a href="#">Section 4.2.1</a> for DDR2 termination recommendations.	
A[24:0], POE#, PB_RSTOUT#	Unused pins are left unconnected. When used refer to <a href="#">Section 7.2</a> for PBI bus connection recommendations.	
D[15:0], PCE[1:0]#, PWE#	These are used for reset straps refer to the Reset Strap <a href="#">Table 53</a> . Also refer to <a href="#">Section 7.2</a> for PBI bus connection recommendations.	
HS_ENUM#	Left unconnected when hot swap not used.	
HS_LSTAT	When Compact PCI Hot Swap is not supported, this signal must be tied to <b>GND</b> .	Hot Swap Latch Status: An input indicating the state of the ejector switch. 0 = Indicates the ejector switch is closed. 1 = Indicates the ejector switch is open. 1 = 8.2K pull-up to <b>VCC</b> 0 = connect to <b>GND</b> .





Table 52. Terminations: Pull-up/Pull-down (Sheet 5 of 7)

Signal	Recommendations	Comments
HS_LED_OUT	Connect to Hot Swap blue LED. When Compact PCI Hot Swap is not supported this signal is left unconnected.	
HS_FREQ[1:0] / CR_FREQ[1:0]	See comments	Hot Swap Frequency: While in Hot Swap mode, (these are only valid when <b>PCIX_EP#</b> = 0 and <b>HS_SM#</b> = 0). 00 = 133MHz PCI-X 01 = 100MHz PCI-X 10 = 66MHz PCI-X 11 = 33 or 66MHz. PCI (frequency depends on <b>P_M66EN</b> ) Central Resource Frequency: While in Central Resource mode, (these are only valid when <b>PCIX_EP#</b> = 1). 00 = 133 MHz 01 = 100 MHz 10 = 66 MHz 11 = 33 MHz <b>Note:</b> 1 = internal pull-up 0 = connect to <b>GND</b>
P_INT[D:A]# / XINT[3:0]# / GPIO[11:8]	<ul style="list-style-type: none"> <li>when <b>PCIX_EP#</b>=0: no termination is required</li> <li>when <b>PCIX_EP#</b>=1: 8.2 K pull-up required</li> <li>when used as GPIOs, these signals need 8.2 K pull-up</li> </ul>	When <b>INTERFACE_SEL_PCIX#</b> = "0": PCI Interrupt: These outputs are level sensitive. When <b>INTERFACE_SEL_PCIX#</b> = "1": External Interrupt: requests are used by external devices to request interrupt service. These pins are level-detect inputs and are internally synchronized. These pins go to the <b>XINT[3:0]#</b> inputs of the Interrupt Controller.
HPI#, NMIO#, NMI1#, XINT[7:4]#	8.2 K pull-ups	
GPIO[7:0] / XINT[15:8]# / CHAPOUT	8.2 K pull-ups	General Purpose I/O (default mode). <b>External Interrupt:</b> These pins are level-detects and are internally synchronized. CHAPOUT: <b>GPIO[7]</b> When enabled it overrides the normal <b>GPIO[7]</b> function.
SCL0, SDA0, SCL1, SDA1, SCL2, SDA2	When used external pull-up to <b>VCC</b> is required. Refer to the I <sup>2</sup> C specification for information on calculating the pull-up. 2K pull-up when unused.	The pull-up value is dependent on the bus loading. Refer to the I <sup>2</sup> C specification <a href="http://www.semiconductors.philips.com/acrobat/literature/9398/39340011.pdf">http://www.semiconductors.philips.com/acrobat/literature/9398/39340011.pdf</a>
SMBCLK	For PCI Express adapter cards: <ul style="list-style-type: none"> <li>When the SMBus is used, there is an isolation device such as the LTC4301 between this signal and PE_SMCK on the PCI Express connector.</li> </ul> For PCI Express motherboard applications: <ul style="list-style-type: none"> <li>When SMBus is used a pull-up is required (value is dependent on the loading).</li> <li>When SMBus is unused, a 8.2K pull-up is required.</li> </ul>	LTC4301 is a hotswappable 2-wire bus buffer that allows card insertion without corruption of the data and clock buses. Refer to the Linear Technology website <a href="http://www.linear.com/pc/downloadDocument.do?navId=H0,C1,C1007,C1070,P2460,D3045">http://www.linear.com/pc/downloadDocument.do?navId=H0,C1,C1007,C1070,P2460,D3045</a> Refer also to the <a href="http://www.smbus.org">http://www.smbus.org</a> for the latest specification.

Table 52. Terminations: Pull-up/Pull-down (Sheet 6 of 7)

Signal	Recommendations	Comments
<b>SMBDAT</b>	<p>For PCI Express adapter cards:</p> <ul style="list-style-type: none"> <li>When the SMBus is used, there is an isolation device such as the LTC4301 between this signal and <b>PE_SMDAT</b> on the PCI Express connector.</li> </ul> <p>For PCI Express motherboard applications:</p> <ul style="list-style-type: none"> <li>When SMBus is used a pull-up is required (value is dependent on the loading).</li> <li>When SMBus is unused, a 8.2K pull-up is required.</li> </ul>	<p>LTC4301 is a hotswappable 2-wire bus buffer that allows card insertion without corruption of the data and clock buses. Refer to the Linear Technology website <a href="http://www.linear.com/pc/downloadDocument.do?navId=H0,C1,C1007,C1070,P2460,D3045">http://www.linear.com/pc/downloadDocument.do?navId=H0,C1,C1007,C1070,P2460,D3045</a></p> <p>Refer also to the <a href="http://www.smbus.org">http://www.smbus.org</a> for the latest specification.</p>
<b>U0_RXD, U1_RXD</b>	When unused connect to <b>GND</b>	UART 0 is dedicated as the debug port for the transport FW as implemented on Intel development boards. UART 1 is a general purpose port.
<b>U0_TXD, U0_RTS#, U1_TXD, U1_RTS#</b>	Left unconnected when unused.	UART 0 is dedicated as the debug port for the transport FW as implemented on Intel development boards. UART 1 is a general purpose port.
<b>U0_CTS#, U1_CTS#</b>	When unused 8.2K pull-up	UART 0 is dedicated as the debug port for the transport FW as implemented on Intel development boards. UART 1 is a general purpose port.
<b>TCK</b>	<b>GND</b> when unused. 8.2K pull-up when used. Refer to the JTAG chapter.	Test Clock: provides clock input for IEEE 1149.1 Boundary Scan Testing (JTAG).
<b>TDI</b>	<b>NC</b> when unused has weak pull-up. 8.20K pull-up when used. Refer to the JTAG chapter.	Test Data Input: is the JTAG serial input pin.
<b>TDO</b>	<b>NC</b> when unused	Test Data Output: is the serial output pin for the JTAG feature.
<b>TRST#</b>	<b>GND</b> when unused.	Test Reset: This pin has a weak internal pull-up.
<b>TMS</b>	<b>NC</b> when unused has weak pull-up. 8.2K pull-up when used. Refer to the JTAG chapter.	Test Mode Select: This pin has a weak internal pull-up.
<b>WARM_RST#</b>	<ul style="list-style-type: none"> <li>When unused: 1K pull-up.</li> <li>When used: This pin is only used when the sticky bit functionality is required. In this scenario, the <b>WARM_RST#</b> pin must be tied to the system reset <b>PCI_RST#</b> signal while the <b>P_RST#</b> pin is tied to the system power good signal.</li> </ul>	<p>Warm Reset is the same as a cold reset, except sticky configuration bits are not reset.</p> <p><b>Note:</b></p> <ul style="list-style-type: none"> <li>When the PCI Express interface is used as an endpoint, the PCI Express in-band Hot Reset Mechanism is also used to provide the sticky bit functionality.</li> <li>On the customer reference board, <b>WARM_RST#</b> is tied to the <b>SRST_N</b> to provide a JTAG debugger reset.</li> <li>Driving <b>WARM_RST#</b> using any other methods than suggested above results in unpredictable behavior of the device.</li> </ul>
<b>NC</b>	No Connect: pins have no usable function and must not be connected to any signal, power or ground.	

**Table 52. Terminations: Pull-up/Pull-down (Sheet 7 of 7)**

Signal	Recommendations	Comments
<b>THERMDA</b>	Connect to the anode of the thermal diode. <b>NC</b> when unused	
<b>THERMDC</b>	Connect to the cathode of the thermal diode. <b>NC</b> when unused	
<b>PUR1</b>	This pin must be pulled up to VCC3P3 with an external 4.7K $\Omega$ 5%, $\frac{1}{16}$ W resistor for proper operation.	

## 11.3 Reset Straps

The following table provides a list of reset straps which are multiplexed on the Peripheral Address Bus A[24:0]. These pins are latched on the rising edge of **P\_RST#**. All reset strap signals are internally pulled to logic 1 by default. An external 4.7KΩ 5% pull-down resistor is required to force a logic 0 on these pins.

**Table 53. Reset Straps (Sheet 1 of 3)**

Signal	Recommendations	Comments
<b>BOOT_WIDTH_8#</b>	8 bits wide, 0 = 4.7K ohms resistor pull down 16 bits wide (Default mode internal pull-up)	<b>Note:</b> Muxed onto signal <b>A[0]</b> .
<b>CFG_CYCLE_EN#</b>	Configuration Cycles enabled, 0 = 4.7K ohms resistor pull down Configuration Retry enabled (Default mode internal pull-up)	<b>Note:</b> Muxed onto signal <b>A[1]</b>
<b>HOLD_X0_IN_RST#</b>	0 - Hold Scale in reset, 0 = 4.7 K ohms resistor pull down 1 - Do not hold in reset (Default mode internal pull-up)	<b>Note:</b> Muxed onto signal <b>A[2]</b>
<b>HOLD_X1_IN_RST#</b>	Hold in reset, 0 = 4.7K ohms resistor pull down Do not hold in reset (Default mode internal pull-up).	<b>Note:</b> Muxed onto signal <b>A[3]</b>
<b>MEM_FREQ[1:0]</b>	10 = 533MHz 11 = 400MHz (Default mode). 0 = 4.7K ohms resistor pull down 1 = internal pull-up.	<b>MEM_FREQ[1:0]</b> muxed onto signal <b>A[5]</b> and <b>A[4]</b> respectively 0 = 4.7K ohms resistor pull down 1 = internal pull-up.
<b>EXT_ARB#</b>	0 =) External arbiter, 0 = 4.7K ohms resistor pull down 1 =) Internal arbiter (Default mode internal pull-up)	Muxed onto signal <b>A[6]</b>
<b>INTERFACE_SEL_PCIX#</b>	ATU-X is function 0 (4.7 KΩ pull-down resistor) and ATUE is function 5. ATU-E is function 0 and ATUX is function 5. Refer to comments.	Interface Select PCI-X: determines which ATU is function 0. 0 = 4.7K ohms resistor pull down 1 = internal pull up. <b>Note:</b> Muxed onto signal <b>A[10]</b>
<b>PCIX_EP#</b>	Refer to comments.	PCI-X End Point: 0 = Endpoint 1 = Central Resource (Default mode) <b>Note:</b> muxed onto signal <b>A[11]</b> 0 = 4.7K ohms resistor pull down 1 = internal pull up.
<b>PCIE_RC#</b>	Refer to comments.	PCI-E Root Complex: 0 = Root Complex 1 = Endpoint (Default mode) muxed onto signal 0 = 4.7K ohms resistor pull down 1 = internal pull up.

**Table 53. Reset Straps (Sheet 2 of 3)**

Signal	Recommendations	Comments
<b>SMB_A5, SMB_A3, SMB_A2, SMB_A1</b>	Refer to comments.	<p>SM Bus Address: maps to address bits 5,3,2, and 1 where bits 7- 0 represent the address the SMBus slave port responds to when access is attempted. address bit is low address bit is high (Default mode)</p> <p><b>Note:</b> <b>SMB_A5</b> muxed onto signal <b>A[16]</b>  <b>Note:</b> <b>SMB_A3</b> muxed onto signal <b>A[15]</b>  <b>Note:</b> <b>SMB_A2</b> muxed onto signal <b>A[14]</b>  <b>Note:</b> <b>SMB_A1</b> muxed onto signal <b>A[13]</b></p> <p>0 = 4.7K ohms resistor pull down 1 = internal pull up.</p>
<b>PCIX_PULLUP#</b>	When pulled-low enables the following signal pull-ups: <b>P_AD[63:32]</b> , <b>P_C/BE[7:4]#</b> , <b>P_PAR64</b> , <b>P_REQ64#</b> , <b>P_ACK64#</b> , <b>P_FRAME#</b> , <b>P_IRDY#</b> , <b>P_TRDY#</b> , <b>P_STOP#</b> , <b>P_DEVSEL#</b> , <b>P_SERR#</b> , <b>P_PERR#</b> , <b>P_INT[D:A]#</b>	<p>PCI-X Pull Up: 0 = enable PCI pull up resistors 1 = disable PCI pull up resistors (Default mode)</p> <p><b>Note:</b> Muxed onto signal <b>A[17]</b> 0 = 4.7K ohms resistor pull down 1 = internal pull up.</p>
<b>PCIX_32BIT#</b>	When 32 PCI-X bus enabled the following signals have internal pull-ups: <b>P_AD[63:32]</b> , <b>P_C/BE[7:4]#</b> and <b>P_PAR64</b> and left as NCs.	<p>32-Bit PCI-X Bus: 0 = 32 bit wide PCI-X bus. 1 = 64 bit wide PCI-X bus. (Default mode)</p> <p><b>Note:</b> Muxed onto signal <b>A[18]</b> 0 = 4.7K ohms resistor pull down 1 = internal pull up.</p>
<b>PCIXM1_100#</b>	Refer to comments.	<p>PCI-X Mode 1 100MHz Enable: 0 = limit PCI-X mode 1 to 100MHz 1 = 133MHz enabled (Default mode)</p> <p><b>Note:</b> Muxed onto signal <b>A[19]</b> 0 = 4.7K ohms resistor pull down 1 = internal pull up.</p>
<b>HS_SM#</b>	Refer to comments	<p>Hot Swap Startup Mode: 0 = Hot Swap mode enabled 1 = Hot Swap mode disabled (Default mode)</p> <p><b>Note:</b> Muxed onto signal <b>A[21]</b> 0 = 4.7K ohms resistor pull down 1 = internal pull up.</p>
<b>FW_TIMER_OFF#</b>	Refer to comments.	<p>Firmware Timer Off: 0 = firmware timer disabled 1 = firmware timer enabled (Default mode)</p> <p><b>Note:</b> Muxed onto signal <b>A[22]</b> 0 = 4.7K ohms resistor pull down 1 = internal pull up.</p>



Table 53. Reset Straps (Sheet 3 of 3)

Signal	Recommendations	Comments
CLK_SRC_PCIE#	Refer to comments.	Clock Source PCI-E: selects the PCI Express Refclk pair as the input clock to the PLLs that control most internal logic. 0 = source clock is REFCLKP/REFCLKN 1 = source clock comes from the active PCI interface (Default mode) <b>Note:</b> Muxed onto signal PWE# 0 = 4.7K ohms resistor pull down 1 = internal pull up.
LK_DN_RST_BYPASS#	Use for PCI Express mode	Link Down Reset Bypass: Disables the full chip reset that are normally be caused by a Link Down or hot reset. 0 = Do not reset on Link Down 1 = Reset on Link Down (default mode) Muxed onto signal A[24]
PCE[1:0]#	Pull up both these signals with 8.2K resistor	



## 11.4 Analog Filters

This section describes filters needed for the PLL circuitry. [Figure 54](#) lists the PLLs that are required for this part.

**Table 54. Required PLLs**

Interface	Filtered Voltage	VCC PLL Balls	VSS PLL Balls	Layout Guideline Table
PCI-X	1.2V	<b>WARM_RST#</b>	VSSPLL	<a href="#">Table 55</a>
Core Digital Logic	1.2V	VCC1P2PLLD	VSSPLLD	<a href="#">Table 55</a>
Intel XScale <sup>®</sup> microarchitecture and XSI bus logic	3.3V	VCC3P3LLX	VSSPLLX	<a href="#">Table 56</a>

### 11.4.1 VCC1P2PLL, VCC1P2PLLD Filter Requirements

The lowpass filter, as shown in Figure 54 reduces noise induced clock jitter and its effects on timing relationships in system designs. The Table 55 filter circuit is recommended for each of the PLL pairs: VCC1P2PLL-VSSPLL and VCC1P2PLLD - VSSPLLD.

The filter has the following characteristics:

- The filter components must be able to handle a DC current of 30mA.
- < 0.2dB gain in pass band and < 0.5dB attenuation in pass band < 1Hz. The passband is DC through 1Hz.
- > 34dB attenuation from 1MHz to 66MHz
- > 28dB attenuation from 66MHz to core frequency

**Note:** When the PCI-X interface is not used the VCC1P2PLL and VSSPLL pins are grounded.

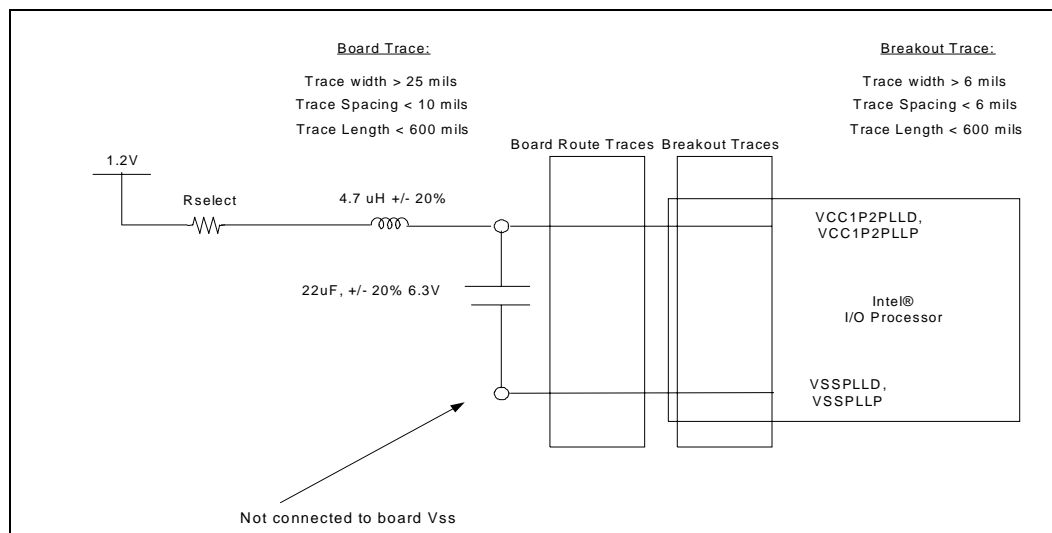
**Table 55. VCC1P2PLL, VCC1P2PLLD Layout Guideline**

Parameter	Specification
Reference Plane	<ul style="list-style-type: none"> <li>• Ground</li> <li>• VCC1P2PLL, VCC1P2PLLD traces must be ground referenced (no <b>VCC</b> references)</li> </ul>
Inductor	<ul style="list-style-type: none"> <li>• 4.7 <math>\mu</math>H +/- 25%</li> <li>• L must be magnetically shielded</li> <li>• ESR: max &lt; 0.3 <math>\Omega</math></li> <li>• rated at 45 mA</li> </ul>
Capacitor	<ul style="list-style-type: none"> <li>• 33 <math>\mu</math>F +/- 20% 6.3V (Capacitor)</li> <li>• ESR: max &lt; 0.3 <math>\Omega</math></li> <li>• ESL &lt; 2.5 nH</li> <li>• Place 33 <math>\mu</math>F capacitor as close as possible to package pin.</li> </ul>
Resistor	<ul style="list-style-type: none"> <li>• Rselect: choose resistor such that both of the following conditions are met: <ol style="list-style-type: none"> <li>1. 1.2V plane to the top end of the capacitor is &gt; 0.35 <math>\Omega</math> (ινχλυδινγ βοαρδ ανδ χομπονεντ ρεσιστανχε)</li> <li>2. 1.2V plane to VCC1P2PLL &lt; 1.5 <math>\Omega</math> <ul style="list-style-type: none"> <li>• 1/16 W 6.3 V</li> <li>• resistor must be placed between VCC1P2 and L.</li> </ul> </li> </ol> </li> </ul> <p><b>Note:</b> When trace and component resistance is large enough the discrete resistor is not required</p>
Breakout Trace	<ul style="list-style-type: none"> <li>• Trace Width &gt; 6 mils</li> <li>• Trace Spacing &lt; 6 mils</li> <li>• Trace Length &lt; 600 mils</li> </ul>
Board Trace	<ul style="list-style-type: none"> <li>• Trace Width &gt; 25 mils</li> <li>• Trace Spacing &lt; 10 mils</li> <li>• Trace Length &lt; 600 mils</li> </ul>
Trace Spacing	<ul style="list-style-type: none"> <li>• <math>\geq</math> 30 mils from any other signals.</li> </ul>
Trace Length maximum	1.2"
Routing Guideline 1	Route VCC1P2PLLD and VSSPLLD, VCC1P2PLL and VSSPLL as differential traces.
Routing Guideline 2	The nodes connecting VCC1P2PLLD and the capacitor, VCC1P2PLL and the capacitor must be as short as possible.
Routing Guideline 3	The 1.2 V supply regulator used for the PLL filter must have less than +/- 3% tolerance





**Figure 54. VCC1P2PLLD, VCC1P2PLL Lowpass Filter Configuration**



## 11.4.2 VCC3P3PLLX PLL Requirements

To reduce clock skew, a PLL is implemented for Intel XScale® microarchitecture and core logic. The balls associated with this PLL are VCC3P3PLLX and VSSPLLX. The lowpass filter, as shown in Figure 55, reduces noise induced clock jitter and its effects on timing relationships in system designs. The node connecting VCC3P3PLLX and the capacitor must be as short as possible.

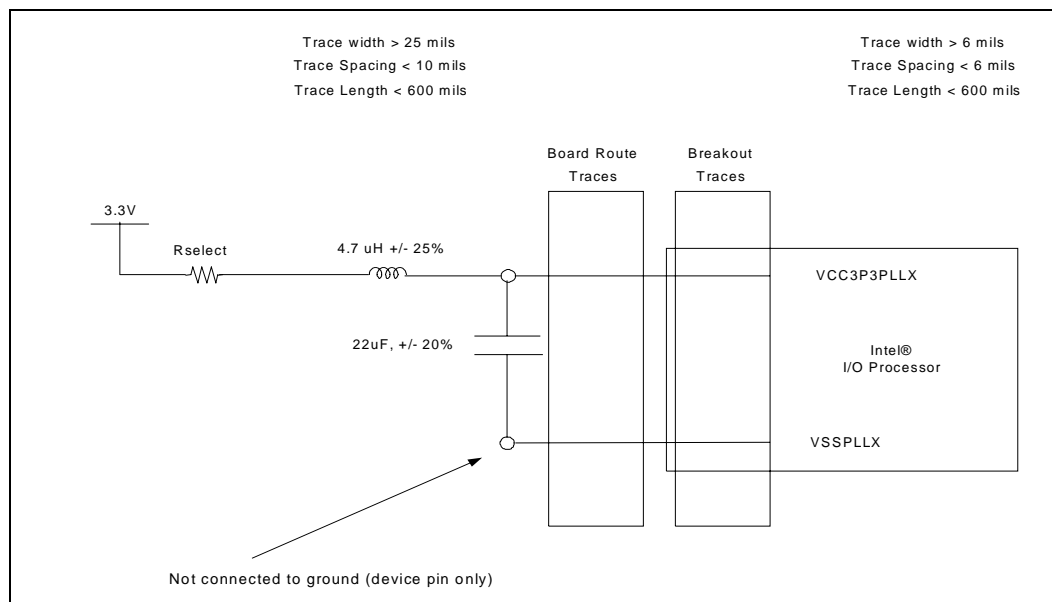
The filter has the following characteristics:

- The filter components must be able to handle a DC current of 30mA.
- < 0.2dB gain in pass band and < 0.5dB attenuation in pass band < 1Hz. Passband is DC through 1Hz.
- > 34dB attenuation from 1MHz to 66MHz
- > 28dB attenuation from 66MHz to core frequency

The following notes list the layout guidelines for this filter:

**Table 56. VCC3P3PLL Layout Guideline**

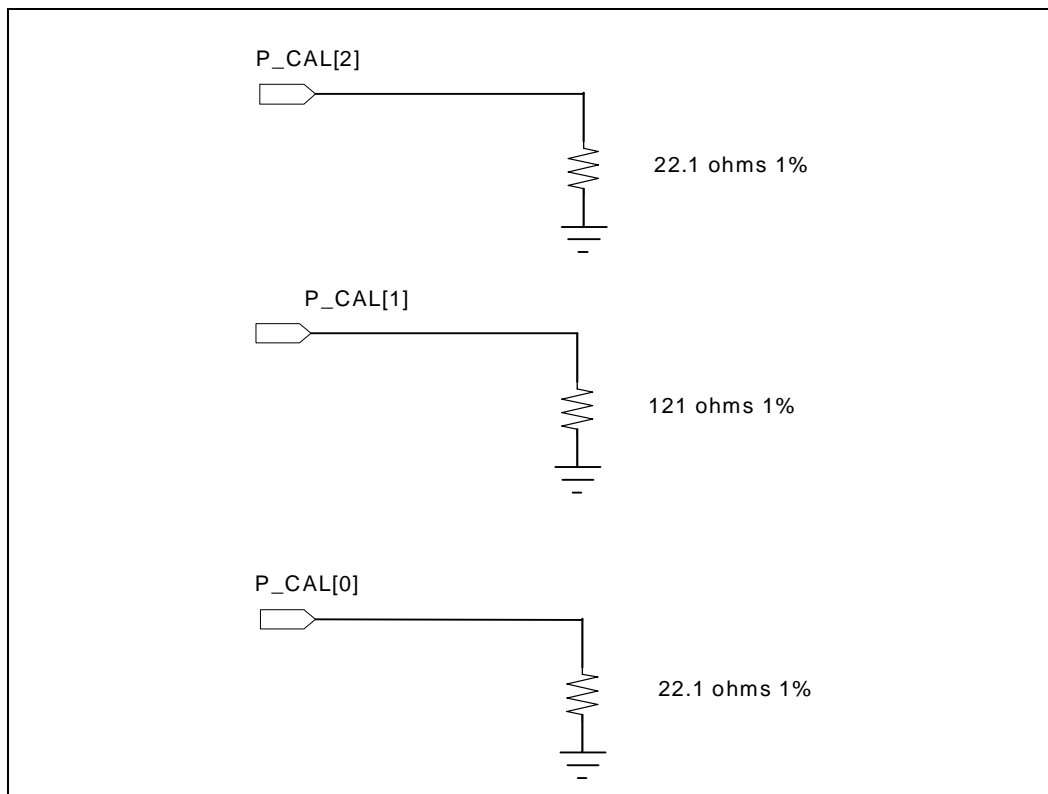
Parameter	Specification
Reference Plane	<ul style="list-style-type: none"> <li>• Ground referenced</li> <li>• VCC3P3PLL and VSSPLLX traces must be ground referenced (no <b>VCC</b> references)</li> </ul>
Inductor	<ul style="list-style-type: none"> <li>• 4.7 <math>\mu</math>H</li> <li>• L must be magnetically shielded</li> <li>• ESR: max &lt; 0.4 <math>\Omega</math></li> <li>• rated at 45 mA</li> <li>• An example of this inductor is TDK part number MLZ2012E4R7P.</li> </ul>
Capacitor	<ul style="list-style-type: none"> <li>• 33 <math>\mu</math>F (Capacitor)</li> <li>• ESR: max &lt; 0.4 <math>\Omega</math></li> <li>• ESL &lt; 3.0 nH</li> <li>• Place 33 <math>\mu</math>F capacitor as close as possible to package pin.</li> </ul>
Resistor	<ul style="list-style-type: none"> <li>• Rselect: choose resistor such that both of the following conditions are met: <ol style="list-style-type: none"> <li>1. 3.3V plane to the top end of the capacitor is &gt; 0.35 <math>\Omega</math></li> <li>2. 3.3V plane to VCC3P3PLL &lt; 1.5 <math>\Omega</math></li> </ol> </li> <li>• resistor ratings: 1/16 W 6.3 V</li> <li>• resistor must be placed between VCC3P3 and L.</li> <li>• Note: when trace and component resistance is large enough the discrete resistor is not required</li> </ul>
Breakout Trace	<ul style="list-style-type: none"> <li>• Trace Width &gt; 6 mils</li> <li>• Trace Spacing &lt; 6 mils</li> <li>• Trace Length &lt; 600 mils</li> </ul>
Board Trace	<ul style="list-style-type: none"> <li>• Trace Width &gt; 25 mils</li> <li>• Trace Spacing &lt; 10 mils</li> <li>• Trace Length &lt; 600 mils</li> </ul>
Trace Spacing	<ul style="list-style-type: none"> <li>• <math>\geq</math> 30 mils from any other signals.</li> </ul>
Trace Length maximum	1.2"
Routing Guideline 1	Route VCC3P3PLLX and VSSPLLX as differential traces.
Routing Guideline 2	The nodes connecting VCC3P3PLL and the capacitor must be as short as possible.

**Figure 55. VCC3P3PLL Filter Configuration**

## 11.5 PCI Resistor Calibration

Figure 56 shows the termination required for the PCI calibration circuitry. **PCI** Calibration pins P\_CAL[1:0] are connected to an external calibration resistors. The PCI output drivers reference the resistor for dynamic adjustment of slew rate and drive strength to compensate for voltage and temperature variations.

**Figure 56. PCI Resistor Calibration**

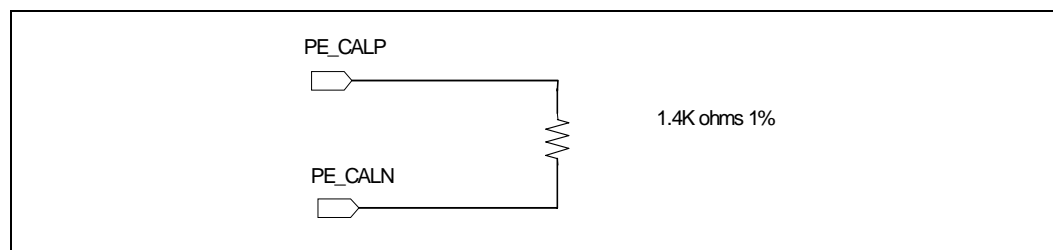




## 11.6 PCI Express Resistor Compensation

Figure 57 shows the termination required for the PCI Express RCOMP circuit.

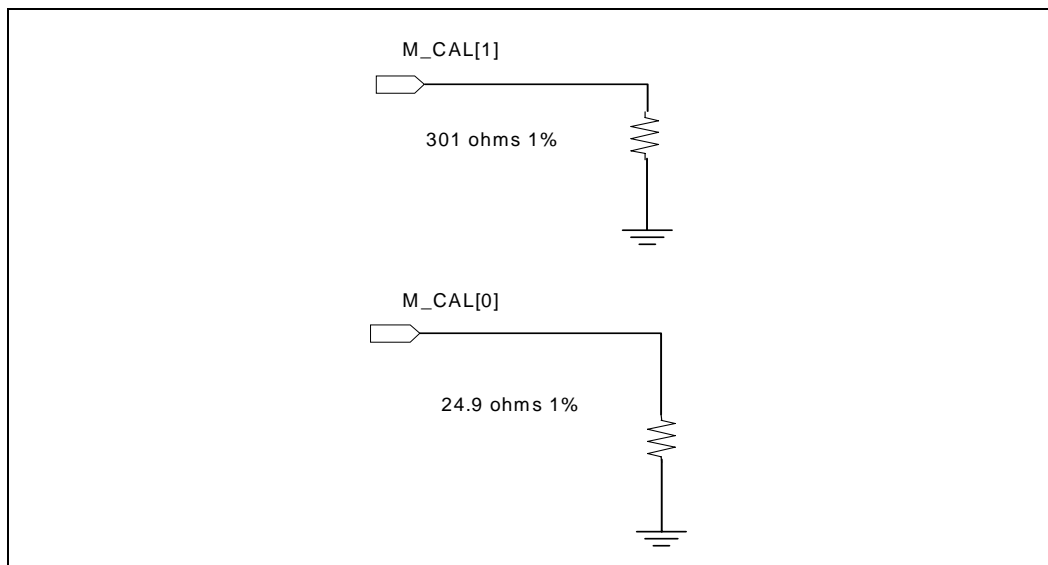
**Figure 57. PCI Express RCOMP**



## 11.7 Memory Calibration Circuitry

The [Figure 58](#) shows the memory calibration pins **M\_CAL[1]** and **M\_CAL[0]** connected to external calibration resistors to ground. The memory output drivers reference the resistor for dynamic adjustment of the drive strength to compensate for temperature and voltage variations.

**Figure 58. Memory Calibration Circuitry**





## 12.0 Layout Checklist

### 12.1 Intel® 81341 and 81342 I/O processors Layout Checklist

The [Table 57](#) provides a summary of layout guidelines for each of the Intel® 81341 and Intel® 81342 I/O Processors interfaces described in detail in the previous sections. The spacing and width specifications are based on the stackup provided in [Chapter 3.0](#).

**Table 57. Intel® 81341 and 81342 I/O processors Layout Checklist (Sheet 1 of 20)**

Checklist Items	Recommendations	Comments
DDR2 DIMM Synchronous (DQ/DQS/DM/CB)		Refer to Section 4.0
Reference Layer	Route over unbroken ground plane.	
Preferred Layer	Stripline	
Impedance	DQ Stripline Trace Impedance: 50 ohms +/- 15% motherboard and 60 ohms +/- 15% adapter card	
	DQS Differential Stripline Trace Impedance: 85 ohms +/- 15% and 100 ohms +/- 15% adapter card	Refer to stackup <a href="#">Chapter 3.0</a>
Spacing (edge to edge)	Spacing with-in the same group $\geq 12$ mils min. Spacing from other DQ groups $\geq 20$ mils min. Spacing of DQS to other signals $\geq 20$ min.	
Length Matching	Trace Length Matching: within DQS group: +/- 0.05" within one pair DQS +/-: +/- 0.0250"	
Length Matching: DQS with respect to clock	When total length: 0" < total length $\leq 6$ ", matching < +/- 0.5" 6" < total length $\leq 8$ ", matching < +/- 0.1"	
DQ Break out Exception	Microstrip spacing: 5 mils, width 5 mils Length 0" - 0.5"	
DQ Lead-in Length	0.5" to 8"	
DQS Break out Exception	Microstrip: spacing: 5 mils, width: 5 mils length 0" - 0.5"	
DQS Lead-in Length	0.5" to 8"	
Via counts	$\leq 2$ (for differential signals the number of vias on + and - signals must be the same)	
Routing Guideline	Route all data signals and their associated strobes on the same layer	
DQ and DQS ODT	150 ohms ODT enabled on memory for reads 75 ohms ODT enabled on IOP for writes	


**Table 57. Intel® 81341 and 81342 I/O processors Layout Checklist (Sheet 2 of 20)**

Checklist Items	Recommendations	Comments
<b>DDR2 DIMM Clock Routing (M_CK[2:0]/M_CK#[2:0])</b>		<b>Refer to Section 4.0</b>
Reference Plane	Route over unbroken ground plane or unbroken power plane.	
Preferred Topology	Microstrip	
Trace Impedance	Microstrip differential impedance: 85 ohms +/- 15% for motherboard and 100 +/- 15% for add-in card.	Refer to stackup <a href="#">Chapter 3.0</a>
Trace Spacing (edge to edge)	Between other groups $\geq 25$ mils	
Length Matching	Within M_CK/M_CK# differential clock + /- pair: +/- 0.0250" With respect to DQS group: when total length: $0 < \text{total length} \leq 6"$ , matching $\leq +/- 0.5"$ when total length: $6" < \text{total length} \leq 8"$ , matching $\leq +/- 0.1"$ With respect to address/command group +8"/-3" motherboard and +8"/-2" add-in card With respect to CS/CKE group +/-2" motherboard and +1/-3" add-in card	
Breakout:	Microstrip or Stripline spacing: 5 mils, width: 5 mils Length 0" - 0.5"	
Lead-in Length	Microstrip: 0.5" to 8"	
Via Count	Maximum of 1 via/layer change for M_CK/M_CK# differential clocks. Use the same number of vias for + and - lines.	
<b>DDR2 DIMM Address/Command Routing (MA[13:0], CS[1:0], CKE[1:0], ODT[1:0])</b>		<b>Refer to Section 4.0</b>
Reference Plane	Route over unbroken ground plane or unbroken voltage plane.	
Preferred Topology	Microstrip	
Trace Impedance	Impedance: 50 ohms +/- 15% for motherboard and 60 +/- 15% for add-in card.	
Trace Spacing (edge to edge)	Within the group $\geq 12$ mils Between other groups $\geq 20$ mils	
Breakout	Microstrip spacing: 5 mils, width: 5 mils Length 0" - 0.5"	
Lead-in Length	0.5" - 10"	
Length Matching: address/command group (except CS and CKE lines) with respect to clock (from controller to DIMM connector)	+8"/-3" maximum for motherboard and +8"/-2" maximum for add-in card	
Length Matching: CS and CKE lines with respect to clock (from controller to DIMM connector)	+/-2" maximum for motherboard and +1"/-3" maximum for add-in card	
Length to Parallel Termination	Microstrip: 0.15" - 0.5"	Place terminations in VTT island.
Parallel Termination: single	51.1 ohms +/- 1% to VTT. Refer to <a href="#">Figure 11</a>	
Parallel Termination: split	100 ohms +/- 1% to 1.8V and 100ohms +/-1% to ground. Refer to <a href="#">Figure 12</a>	



**Table 57. Intel® 81341 and 81342 I/O processors Layout Checklist (Sheet 3 of 20)**

Checklist Items	Recommendations	Comments
Via counts	2 vias or less	
<b>Embedded DDR2 Synchronous (DQ/DQS/DM/CB)</b>		<b>Refer to Section 4.0</b>
Reference Layer	Route over unbroken ground plane.	
Preferred Layer	Stripline	
Impedance	DQ Stripline Trace Impedance: 50 ohms +/- 15% DQS Differential Stripline Trace Impedance: 85 ohms +/- 15%	Refer to stackup Chapter 3.0
Spacing (edge to edge)	Spacing with-in the same group $\geq$ 12 mils min Spacing from other DQ groups $\geq$ 20 mils min. Spacing of DQS to other signals $\geq$ 20 min.	
Length Matching DQS pair and group	Trace Length Matching: within DQS group: +/- 0.05" within one pair DQS +/-: +/- 0.0250"	
Length Matching DQS group with respect clock	DQS length = clock length - 1" (tolerance +/- 0.1")	
DQ/DQS Break out Exception	Microstrip spacing: 5 mils, width 5 mils Length 0" - 0.5"	
Overall Trace Length	0.5" to 9.5"	
Via counts	$\leq$ 4 (for differential signals the number of vias on + and - signals must be the same)	
DQ and DQS ODT	150 ohms ODT enabled on IOP for reads 75 ohms ODT enabled on SDRAM	
Routing Guideline	Route all data signals and their associated strobes on the same layer	


**Table 57. Intel® 81341 and 81342 I/O processors Layout Checklist (Sheet 4 of 20)**

Checklist Items	Recommendations	Comments
<b>Embedded DDR2 Clock Routing (M_CK[2:0]/M_CK#[2:0])</b>		<b>Refer to Section 4.0</b>
Reference Plane	Route over unbroken ground plane	
Preferred Topology	Microstrip	
Trace Impedance	Microstrip differential impedance: 85 ohms +/- 15%	Refer to stackup Chapter 3.0
Trace Spacing (edge to edge)	Between other groups $\geq$ 25 mils	
Length Matching: With respect to DQ/DQS group (from controller to memory ball)	DQ/DQS length = clock length - 1"	
Length Matching: With respect to address/command group (except CS, CKE, ODT) from controller to memory ball	ADDR/CMD $\leq$ clock length + 2" ADDR/CMD $\geq$ clock length - 1"	
Length Matching with respect to CS/CKE group	For Daisy chain Topology: when CS/CKE length is $\leq$ 4": clock length + 1" when CS/CKE length is $>$ 4": clock length + 3" For balanced segment topology: when CS/CKE length is $\leq$ 2": clock length + 1" when CS/CKE length is $>$ 2": clock length +/- 0.5"	
Breakout:	Microstrip or Stripline spacing: 5 mils, width: 5 mils Length 0" - 0.5"	
Lead-in Length	Microstrip: 0.5" to 10.5"	
Routing Guideline	Maximum of 2 via/layer changes for M_CK/M_CK# clocks (same number of vias between + and - signals of the differential clock).	

**Table 57. Intel® 81341 and 81342 I/O processors Layout Checklist (Sheet 5 of 20)**

Checklist Items	Recommendations	Comments
<b>Embedded DDR2 Address/CMD Routing</b>		<b>Refer to Section 4.2.4.3</b>
Reference Plane	Route over unbroken ground plane preferred	
Preferred Topology	Microstrip lines	
Trace Spacing (edge to edge) <sup>1</sup>	5 mils acceptable between the pins and the breakout regions. $\geq 12$ mils within group $\geq 20$ mils from any other clock/DQ/DQS groups.	
Trace Impedance	50 ohms +/- 15% for a motherboard	
Trace Length: Overall length from signal Ball to SDRAM ball	1" min to 12" maximum	
Trace Length	TL Break out: $\leq 0.5"$	
	TL0: 0.5" to 9"	
	TL1: 0.2" to 0.75"	
	TL2: 0.2" to 0.75"	
	TL3: 0.05" to 0.2	
	TL4: 0.05" to 0.2	
Length Matching: address/command group (except CS, ODT and CKE lines) with respect to clock (from controller to SDRAM ball)	ADDR/CMD $\leq$ clock length + 2" ADDR/CMD $\geq$ clock length - 1"	
Breakout Trace Width and Spacing	5 mils x 5mils. Microstrip is preferred. Maximum length of the breakout trace is 500 mils.	
Split Termination	100 ohms +/- 1% to ground and 100 ohms +/- 1% to 1.8V	
Routing Guideline 1	Place the VTT terminations in the VTT island after the DIMM with a trace length of 0.15" to 0.5"	
Routing Guideline 2	For split terminations place the VTT termination in their respective power islands	
Routine Guideline 3	6 Vias or less	


**Table 57. Intel® 81341 and 81342 I/O processors Layout Checklist (Sheet 6 of 20)**

Checklist Items	Recommendations	Comments
<b>Embedded DDR2 CS, ODT, CKE Routing Topology</b>		<b>Refer to Section 4.2.4.3</b>
Reference Plane	Route over unbroken ground plane preferred	
Preferred Topology	Microstrip lines	
Trace Spacing (edge to edge) <sup>1</sup>	5 mils acceptable between the pins and the breakout regions. ≥ 12 mils within group ≥ 20 mils from any other clock/DQ/DQS groups.	
Trace Impedance	50 ohms +/- 15% for a motherboard	
Trace Length: Overall length from signal Ball to SDRAM ball	1" min to 12" maximum Refer to <a href="#">Table 18</a> for segment lengths.	
Trace Length Routing	Balanced Load Topology	
	TL Break out: ≤ 0.5"	
	TL0: 0.5" to 8"	
	TL1: 0.2" to 0.2"	
	TL3: 0.2" to 0.2"	
	TL4: 0.2" to 0.2"	
	TL5: 0.4" to 0.4"	
	TL6: 0.05" to 0.2"	
	Daisy Chain Topology:	
	TL Break out: ≤ 0.5"	
	TL0: 0.5" to 8"	
	TL1: 0.2" to 0.75"	
	TL3: 0.05" to 0.2"	
	TL4: 0.05" to 0.2"	
Length Matching: With respect to CS/CKE group	For daisy chain topology: when CS/CKE group length is < 4": CK length + 1" when CS/CKE group length is > 4": CK length + 3" For balanced segment topology: when CS/CKE group length is < 2": CK length + 1" when CS/CKE group length is > 2": CK length +/- 0.5"	
Breakout Trace Width and Spacing	5 mils x 5mils. Microstrip is preferred. Maximum length of the breakout trace is 500 mils.	
Split Termination	100 ohms +/- 1% to ground and 100 ohms +/- 1% to 1.8V	
Routing Guideline 1	Place the VTT terminations in the VTT island after the DIMM with a trace length of 0.15" to 0.5"	
Routing Guideline 2	For split terminations place the VTT termination in their respective power islands	
Routine Guideline 3	6 Vias or less	

**Table 57. Intel® 81341 and 81342 I/O processors Layout Checklist (Sheet 7 of 20)**

Checklist Items	Recommendations	Comments
<b>Embedded DDR2 CS, ODT, CKE Routing Daisy Chain Topology</b>		<b>Refer to Section 4.2.4.3</b>
Reference Plane	Route over unbroken ground plane preferred	
Preferred Topology	Microstrip lines	
Trace Spacing (edge to edge) <sup>1</sup>	5 mils acceptable between the pins and the breakout regions. ≥ 12 mils within group ≥ 20 mils from any other clock/DQ/DQS groups.	
Trace Impedance	50 ohms +/- 15% for a motherboard	
Trace Length: Overall length from signal Ball to SDRAM ball	1" min to 12" maximum Refer to Table 18 for segment lengths.	
Trace Length	TL Break out: ≤ 0.5"	
	TL0: 0.5" to 8"	
	TL1: 0.2" to 0.75"	
	TL3: 0.05" to 0.2"	
	TL4: 0.05" to 0.2"	
Length Matching: With respect to CS/CKE group	For daisy chain topology: when CS/CKE group length is < 4": CK length + 1" when CS/CKE group length is > 4": CK length + 3" For balanced segment topology: when CS/CKE group length is < 2": CK length + 1" when CS/CKE group length is > 2": CK length +/- 0.5"	
Breakout Trace Width and Spacing	5 mils x 5mils. Microstrip is preferred. Maximum length of the breakout trace is 500 mils.	
Split Termination	100 ohms +/- 1% to ground and 100 ohms +/- 1% to 1.8V	
Routing Guideline 1	Place the VTT terminations in the VTT island after the DIMM with a trace length of 0.15" to 0.5"	
Routing Guideline 2	For split terminations place the VTT termination in their respective power islands	
Routine Guideline 3	6 Vias or less	

**Table 57. Intel® 81341 and 81342 I/O processors Layout Checklist (Sheet 8 of 20)**

Checklist Items	Recommendations	Comments
<b>PCI Express for Motherboard Layout Recommendations (PETP[7:0]/PETN[7:0],PERP[7:0],PERN[7:0])</b>		<b>Refer to Section 5.2.1</b>
Reference Plane	Routing over unbroken ground plane is preferred. When unbroken ground plane is not available route over unbroken voltage plane.	
Trace Impedance	Single-ended: 50 ohms +/- 15% Differential: 85 ohms +/- 15%	
Microstrip Trace Width	5 mils	
Microstrip Trace Spacing (edge to edge)	between + and - : 7 mils Between other signals $\geq 25$ mils Transmit and Receive pairs are interleaved. For non interleaved pairs interpair spacing $\geq 45$ mils.	
Stripline Trace Width	5 mils	
Stripline Trace Spacing (edge to edge)	between + and - : 7 mils Between other signals $\geq 25$ mils Transmit and Receive pairs are interleaved. For non interleaved pairs interpair spacing $\geq 45$ mils.	
Length Matching	Total allowable pair mismatch on system board $\leq 10$ mils Length matched on a segment by segment basis.	
AC coupling capacitor	75 nF - 200 nF located at the transmitter	
Total Trace Length - (Transmitter/Receiver) from device signal pin to AC coupling capacitor and AC coupling capacitor to PCI Express device pin	1" - 30" max.	
Via counts	4 vias or less	

**Table 57. Intel® 81341 and 81342 I/O processors Layout Checklist (Sheet 9 of 20)**

Checklist Items	Recommendations	Comments
<b>PCI Express Baseboard (for Motherboard-Adapter Card) Layout Recommendations (PETP[7:0]/PETN[7:0],PERP[7:0],PERN[7:0])</b>		<b>Refer to Section 5.2.2</b>
Reference Plane	Routing over unbroken ground plane is preferred. When unbroken ground plane is not available route over unbroken voltage plane.	
Trace Impedance motherboard	Single -ended: 50 ohms +/- 15% Differential microstrip: 85 ohms +/- 15%	
Trace Impedance adapter card	Single Ended: 60 +/-15% ohms nominal Differential: 100 +/-15% ohms nominal	
Microstrip Trace Width	5 mils	
Microstrip Trace Spacing (edge to edge)	between + and - : 7 mils Between other signals $\geq$ 25 mils Transmit and Receive pairs are interleaved. When interleaving For non interleaved pairs interpair spacing $\geq$ 45 mils.	
Stripline Trace Width	5 mils	
Stripline Trace Spacing (edge to edge)	<ul style="list-style-type: none"> <li>Between + (P) and - (N) of pair: 7 mils</li> <li>Between other signals <math>\geq</math> 25 mils</li> <li>Transmit and Receive pairs are interleaved.</li> <li>For non interleaved pairs interpair spacing <math>\geq</math> 45 mils.</li> </ul>	
Length Matching	<ul style="list-style-type: none"> <li>Total allowable length skew between + and - signals of the pair trace mismatch for a lane that consists of a motherboard and an add-in card must not exceed 15 mils (10 mils motherboard and 5 mils adapter card).</li> <li>Total skew across all lanes must be less than 20 ns.</li> </ul>	
AC coupling capacitor	<ul style="list-style-type: none"> <li>75nF - 200 nF located at the transmitter</li> </ul>	
Total Length: Topology 1: Intel® 81341 and Intel® 81342 I/O Processors transmitter on motherboard with PCI-E device receiver on adapter card	<ul style="list-style-type: none"> <li>1.0" - 27" max</li> </ul>	
Total Length: Topology 2: transmitter on adapter card and the PCI-E device receiver on motherboard	<ul style="list-style-type: none"> <li>1.0" min. - 25" max.</li> </ul>	
Via counts	4 vias or less	


**Table 57. Intel® 81341 and 81342 I/O processors Layout Checklist (Sheet 10 of 20)**

Checklist Items	Recommendations	Comments
<b>PCI Express Clock Layout Recommendations (REFCLKP, REFCLKN)</b>		<b>Refer to Section 5.2.3</b>
Reference Plane	Routing over unbroken ground plane is preferred. When unbroken ground plane is not available route over unbroken voltage plane.	
Trace Impedance	Differential target: 100 ohm, tolerance +/-15% Single Ended: 50 ohms +/- 15%	
Trace Width	5 mils	
<b>REFCLKP, REFCLKN</b> differential Clock Pair Spacing	< 1.4 x Space Width	
Serpentine Spacing (spacing of clock lines from itself)	<ul style="list-style-type: none"> <li>spacing <math>\geq</math> 25 mils.</li> </ul>	
Clock to Other Spacing (edge to edge)	<ul style="list-style-type: none"> <li>Spacing from clock to other groups <math>\geq</math> 25 mils.</li> </ul>	
Trace Lengths <sup>2</sup>	L1, L1: 0.5" max	
	L2, L2: 0.2" max	
	L3, L3: 0.2" max	
	L4, L4 <ul style="list-style-type: none"> <li>Device down: 2" to 15.3"</li> </ul> or <ul style="list-style-type: none"> <li>Connector: 2" to 11.3</li> </ul>	
	Total Length = L1+L2+L4 <ul style="list-style-type: none"> <li>Device Down: 3" to 16"</li> </ul> or <ul style="list-style-type: none"> <li>Connector: 3" to 12"</li> </ul>	
Length Matching Requirements within differential pair	+/- 5 mils	
Rs Series Resistor	33 +/- 5% ohms	
Rt Shunt Resistor	49.9 +/- 1% ohms	
Number of Vias	4 max	



**Table 57. Intel® 81341 and 81342 I/O processors Layout Checklist (Sheet 11 of 20)**

Checklist Items	Recommendations	Comments
<b>PBI Interface (A[24:0], D[15:0]) One, Two and Three Loads</b>		<b>Refer to Section 7.4</b>
Reference Plane	Route over unbroken ground plane or unbroken power plane.	
Recommended Layer	Microstrip or stripline or combination	
Trace Impedance	Motherboard: 50 ohms +/- 15% Add-in Card: 60 ohms +/- 15%	
Trace Spacing (edge to edge)	$\geq 5$ mils between all Address and Data lines $\geq 20$ mils must be maintained from all other signals or vias.	
Breakout TL0	5 mils on 5 mils spacing. Maximum length of breakout region is 500mils.	
Trace Length TL1 single load	0" to 20.0"	
Trace Length TL1 multiple loads	2" to 20.0"	
Trace Length TL2, TL3	0.5" to 2.0" from the last device on the bus.	
Trace Length to strapping resistors TL4	0.5" to 3.0" from the last device on the bus.	
Routing Guideline	Route as daisy-chain only.	
Via counts	8 vias or less	

**Table 57. Intel® 81341 and 81342 I/O processors Layout Checklist (Sheet 12 of 20)**

Checklist Items	Recommendations	Comments
<b>PCI-X Routing Recommendations (Clocks P_CLK[0-3], PCLKIN, PCLKOUT)</b>		<b>Refer to Section 6.2</b>
Reference Plane	Route over unbroken ground plane.	
Recommended Layer	Stripline	
Trace Impedance: Motherboard	Microstrip: 50 ohm +/- 15%, stripline: 50 ohm +/- 10%	
Trace Impedance: Adapter Card	Microstrip or stripline: 60 ohm +/- 15%	
Trace Spacing (edge to edge)	between two different clock lines $\geq 25$ mils between two segments of the same clock line $\geq 25$ mils between clock and other signals $\geq 50$ mils	
Series Resistors	28 ohms 1% for connectors 26 ohms 1% for embedded	
Trace Length TL1 from buffer to the resistor	1.0" max	
Total Trace Length: from device ball to device (including resistor segment)	11" max	
Length Matching:	All clock lines including PCLKOUT to PCLKIN (feedback clock) must be matched to within 25 mils. Refer to Figure 31.	
Topologies with only embedded devices.	Match clocks to within 25 mils	
Topologies with only connectors .	Match clocks to within 25 mils. Route feedback clock longer to compensate for the adapter card length (2.4" to 2.6") + 0.85" (for the connector delay)	
Topologies with both slots and devices used in the design	Match Clocks to within 25 mils Route feedback clock longer to compensate for the adapter card length (2.4" to 2.6") + 0.85" (for the connector delay) PCLKs going to the embedded devices must be compensate for the adapter card length (2.4" to 2.6") + 0.85" (for the connector delay)	
Vias	$\leq 2$ vias	
<b>PCI-X Point to Point Signals (REQ#, GNT#)</b>		
Signal Group	REQ# and GNT# lines	
Reference Plane	Route over unbroken reference plane.	
Motherboard Impedance (microstrip)	50 ohm +/- 15% microstrip and 50 ohm +/- 10% stripline	
Motherboard Trace Spacing	14 mils microstrip and 12 mils stripline	
Add-in Card Impedance	60 ohm +/- 15% microstrip and stripline	
Add-in Card Trace Spacing	18 mils microstrip and 14 mils stripline	
Group Spacing: Spacing from other groups	25 mils minimum, edge to edge	
Trace Length TL1 - from buffer to the connector	0.5" min - 4.5" max 0.5" - 12.0" for 100MHz 0.5" - 12.0" for 66MHz	
Trace Length TL2 - from connector to the receiver	2.4" - 2.6" max	
Vias	$\leq 3$ vias	

**Table 57. Intel® 81341 and 81342 I/O processors Layout Checklist (Sheet 13 of 20)**

Checklist Items	Recommendations	Comments
<b>PCI-X 133 MHz Single Slot Topology (AD lines)</b>		
Signal Group	Address, data and control lines	
Reference Plane	Route over unbroken reference plane.	
Motherboard Impedance (microstrip)	50 ohm +/- 15%	
Motherboard Impedance (stripline)	50 ohm +/- 10%	
Motherboard Trace Spacing	14 mils microstrip 12 mils stripline	
Add-in Card Impedance	60 ohm +/- 15% microstrip and stripline	
Add-in Card Trace Spacing	14 mils microstrip and 12 mils stripline	
Group Spacing	Spacing from other groups: 25 mils minimum, edge to edge	
<b>Lower AD:</b> Trace Length TL1 - from SL ball to the connector	1.0" - 6.0" max	
<b>Lower AD:</b> Trace Length TL2 - from connector to the receiver	0.75" - 1.5" Max	
<b>Upper AD:</b> Trace Length TL1 - from SL ball to the connector	0.5" - 5.0" max	
<b>Upper AD:</b> Trace Length TL2 - from connector to the receiver	1.75" - 2.75" Max	
Vias	≤ 2 vias	
<b>PCI-X 133 MHz Embedded Topology (AD lines)</b>		
Signal Group	Address, data and control lines	
Reference Plane	Route over unbroken reference plane.	
Motherboard Impedance (microstrip)	50 ohm +/- 15%	
Motherboard Impedance (stripline)	50 ohm +/- 10%	
Motherboard Trace Spacing	14 mils microstrip 12 mils stripline	
Group Spacing edge to edge	Spacing from other groups: 25 mils minimum	
Trace Length TL1 - from SL ball to the junction	0.75" min - 2.5" max	
Trace Length TL2, TL4 from connector to the receiver	0.75" min - 2.5" Max	
Trace Length TL3 from junction to junction	0.75 "min. to 2.5" max	
Vias	≤ 3 vias	



**Table 57. Intel® 81341 and 81342 I/O processors Layout Checklist (Sheet 14 of 20)**

Checklist Items	Recommendations	Comments
<b>PCI-X 133 MHz Mixed Topology (AD lines)</b>		
Signal Group	Address, data and control lines	
Reference Plane	Route over unbroken reference plane.	
Motherboard Impedance (microstrip)	50 ohm +/- 15%	
Motherboard Impedance (stripline)	50 ohm +/- 10%	
Motherboard Trace Spacing	18 mils microstrip 14 mils stripline	
Add-in Card Impedance	60 ohm +/- 15% microstrip and stripline	
Add-in Card Trace Spacing	18 mils microstrip and 14 stripline	
Group Spacing edge to edge	Spacing from other groups: 25 mils minimum	
Lower AD: Trace Length TL1 - from SL ball to the junction	0.5" min. to 2.0" max	
Lower AD: Trace Length TL2 - from junction to AD1	0.5" min. to 2.0" max	
Lower AD: Trace Length TL3, from junction to CONN	0.5" min. to 3.5" max	
Lower AD: Trace Length TL4, from CONN to adapter	0.75" min. to 1.5" max	
Upper AD: Trace Length TL1 - from SL ball to the junction	0.5" min. to 2.0" max	
Upper AD: Trace Length TL2 - from junction to AD1	0.5" min. to 2.0" max	
Upper AD: Trace Length TL3, from junction to CONN	0.5" min. to 2.25" max	
Upper AD: Trace Length TL4, from CONN to adapter	1.75" min. to 2.75" max	
Vias	< 3 vias	

**Table 57. Intel® 81341 and 81342 I/O processors Layout Checklist (Sheet 15 of 20)**

Checklist Items	Recommendations	Comments
<b>PCI-X 100 MHz Slot Topology (AD lines)</b>		
Signal Group	Address/data and control lines	
Reference Plane	Route over unbroken reference plane.	
Motherboard Impedance (microstrip)	50 ohm +/- 15%	
Motherboard Impedance (stripline)	50 ohm +/- 10%	
Motherboard Trace Spacing	18 mils microstrip 14 mils stripline	
Add-in Card Impedance	60 ohm +/- 15% microstrip and stripline	
Add-in Card Trace Spacing	18 mils microstrip and 14 stripline	
Group Spacing edge to edge	Spacing from other groups: 25 mils minimum	
Lower AD: Trace Length TL1 - from ball to the junction	0.5" - 12.0" max	
Lower AD: Trace Lengths TL3 - Between connectors	0.5" - 3.0" max	
Lower AD: Trace Lengths TL2 - from connector to the first receiver, TL4 - from connector to the second receiver	0.75" - 1.50" max	
Upper AD: Trace Length TL1 - from ball to the junction	0.5" - 10.0" max	
Upper AD: Trace Lengths TL3 - Between connectors	0.5" - 3.0" max	
Upper AD: Trace Lengths TL2 - from connector to the first receiver, TL4 - from connector to the second receiver	1.75" - 2.75" max	
Vias	< 3 vias	
<b>PCI-X 100 MHz Embedded Topology (AD lines)</b>		
Signal Group	Address, data and control lines	
Reference Plane	Route over unbroken reference plane.	
Motherboard Impedance (microstrip)	50 ohm +/- 15%	
Motherboard Impedance (stripline)	50 ohm +/- 10%	
Motherboard Trace Spacing	18 mils microstrip 14 mils stripline	
Add-in Card Impedance	60 ohm +/- 15% microstrip and stripline	
Add-in Card Trace Spacing	14 mils microstrip and stripline	
Group Spacing	Spacing from other groups: 25 mils minimum, edge to edge	
Trace Length TL1 - from SL ball to the junction	0.5" min. to 3.0" max (3 loads, 5 loads)	
Trace Length TL3, TL5, TL7, TL9: from junction to junction	0.5" min. to 2.0" max (3 loads) 0.5" min. to 1.0" max (5 loads)	
Trace Length TL2, TL4, TL6, TL8, TL10: from junction to receiver	0.5" min. to 3.0" max (3 loads) 0.5" min to 2.0" max (5 loads)	
Vias	≤ 4 vias	


**Table 57. Intel® 81341 and 81342 I/O processors Layout Checklist (Sheet 16 of 20)**

Checklist Items	Recommendations	Comments
<b>PCI-X 100 MHz Mixed Topology (AD lines)</b>		
Signal Group	Address, data and control lines	
Reference Plane	Route over unbroken reference plane.	
Motherboard Impedance (microstrip)	50 ohm +/- 15%	
Motherboard Impedance (stripline)	50 ohm +/- 10%	
Motherboard Trace Spacing	18 mils microstrip and 14 mils stripline	
Add-in Card Impedance	60 ohm +/- 15% microstrip and stripline	
Add-in Card Trace Spacing	18 mils microstrip and 14 mils stripline	
Group Spacing	Spacing from other groups: 25 mils minimum, edge to edge	
Lower AD: Trace Length TL1 - from SL ball to the junction	0.5" min. to 2.5" max	
Lower Trace Length TL2 - from junction to AD1	0.5" min. to 2.0" max	
Lower Trace Length TL3, from junction to first CONN and TL5, from junction to second CONN	0.5" min. to 3.5" max	
Lower Trace Length TL4, from 1st CONN to AD2 Lower AD: Trace Length TL6, from 2nd CONN to AD3	0.75" min. to 1.5" max	
Upper AD: Trace Length TL1 - from SL ball to the junction	0.5" min. to 2.5" max	
Upper AD: Trace Length TL2 - from junction to AD1	0.5" min. to 2.0" max	
Upper AD: Trace Length TL3, from 1st junction to first CONN	0.5" min. to 3.0" max	
Upper AD: From 2nd junction to second CONN	0.5" min. to 3.5" max	
Upper AD: Trace Length TL4, from 1st CONN to AD2 Upper AD: Trace Length TL6, from 2nd CONN to AD3	1.75" min. to 2.75" max	
Vias	≤ 3 vias	

**Table 57. Intel® 81341 and 81342 I/O processors Layout Checklist (Sheet 17 of 20)**

Checklist Items	Recommendations	Comments
<b>PCI-X 66 MHz Slot Topology (AD lines)</b>		
Signal Group	Address/data and control lines	
Reference Plane	Route over unbroken reference plane.	
Motherboard Impedance (microstrip)	50 ohm +/- 15%	
Motherboard Impedance (stripline)	50 ohm +/- 10%	
Motherboard Trace Spacing	18 mils microstrip 14 mils stripline	
Add-in Card Impedance	60 ohm +/- 15% microstrip and stripline	
Add-in Card Trace Spacing	12 mils microstrip and 12 mils stripline	
Group Spacing	Spacing from other groups: 25 mils minimum, edge to edge	
Lower AD: Trace Length TL1 - from ball to the connector	0.5" - 12.0" max	
Lower AD: Trace Lengths TL3, TL5, TL7 - Between connectors	0.5" - 2.0" max	
Lower AD: Trace Lengths TL2, TL4, TL6, TL8 - from connector to the receivers	0.75" - 1.50" max	
Upper AD: Trace Length TL1 - from ball to the connector	0.5" - 9.0" max	
Upper AD: Trace Lengths TL3, TL5, TL7 - Between connectors	0.5" - 2.0" max	
Upper AD: Trace Lengths TL2, TL4, TL6, TL8 - from connector to the receivers	1.75" - 2.75" max	
Vias	≤ 4 vias	
<b>PCI-X 66 MHz Embedded Topology (AD lines)</b>		
Signal Group	Address/data and control lines	
Reference Plane	Route over unbroken reference plane.	
Motherboard Impedance (microstrip)	50 ohm +/- 15%	
Motherboard Impedance (stripline)	50 ohm +/- 10%	
Motherboard Trace Spacing	18 mils microstrip 14 mils stripline	
Group Spacing	Spacing from other groups: 25 mils minimum, edge to edge	
Trace Length TL1 - from SL ball to the junction	0.5" min. to 3.0" max (8 loads) 0.5" min. to 3.5" max (6 loads)	
Trace Length TL3, TL5, TL7, TL9, TL11, TL13, TL15: from junction to junction	0.5" min. to 1.5" max (8 loads) 0.5" min. to 2.5" max (6 loads)	
Trace Length TL2, TL4, TL6, TL8, TL10, TL12, TL14, TL16: from junction to receiver	0.5" min. to 1.5" max (8 loads) 0.5" min to 2.0" max (6 loads)	
Vias	≤ 4 vias	


**Table 57. Intel® 81341 and 81342 I/O processors Layout Checklist (Sheet 18 of 20)**

Checklist Items	Recommendations	Comments
<b>PCI-X 66 MHz Mixed Topology (AD lines)</b>		
Signal Group	Address/data and control lines	
Reference Plane	Route over unbroken reference plane.	
Motherboard Impedance (microstrip)	50 ohm +/- 15%	
Motherboard Impedance (stripline)	50 ohm +/- 10%	
Motherboard Trace Spacing	18 mils microstrip and 14 mils stripline	
Adapter Card Trace Impedance	60 ohm +/- 15% (microstrip and stripline)	
Adapter Card Trace Spacing	12 mils microstrip and mils stripline	
Group Spacing	Spacing from other groups: 25 mils minimum, edge to edge	
Lower AD: Trace Length TL1 - from SL ball to the junction	0.5" min. to 11" max	
Lower AD: Trace Length TL2, TL4 - from junction to AD1, AD2	0.5" min. to 4.5" max	
Lower AD: Trace Length TL3, TL5, TL7 from junction to junction	0.5" min. to 4.0" max	
Lower AD: Trace Length TL6 from 1st CONN to AD3, TL8: from 2nd CONN to AD4	0.75" min. to 1.5" max	
Upper AD: Trace Length TL1 - from SL ball to the junction	0.5" min. to 10" max	
Upper AD: Trace Length TL2, TL4 - from junction to AD1, AD2	0.5" min. to 4.0" max	
Upper AD: Trace Length TL3, TL5, TL7 from junction to junction	0.5" min. to 4.0" max	
Upper AD: Trace Length TL6 from 1st CONN to AD3, TL8: from 2nd CONN to AD4	1.75" min. to 2.75" max	
Vias	≤ 4 vias	



**Table 57. Intel® 81341 and 81342 I/O processors Layout Checklist (Sheet 19 of 20)**

Checklist Items	Recommendations	Comments
<b>VCC1P2PLLD - VSSPLL PCI-X PLL Filters</b>		<b>Refer to Section 11.4</b>
Reference Plane	Ground VCC1P2PLL, VCC1P2PLLD traces must be ground referenced (no $V_{CC}$ references)	
Inductor	4.7 $\mu$ H +/- 25% 45 mA L must be magnetically shielded ESR: max < 0.3 ohms rated at 45 mA	
Capacitor	22 $\mu$ F +/- 20% 6.3V (Capacitor) ESR: max < 0.3 ohms ESL < 2.5 nH Place 22 $\mu$ F capacitor as close as possible to package pin.	
Resistor	Rselect: choose resistor such that both of the following conditions are met: 1.2V plane to the top end of the capacitor is > 0.35 $\Omega$ (ινχλυδινγ βοαρδ ανδ χομπονεντ ρεσιςτανχε) 1.2V plane to $V_{CC1P2PLL}$ < 1.5 $\Omega$ (ινχλυδινγ βοαρδ ανδ χομπονεντ ρεσιςτανχε) 1/16 W 6.3 V resistor must be placed between $V_{CC1P2}$ and L. Note: when trace and component resistance is large enough a discrete resistor is not required	
Breakout Trace	Trace Width > 6 mils Trace Spacing < 6 mils Trace Length < 600 mils	
Board Trace	Trace Width > 25 mils Trace Spacing < 10 mils Trace Length < 600 mils	
Trace Spacing	$\geq$ 30 mils from any other signals.	
Trace Length maximum	1.2"	
Routing Guideline 1	Route VCC1P2PLLD and VSSPLL, VCC1P2PLL and VSSPLL as differential traces.	
Routing Guideline 2	The nodes connecting VCC1P2PLLD and the capacitor, VCC1P2PLL and the capacitor must be as short as possible.	



**Table 57. Intel® 81341 and 81342 I/O processors Layout Checklist (Sheet 20 of 20)**

Checklist Items	Recommendations	Comments
<b>VCC3P3PLLX - VSSPLLX PLL Filters</b>		
Reference Plane	Ground referenced VCC3P3PLL and VSSPLLX traces must be ground referenced (no V <sub>CC</sub> references)	
Inductor	4.7 $\mu$ H L must be magnetically shielded ESR: max < 0.4 ohms rated at 45 mA An example of this inductor is TDK part number MLZ20 12E4R7P.	
Capacitor	22 $\mu$ F 20% 6.3V (Capacitor) ESR: max < 0.4 ohms ESL < 3.0 nH Place 22 $\mu$ F capacitor as close as possible to package pin.	
Resistor	Rselect: choose resistor such that both of the following conditions are met: 3.3V plane to the top end of the capacitor is > 0.35 $\Omega$ 3.3V plane to V <sub>CC3P3PLL</sub> < 1.5 $\Omega$ resistor ratings: 1/16 W 6.3 V resistor must be placed between V <sub>CC3P3</sub> and L. Note: when trace and component resistance is large enough the discrete resistor is not required	
Breakout Trace	Trace Width > 6 mils Trace Spacing < 6 mils Trace Length < 600 mils	
Board Trace	Trace Width > 25 mils Trace Spacing < 10 mils Trace Length < 600 mils	
Trace Length maximum	1.2"	
Trace Spacing	$\geq$ 30 mils from any other signals.	



## 13.0 References

The following manuals and specifications are helpful in designing an application using Intel® 81341 and Intel® 81342 I/O Processors

### 13.1 Relevant Documents

- Intel® 81341 and Intel® 81342 I/O Processors Developer's Manual, Intel Corporation
- Intel® 81341 and Intel® 81342 I/O Processors Datasheet, Intel Corporation
- Intel® 81341 and Intel® 81342 I/O Processors *Thermal Application Note*, Intel Corporation
- *PCI Express Specification*, Revision 1.0a
- PCI Express Base Specification 1.0a
- PCI Express Card Electromechanical Specification 1.0a
- *PCI Local Bus Specification*, Revision 2.3 - PCI Special Interest Group
- *PCI-X Specification*, Revision 1.0b - PCI Special Interest Group
- *PCI Hot-Plug Specification*, Revision 1.0 - PCI Special Interest Group
- *PCI Bus Power Management Interface Specification*, Revision 1.1 - PCI Special Interest Group
- IEEE Standard Test Access Port and Boundary-Scan Architecture (IEEE JTAG-1149.1-1990)
- I2C Bus Specification version 2.1 at <http://www.semiconductors.philips.com/acrobat/literature/9398/39340011.pdf>
- SMBus Specification at <http://www.smbus.org/specs/>

**Table 58. Intel Related Documentation**

Document Title	Order #
Intel® Packaging Databook	240800



## 13.2 Design References

**Table 59. Design References**

Design References
<i>Transmission Line Design Handbook</i> , Brian C. Wadell
<i>Microstrip Lines and Slotlines</i> , K. C. Gupta. Et al.
<i>Design, Modeling and Simulation Methodology for High Frequency PCI-X Subsystems</i> , Moises Cases, Nam Pham, Dan Neal <a href="http://www.pcisig.com">www.pcisig.com</a>
<i>High-Speed Digital Design "A Handbook of Black Magic"</i> Howard W. Johnson, Martin Graham
"Terminating Differential Signals on PCBs", Steve Kaufer, Kelee Crisafulli, Printed Circuit Design, March 1999
"Board Design Guidelines for PCI Express™ Interconnect", <a href="http://www.intel.com/technology/pciexpress/downloads/PCI_EI_PCB_Guidelines.pdf">http://www.intel.com/technology/pciexpress/downloads/PCI_EI_PCB_Guidelines.pdf</a>

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Denver, CO 80217-9808  
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## 13.3 Electronic Information

**Table 60. Electronic Information**

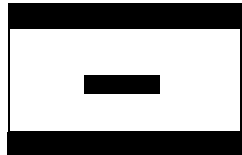

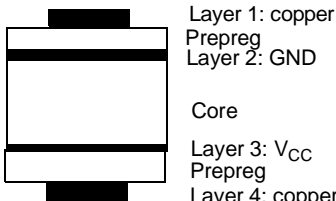
The Intel World-Wide Web (WWW) Location:	<a href="http://www.intel.com">http://www.intel.com</a>
Customer Support (US and Canada):	800-628-8686

## Appendix A Terminology, Definitions and Conditions

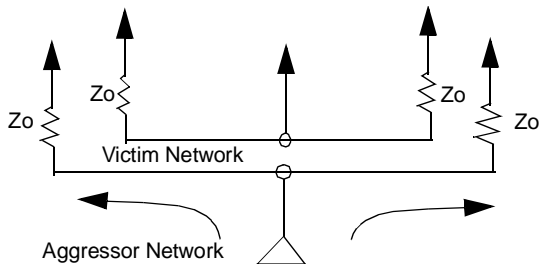
### A.1 Terminology

To aid the discussion of the Intel® 81341 and Intel® 81342 I/O Processors, [Table 61](#) provides the terminology used in this document.

**Table 61. Terminology and Definitions (Sheet 1 of 3)**

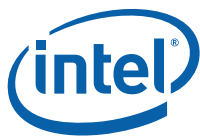
Term	Definition	
Stripline	 Side View	<p>Stripline in a PCB is composed of the conductor inserted in a dielectric with GND planes to the top and bottom.</p> <p>An easy way to distinguish stripline from microstrip is to strip away layers of the board to view the trace on stripline.</p>
Microstrip	 Side View	<p>Microstrip in a PCB is composed of the conductor on the top layer above the dielectric with a ground plane below</p>
Prepreg	Material used for the lamination process of manufacturing PCBs. It consists of a layer of epoxy material that is placed between two cores. This layer melts into epoxy when heated and forms around adjacent traces.	
Core	Material used for the lamination process of manufacturing PCBs. This material is two sided laminate with copper on each side. The core is an internal layer that is etched.	
PCB	 <p>Example of a Four-Layer Stack</p>	<p>Printed circuit board.</p> <p>Example manufacturing process consists of the following steps:</p> <ul style="list-style-type: none"> <li>Consists of alternating layers of core and prepreg stacked</li> <li>The finished PCB is heated and cured.</li> <li>The via holes are drilled</li> <li>Plating covers holes and outer surfaces</li> <li>Etching removes unwanted copper</li> <li>Board is tinned, coated with solder mask and silk screened</li> </ul>
DDR	Double Data Rate Synchronous DRAM. Data clocked on both rising and falling edges of clock.	
DDR2	DDR2 is backward compatible with DDR I. It allows 4.3GB/bytes/sec. for a clock rate of 533MHz and 3.2GB/bytes/sec. for a clock rate of 400 MHz.	
DIMM	Dual Inline Memory Module	
Source Synchronous DDR	With source-synchronous DDR interfaces, data and clock transport from a transmitter to a receiver, and the receiver interface uses the clock to latch the accompanying data.	
SSTL_2	Series Stub Terminated Logic for 2.5 V	

**Table 61. Terminology and Definitions (Sheet 2 of 3)**

Term	Definition
JEDEC	Provides standards for the semiconductor industry.
DLL	Delay Lock Loop - DDR feature used to provide appropriate strobe delay to clock in data.
PLL	Phase Lock Loop - A phase-locked loop (PLL) is an electronic circuit with a voltage- or current-driven oscillator that is constantly adjusted to match in phase (and thus lock on) the frequency of an input signal.
Aggressor	<p>A network that transmits a coupled signal to another network is aggressor network.</p> 
Victim	A network that receives a coupled cross-talk signal from another network is a victim network.
Network	The trace of a PCB that completes an electrical connection between two or more components.
Stub	Branch from a trunk terminating at the pad of an agent.
ISI	<p>Intersymbol Interference (ISI). This occurs when a transition that has not been completely dissipated, interferes with a signal being transmitted down a transmission line. ISI impacts both the timing and signal integrity. It is dependent on frequency, time delay of the line and the reflection coefficient at the driver and receiver. Examples of ISI patterns that are used in testing at the maximum allowable frequencies are the sequences shown below:</p> <p>01010101010101  0011001100110011  0001110001110001111</p>
CRB	Customer Reference Board
PC1600	<p>JEDEC Names for DDR based on peak data rates.</p> <p>PC1600= clock of 100 MHz * 2 data words/clock * 8 bytes = 1600 MB/sec.</p>
PC2100	<p>JEDEC Names for DDR based on peak data rates.</p> <p>PC2100= clock of 133 MHz * 2 data words/clock * 8 bytes = 2128 MB/sec.</p>
PC2700	<p>JEDEC Names for DDR2 based on peak data rates.</p> <p>PC2700= clock of 167 MHz * 2 data words/clock * 8 bytes = 2672 MB/sec</p>
PC3200	<p>JEDEC Names for DDR2 400 based on peak data rates.</p> <p>PC3200= clock of 200 MHz * 2 data words/clock * 8 bytes = 3200 MB/sec  clock of 266 * 2 data words/clock * 8 bytes =</p>
PC4300	<p>JEDEC Names for DDR2 533 based on peak data rates.</p> <p>PC4300= clock of 266 MHz * 2 data words/clock * 8 bytes = 4256 MB/sec</p>
Host processor	Processor located upstream from the Intel® 81341 and Intel® 81342 I/O Processors
Local processor	Intel XScale® processor within Intel® 81341 and Intel® 81342 I/O Processors
Downstream	<ul style="list-style-type: none"> <li>• PCI Express: At or toward a PCI Express port directed away from root complex (to a bus with a higher number).</li> <li>• PCI-X: At or toward a PCI bus with a higher number (after configuration) away from host processor.</li> </ul>
Upstream	<ul style="list-style-type: none"> <li>• PCI Express: At or toward a PCI Express port directed to the PCI Express root complex (to a bus with a lower number).</li> <li>• PCI-X: At or toward a PCI bus with a higher number (after configuration) toward host processor.</li> </ul>

**Table 61. Terminology and Definitions (Sheet 3 of 3)**

Term	Definition
Local memory	Memory subsystem on the Intel XScale® processor DDR SDRAM or Peripheral Bus Interface busses.
WORD	16-bits of data.
DWORD	32-bit data word.
QWORD	64-bit data word
Local bus	Internal Bus.
Outbound	At or toward the PCI interface of the ATU from the Internal Bus.
Inbound	At or toward the Internal Bus from the PCI interface of the ATU.
Core processor	Intel XScale® processor within the part.
Flip Chip	FC-BGA (flip chip-ball grid array) chip packages are designed with core flipped up on the back of the chip, facing away from the PCB. This allows more efficient cooling of the package.
Mode Conversion	Mode Conversions are due to imperfections on the interconnect which transform differential mode voltage to common mode voltage and common mode voltage to differential voltage.
ROMB	Raid on motherboard
ODT	On Die Termination - eliminates the need for termination resistors by placing the termination at the chip.



## A.2 Right Angle Connector DDR2 Skews for Length Matching

Use [Table 62](#) to compensate for the length differences for the right angle connector when performing length matching. Refer to [Section 4.2.2.1](#) for additional information.

For example when compensating for the length difference between CB0, CB1 two choices either:

1. CB0 length is not adjusted but subtract 75 from CB1 length.
2. add a constant value of 285 to CB0 and a constant value of 210 to CB1

When chosen method was number 1 then the entire "Shorter by" column is used to compensate for length for the entire connector. When the choose method was number 2 then the entire "Longer by" column is used to compensate for length for the entire connector.

**Note:** The rows that are shaded in [Table 62](#) are not critical signals and do not have to be length compensated.

**Table 62. Right Angle Connector Skews (length matching compensation) (Sheet 1 of 7)**

Connector Pin	Signal Name	Length Skews (choose one column below)	
		Shorter by (mils)	Longer by (mils)
42	CB0	0	285
43	CB1	75	210
48	CB2	0	285
49	CB3	75	210
161	CB4	285	0
162	CB5	210	75
167	CB6	285	0
168	CB7	210	75
185	CK0	285	0
186	CK0_N	210	75
137	CK1_RFU	285	0
138	CK1_RFU_N	210	75
220	CK2_RFU	210	75
221	CK2_RFU_N	285	0
125	DM0_DQS9	285	0
134	DM1_DQS10	210	75
146	DM2_DQS11	210	75
155	DM3_DQS12	285	0
202	DM4_DQS13	210	75
211	DM5_DQS14	285	0
223	DM6_DQS15	285	0
232	DM7_DQS16	210	75
164	DM8_DQS17	210	75
3	DQ0	75	210
4	DQ1	0	285



**Table 62. Right Angle Connector Skews (length matching compensation) (Sheet 2 of 7)**

Connector Pin	Signal Name	Length Skews (choose one column below)	
		Shorter by (mils)	Longer by (mils)
21	DQ10	75	210
22	DQ11	0	285
131	DQ12	285	0
132	DQ13	210	75
140	DQ14	210	75
141	DQ15	285	0
24	DQ16	0	285
25	DQ17	75	210
30	DQ18	0	285
31	DQ19	75	210
9	DQ2	75	210
143	DQ20	285	0
144	DQ21	210	75
149	DQ22	285	0
150	DQ23	210	75
33	DQ24	75	210
34	DQ25	0	285
39	DQ26	75	210
40	DQ27	0	285
152	DQ28	210	75
153	DQ29	285	0
10	DQ3	0	285
158	DQ30	210	75
159	DQ31	285	0
80	DQ32	0	285
81	DQ33	75	210
86	DQ34	0	285
87	DQ35	75	210
199	DQ36	285	0
200	DQ37	210	75
205	DQ38	285	0
206	DQ39	210	75
122	DQ4	210	75
89	DQ40	75	210
90	DQ41	0	285
95	DQ42	75	210
96	DQ43	0	285
208	DQ44	210	75
209	DQ45	285	0



**Table 62. Right Angle Connector Skews (length matching compensation) (Sheet 3 of 7)**

Connector Pin	Signal Name	Length Skews (choose one column below)	
		Shorter by (mils)	Longer by (mils)
214	DQ46	210	75
215	DQ47	285	0
98	DQ48	0	285
99	DQ49	75	210
123	DQ5	285	0
107	DQ50	75	210
108	DQ51	0	285
217	DQ52	285	0
218	DQ53	210	75
226	DQ54	210	75
227	DQ55	285	0
110	DQ56	0	285
111	DQ57	75	210
116	DQ58	0	285
117	DQ59	75	210
128	DQ6	210	75
229	DQ60	285	0
230	DQ61	210	75
235	DQ62	285	0
236	DQ63	210	75
129	DQ7	285	0
12	DQ8	0	285
13	DQ9	75	210
7	DQS0	75	210
6	DQS0_N	0	285
16	DQS1	0	285
15	DQS1_N	75	210
28	DQS2	0	285
27	DQS2_N	75	210
37	DQS3	75	210
36	DQS3_N	0	285
84	DQS4	0	285
83	DQS4_N	75	210
93	DQS5	75	210
92	DQS5_N	0	285
105	DQS6	75	210
104	DQS6_N	0	285
114	DQS7	0	285
113	DQS7_N	75	210

**Table 62. Right Angle Connector Skews (length matching compensation) (Sheet 4 of 7)**

Connector Pin	Signal Name	Length Skews (choose one column below)	
		Shorter by (mils)	Longer by (mils)
46	DQS8	0	285
45	DQS8_N	75	210
193	S0_N	285	0
76	S1_N	0	285
188	A0	210	75
183	A1	285	0
70	A10_AP	0	285
57	A11	75	210
176	A12	210	75
196	A13	210	75
174	A14	210	75
173	A15	285	0
54	A16_BA2	0	285
63	A2	75	210
182	A3	210	75
61	A4	75	210
60	A5	0	285
180	A6	210	75
58	A7	0	285
179	A8	285	0
177	A9	285	0
71	BA0	75	210
190	BA1	210	75
74	CAS_N	0	285
52	CKE0	0	285
171	CKE1	285	0
135	NC_DQS10_N	285	0
147	NC_DQS11_N	285	0
156	NC_DQS12_N	210	75
203	NC_DQS13_N	285	0
212	NC_DQS14_N	210	75
224	NC_DQS15_N	210	75
233	NC_DQS16_N	285	0
165	NC_DQS17_N	285	0
126	NC_DQS9_N	210	75
19	NC0	75	210
68	NC1	0	285
102	NC2	0	285
195	ODT0	285	0


**Table 62. Right Angle Connector Skews (length matching compensation) (Sheet 5 of 7)**

Connector Pin	Signal Name	Length Skews (choose one column below)	
		Shorter by (mils)	Longer by (mils)
77	ODT1	75	210
192	RAS_N	210	75
55	RC0	75	210
18	RESET_N	0	285
239	SA0	285	0
240	SA1	210	75
101	SA2	75	210
120	SCL	0	285
119	SDA	75	210
53	VDD0	75	210
59	VDD1	75	210
197	VDD10	285	0
64	VDD2	0	285
67	VDD3	75	210
69	VDD4	75	210
172	VDD5	210	75
178	VDD6	210	75
184	VDD7	210	75
187	VDD8	285	0
189	VDD9	285	0
194	VDDQ0	210	75
51	VDDQ1	75	210
191	VDDQ10	285	0
56	VDDQ2	0	285
62	VDDQ3	0	285
72	VDDQ4	0	285
75	VDDQ5	75	210
78	VDDQ6	0	285
170	VDDQ7	210	75
175	VDDQ8	285	0
181	VDDQ9	285	0
238	VDDSPD	210	75
1	VREF	75	210
2	VSS0	0	285
5	VSS1	75	210
32	VSS10	0	285
35	VSS11	75	210
38	VSS12	0	285
41	VSS13	75	210

**Table 62. Right Angle Connector Skews (length matching compensation) (Sheet 6 of 7)**

Connector Pin	Signal Name	Length Skews (choose one column below)	
		Shorter by (mils)	Longer by (mils)
44	VSS14	0	285
47	VSS15	75	210
50	VSS16	0	285
65	VSS17	75	210
66	VSS18	0	285
79	VSS19	75	210
8	VSS2	0	285
82	VSS20	0	285
85	VSS21	75	210
88	VSS22	0	285
91	VSS23	75	210
94	VSS24	0	285
97	VSS25	75	210
100	VSS26	0	285
103	VSS27	75	210
106	VSS28	0	285
109	VSS29	75	210
11	VSS3	75	210
112	VSS30	0	285
115	VSS31	75	210
118	VSS32	0	285
121	VSS33	285	0
124	VSS34	210	75
127	VSS35	285	0
130	VSS36	210	75
133	VSS37	285	0
136	VSS38	210	75
139	VSS39	285	0
14	VSS4	0	285
142	VSS40	210	75
145	VSS41	285	0
148	VSS42	210	75
151	VSS43	285	0
154	VSS44	210	75
157	VSS45	285	0
160	VSS46	210	75
163	VSS47	285	0
166	VSS48	210	75
169	VSS49	285	0



**Table 62. Right Angle Connector Skews (length matching compensation) (Sheet 7 of 7)**

Connector Pin	Signal Name	Length Skews (choose one column below)	
		Shorter by (mils)	Longer by (mils)
17	VSS5	75	210
198	VSS50	210	75
201	VSS51	285	0
204	VSS52	210	75
207	VSS53	285	0
210	VSS54	210	75
213	VSS55	285	0
216	VSS56	210	75
219	VSS57	285	0
222	VSS58	210	75
225	VSS59	285	0
20	VSS6	0	285
228	VSS60	210	75
231	VSS61	285	0
234	VSS62	210	75
237	VSS63	285	0
23	VSS7	75	210
26	VSS8	0	285
29	VSS9	75	210
73	WE_N	75	210



## A.3 Simulation Conditions

This section provides the simulation conditions that were used for this level of analysis for each of the interfaces.

### A.3.1 DDR2 Simulation Conditions

The following list provides the DDR2 simulation conditions used in this analysis:

- Motherboard 50 ohm single ended impedance stackup +/- 15% tolerance and 60 ohm single ended impedance stackup +/- 15%
- Motherboard clock target differential impedance 85 ohms +/- 15% and adapter card differential impedance of 100 ohms +/- 15%
- One Die Termination - ODT value of 75W was assumed for all DDR2 simulations.
- Generic DDR2 memory model
- DIMM models and topologies used the JEDEC model as a reference.
- JEDEC standard recommendations were used as a reference.
- Vias are modeled for all topologies with equal number of vias for differential pair
- Timing analysis was conducted.
- ISI Pattern was simulated.
- Signal Quality analysis covered for Rising flight time, Falling flight time, Low to high ring-back (noise margin high), High to Low ring-back (noise margin Low), and Low and High Overshoot.
- Frequency: 266MHz (DDR2 533MT/s)
- DIMM card microstrip routing is specified by JESD21-C.
- The ODT value used for simulations was 75Ω. Note that this value must be programmed for both the IOP and the SDRAM locations.

### A.3.2 PCI-X Simulation Conditions

The following list provides the PCI-X simulation conditions used in this analysis:

- Simulations were done for 133 MHz, 100 MHz and 66MHz.
- Various combinations of stripline and microstrip routing were analyzed.
- Vias and connectors were modeled using some estimated L and C parasitic values based on previous projects, or commonly used values from the literature.
- Connector Model: distributed PCI/PCI-X connector model
- PCI-X Package Model - Generic PCI-X spec device model
- SL Package Model: ball (RLC) + 1 via (RLC) + Stripline (W element) + 3 via + 1PTH (plated through hole RLC) + 4 via + 1 Ball.
- Motherboard trace: Impedance 50 ohm +/- 15% for stripline.
- Adapter Card Trace: Impedance 60 ohm +/- 15% for both microstrip and stripline.



### A.3.3 PCI Express Simulation Conditions

The following list provides the SAS simulation conditions used in this analysis:

- Jitter and insertion loss budgets used per PCI Express Specifications
- AC coupling capacitors 75 nF with low ESL and ESR
- Both receiver and transmitter eyes were evaluated for the PCI Express mask specifications
- Modified worst case ISI pattern (8b/10b was used)
- Both near end and far end crosstalk were taken into consideration
- SSO simulated but the impact was found to be not significant.

### A.3.4 PBI Simulation Conditions

The following list provides the PBI simulation conditions used in this analysis:

- System Board Stack up: 50 ohm +/- 15%, single ended impedance
- Add-In Card Stack up: 60 ohm +/- 15% single ended impedance
- Flash Model: RC128J3A
- Latch Model: 74LVC573A
- CPLD Model: XC9500XL TQFP package
- NVRAM Model: Same as flash
- Lossy un-coupled transmission lines were used in simulations.
- Trace spacings were set to three times the height of the trace over the reference plane to avoid crosstalk
- Up to 200ns of cycles for AD lines are examined for every topology and are assumed to be equivalent to subsequent cycles.